



MCF5407SWRM/D 1/2003 REV 0

# MCF5407 Low Data Rate Soft Modem Software Developer's Reference Manual

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#### **Tables**

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#### **About This Book**

The primary objective of this user's manual is to define the functionality of the MCF5407 processors for use by software developers.

The information in this book is subject to change without notice, as described in the disclaimers on the title page of this book. As with any technical documentation, it is the readers' responsibility to be sure he is using the most recent version of the documentation.

To locate any published errata or updates for this document, refer to the world-wide web at http://www.motorola.com/coldfire.

#### **Audience**

This manual is intended for system software developers who want to develop products with the MCF5407. It is assumed that the reader understands operating systems, microprocessor system design, basic principles of software and hardware, and basic details of the ColdFire<sup>®</sup> architecture.

#### **Suggested Reading**

This section lists additional reading that provides background for the information in this manual as well as general information about the ColdFire architecture.

#### **General Information**

The following documentation provides useful information about the ColdFire architecture and computer architecture in general:

#### **ColdFire Documentation**

The ColdFire documentation is available from the sources listed on the back cover of this manual. Document order numbers are included in parentheses for ease in ordering.

- ColdFire Programmers Reference Manual, R1.0 (MCF5200PRM/AD)
- ColdFire MCF5407 User's Manual, 0.1 (MCF5407UM/AD)
- Motorola M5407C3 User's Manual, R1.1 (M5407C3UM/AD).
- Silicon Laboratories Si3044 Data Sheet, R2.01 (Si3044-DS201)

#### **Acronyms and Abbreviations**

Table i lists acronyms and abbreviations used in this document.

**Table i. Acronyms and Abbreviated Terms** 

Term	Meaning
AGC	Automatic Gain Control
AFE	Analogue Front End
ANS	Answer tone (see V.25)
BPF	Bandpass filter
DAA	Data Access Arrangement
DTMF	Dual Tone Multi Frequency Signaling
DCE	Data Circuit-terminating Equipment
DTE	Data Terminal Equipment
DPSK	Differential PSK
EC	Echo Canceling
FIR	Finite Impulse Response
FSK	Frequency Shift Keying
FDM	Frequency Division Multiplexing
GSTN	General Switched Telephone Network
IIR	Infinite Impulse Response
LDR	Low Data Rate
LPF	Low-pass filter
HSP	Host Signal Processing
PSK	Phase Shift Keying
PSTN	Public Switched Telephone Network
SM	Soft Modem
QAM	Quadrature Amplitude Modulation
QDPSK	Quadrature DPSK
UART	Universal asynchronous receiver / transmitter
USART	Universal synchronous/asynchronous receiver transmitter

## Chapter 1 Introduction

This document describes the low data rate (LDR) Soft Modem (with data rates up to 2400 bits / second) for the MCF5407 ColdFire processor. The Soft Modem includes support for the Data Pump in real-time software, as well as the AT command set and other protocols (see below). The chosen development platform is the MCF5407C3 EVM, with an additional Daughter card. The Daughter card is used to provide the Data Access Arrangement (codec and phone line interface) such as Silicon Labs Si3044.

In this manual, you will find the information required to use and maintain the LDR Soft Modem interfaces and algorithms.

#### 1.1 Key Features

The LDR Soft Modem supports:

• Table 1-1 lists Data Pump Protocols implemented in the Soft Modem.

Table 1-1. Data Pump Protocols Supported by Soft Modem

ITU-T recommendation	Modulation	Data Rate (bps)	Modulation Rate (bauds)	Carrier Frequency (Hz)	Bits per Baud
V.22bis	QAM	2400	600	1200/2400	4
		1200	600	1200/2400	2
V.22	DPSK	1200	600	1200/2400	2
		600	600	1200/2400	1
V.23	FSK	75/1200	75/1200	420/1700	1
		75/600	75/600	420/1500	1
V.21	FSK	300	300	1080/1750	1

- ITU-T V.21, V.23, V.22 and V22bis Physical Handshake Sequences.
- ITU-T V.42 Error Correction.
- ITU-T V.42bis Data Compression.
- ITU-T V.14 sync-to-asynchronous converter.

#### Related Files

- AT-command set.
- Tone generation and detection.
- DTMF and Pulse dialing.
- DAA interface (Si3044 Modem Daughter Card) via the on-chip USART module.
- DTE interface via the on-chip UART module.

#### 1.2 Related Files

The following files are relevant to the Soft Modem:

- main.c entry point of the Soft Modem.
- modem. h high level functions and main data structure definitions.
- modem. c implementation of high level functions.
- state machine.h definition of state machine data structure.
- state machine.c implementation of main state machine.
- *v21.h* definitions used by V.21 protocol.
- *v21.c* initialization of V.21 data pump protocol and realization of physical handshake sequence for V.21.
- v23.h definitions used by V.23 protocol.
- v23.c initialization of V.23 data pump protocol and realization of physical handshake sequence for V.23.
- *v22.h* definitions used by V.22 protocol.
- *v22.c* initialization of V.22 data pump protocol and realization of physical handshake sequence for V.22.
- *v22bis.h* definitions used by V.22bis protocol.
- *v22bis.c* initialization of V.22bis data pump protocol and realization of physical handshake sequence for V.22bis.
- fsk.h FSK data pump definitions.
- fsk.c realization of FSK data pump.
- dpsk.h DPSK data pump definitions.
- dpsk.c realization of 2,4-DPSK data pump.
- qam.h QAM data pump definitions.
- qam. c realization of 4,16-QAM data pump.
- v14.h definitions used by V.14 protocol.
- v14.c realization of V.14 data handler.
- *V42.h* definitions used by V.42 protocol.
- *V42.c* realization of V.42 protocol.

- *V42bis.h* definitions used by V.42bis protocol.
- *V42bis.c* realization of V.42bis protocol.
- *v*8.*h* definitions used by V.8 protocol.
- *vs.c* realization of V.8 protocol.
- dtmf.h DTMF dialer definitions.
- dtmf.c DTMF dialer realization.
- pulse.h pulse dialer definitions.
- pulse.c pulse dialer realization.
- tone gendet.h definitions used by tone generator and detector.
- tone gendet.c realization of tone generator and detector.
- ring det.h definitions used by ring detector.
- ring\_det.c realization of ring detector.
- at parser.h definitions used by AT command parser.
- at\_parser.c realization of AT command parser.
- at\_handler.h definitions used by AT command handler.
- at handler.c realization of AT command handler.
- text\_response.h definitions used by text response routines according to V.25ter.
- text\_response.c realization of text response routines according to V.25ter.
- countries.h definitions of country or area codes according to ITU-T T.35.
- misc.h definitions used by miscellaneous functions.
- misc.c realization of miscellaneous functions.

#### The following files are HW and Compiler dependent parts of the Soft Modem:

- softMod.mcp Soft Modem project file.
- *init.h* general data types and definitions. Init function prototypes.
- si3044\_daa.h definitions used by the DAA (Si3044 Modem Daughter Card) interface.
- si3044 daa.c realization of the DAA (Si3044 Modem Daughter Card) interface.
- mcf5407 uart.h definitions used by UART module routines.
- mcf5407 uart.c realization of UART module routines.
- mcf5407 timer.h definitions used by Timer module routines.
- mcf5407 timer.c realization of Timer module routines.
- printf uart.h definitions used by printf\_uart() routine.
- printf uart.c realization of printf\_uart() routine. Only for debug requirements.
- mcf5407.h MCF5407 definitions.

- mcf5407 lo.s lowest level routines for the MCF5407.
- *vector.s* MCF5407 vector table.
- int handlers.c interrupt handlers.
- sysinit.c power-on reset configuration of the MCF5407.
- ads 68K mw.c linker command file.
- mwerks.h defines constants used by the CodeWarrior Preprocessor.

#### 1.3 Quick Start

Carefully follow these steps to prepare the LDR Soft Modem for operation

• Properly install and set-up the Soft Modem daughter card:

The daughter card can be installed on the M5407C3 evaluation board the wrong way. Please ensure that the daughter card is correctly fitted before applying power. The Motorola logos on the two cards must be read the opposite way up to each other for correct operation. When correctly installed, the RJ11 connector will be towards the PCI socket of the main board and the speaker will be adjacent to the DIMM memory card.

Figure 1-1 shows the correctly installed daughter card and its jumper settings for correct communication with the M5407C3 evaluation board.

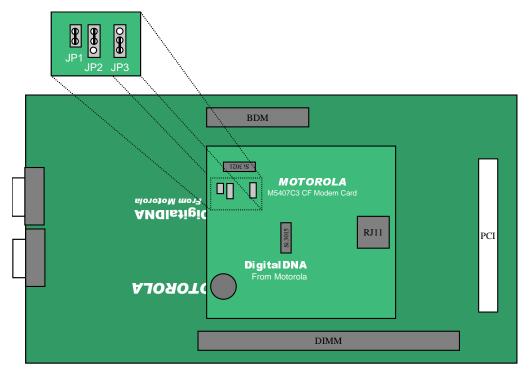


Figure 1-1. Modem Daughter Card Installation. Jumper Settings.

• Proper set-up of the M5407C3 evaluation board:

To setup the M5407C3 evaluation board, refer to the M5407C3 User's Manual, R1.1. When you are preparing the board for the first time, be sure to check that all jumpers are in the correct locations. In practice, the locations of J21, J22, J23 and J24 are incorrect by default. The settings can be

determined by allowing the board to boot up under dBUG control. Since dBUG drives the configuration data on PP[0:3] the data will be seen on the LEDs (D1-D4). If the LED is on then the corresponding jumper should be OFF, and if the LED is off then the jumper should be in position 2-3 (see *M5407C3 User's Manual*, *R1.1*. Chapter 3.1.12 SDRAM DIMM ).

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- Use the RS-232 male/female DB-9 serial cable to connect the PC to the M5407C3.
- Plug in the Modem daughter card to the telephone line through the RJ-11 socket (if you want to make a connection to another modem).
- Power-up the board.
- Invoke the Hyper Terminal or similar terminal program on the PC to which the evaluation board is connected and configure it to:
  - Bits per second: 19200
  - Data bits: 8Parity: noneStop bits: 1
  - Flow control: Hardware (CTS/RTS)
- Load ...\ SoftMod.mcp onto Metrowerks' CodeWarrior and build the target.
- Run project.
- In your terminal program window you should see the Soft Modem welcome message:

LDR Soft Modem

The Soft Modem is now ready to accept your AT commands (see MCF5407 Low Data Rate Soft Modem. AT Command reference, User Manual.)

## **Chapter 2 General Architecture**

The Low Data Rate Soft Modem runs on the M5407C3 boards with an installed Modem daughter card based on the Silicon Labs Si3044 DAA. The current version of the LDR Soft Modem works as a stand-alone application without the need for an operating system (see Fig 2). The two functions performed by the modem are:

- Modem Data Pump functions the modulation/demodulation functions.
- Modem Control functions error correction, hardware compression and AT command interpretation.

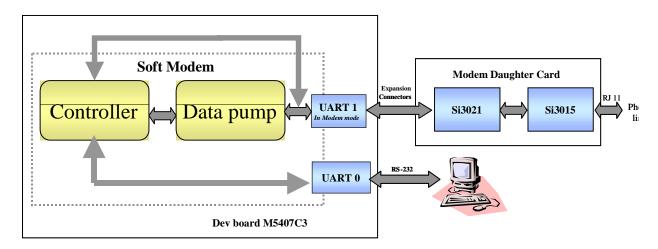


Figure 2-1. Basic Architecture of the Standalone Version of the LDR Soft Modem

All Controller and Data Pump functionality is performed using the ColdFire MCF5407 microprocessor, with no additional DSP processor. In general the MCF5407 performs the following functions:

- Modulation/demodulation
- Data encoding/decoding
- Filtering
- Automatic gain control
- Tone dialing
- Call progress monitoring

- AT command set interpretation
- Scrambling / descrambling
- Dialing
- Synchronous/asynchronous conversion
- Modem configuration control
- Protocol initialization

The Modem Daughter Card is based on the Silicon Laboratories Si3044 Data Access Arrangement (DAA) (combined Si3021 and Si3015 chipset shown in Fig. 2), this provides a programmable line interface to meet global telephone line interface requirements (see Si3044 Data Sheet, R2.01 (Si3044-DS201)).

Modem communication is a real-time process where data must be communicated in both directions within a fixed period of time. If the modem does not meet real-time requirements, then the modem might drop the connection. In the SM, data is sampled at 9.6KHz, so the SM must process data every 0.00010416 Seconds. The LDR Soft Modem is capable of performing a full suite of modem functions in real-time on the MCF5407. Data buffering is used to accommodate stress conditions. The modem accumulates a number of samples in a buffer and processes those samples while gathering the next set of samples.

Code optimization is also important, as optimization has architectural dependencies. The most critical parts of the source code have been rewritten in assembler, taking into account the abilities of the MCF5407 (especially the MAC module). These assembly implementations also include a C interface. To enable the optimized assembler source code, it is required to define USE\_OPTIMIZED\_ASSEMBLER\_CODE (By default this is defined in "init.h").

The most important benefit of using the SM is the ability to upgrade the modem purely via software. Upgrades are typically provided to add new features, fix bugs or enhance modem performance.

The features of the LDR Soft Modem architecture are:

- Flexible addition of new Data Pump and Data Handler protocols with minimum modification required to the current source code.
- There is a potential for use in multi-channel systems.
- The data interface between the Soft Modem modules uses a set of Circular Buffers. It is better to implement circular buffers for the interface. This approach insulates the modem blocks running at different synchronous and asynchronous rates.
- There is a well-defined division of data processing between the Data Pump, the Data Handler, and others modules of the Soft Modem.
- Separate modules can work synchronously by detecting the level to which the Circular Buffers are filled.

#### 2.1 The Basic Blocks of the Soft Modem

This section is important for understanding how the LDR Soft Modem works.

Figure 2-2 shows the Basic Blocks of the Soft Modem and the general flow of data between them.

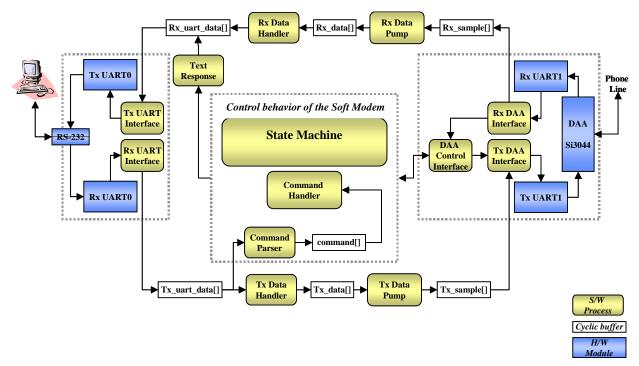


Figure 2-2. The Data Flow Diagram

The Soft Modem uses a set of seven circular buffers, Tx\_uart\_data[], Tx\_data[], Tx\_sample[], Rx\_uart\_data[], Rx\_sample[] and command[], as data interfaces between the functional blocks of the Soft Modem. Circular in this instance, means that the elements are accessed sequentially until the end of the buffer, then wrap around to the beginning again.

#### Block descriptions:

- **Tx\_sample**[] is a circular buffer that the Tx Data Pump writes the samples into. The Tx DAA Interface reads these samples directly from Tx\_sample[].
- **Tx\_data**[] is a circular buffer from which the Tx Data Pump gets the transmit data symbols. The Tx Data Handler writes the data symbols to be transmitted into this buffer.
- Tx\_uart\_data[] is a circular buffer from which the Tx Data Handler gets the transmit data chars. The Rx UART Interface writes data chars received from the DTE to be transmitted into this buffer. Also, this buffer is read by the Command Parser in order to detect AT commands entered by the user.

The Basic Blocks of the Soft Modem

- **Rx\_sample**[] is a circular buffer from which the Rx Data Pump reads samples. The Rx DAA Interface writes these samples directly into Rx\_sample[].
- **Rx\_data**[] is a circular buffer into which the Rx Data Pump places received data symbols. The Rx Data Handler reads demodulated or detected data symbols from this buffer.
- **Rx\_uart\_data**[] is a circular buffer into which the Rx Data Handler puts the received data chars. The Tx UART Interface reads data chars from this buffer and sends them to the DTE. Also this buffer is used to write data by Text Response functions, in order to send informative text messages to the DTE.
- **Command**[] is a circular buffer from which the Command Handler gets the detected commands and their parameters. The Command Parser writes detected sequences of user commands with parameters into this buffer.
- **State Machine** is a top-level entity that determines the behaviour of the Soft Modem, which implements the role of the scheduler (supervisor). And it has the capability to change the parameters of any Soft Modem block.
- **Data Handler** is a top-level entity that is invoked by the scheduler to manage transmission of actual data. Possible Data Handlers can be:
  - 1. Data handler according to the ITU-T V.14 recommendation.
  - 2. Data handler according to the ITU-T V.42 and V.42bis recommendation.
  - 3. Data handler that performs physical handshake, for example, according to V.22 or V.22bis.
  - 4. Data handler according to the ITU-T V.8 recommendation.
- The Data Handler performs the actual handling of bits of data.
- **Data Pump** is a top-level entity that is invoked by the scheduler to manage the conversion of digital data into signal samples that can be sent over a phone line and vice versa. Possible Data Pump arrangements can be:
  - 1. Data Pump according to the ITU-T V.21/V.22/V.22bis or V.23 recommendations.
  - 2. Data Pump that implements a DTMF generator.
  - 3. Data Pump that implements a Tone Generator and Detector.
  - 4. Data Pump that implements Ring detection.
- **Command Parser** is a top-level entity that is invoked by the scheduler to implement AT command syntax handling. This writes all sequences of valid system commands to the command[] circular buffer.
- **Command Handler** is a top-level entity that is invoked by the scheduler to interpret AT commands and execute them. It reads formatted commands from the command[] circular buffer, that are written into it by the Command Parser, in response to valid AT commands.

- **Text Response** is a set of functions that perform informative messaging to the DTE according to V.25ter.
- **DTE Interface** is a set of functions that perform communication between the DCE and the DTE. It uses the UART0 module.
- **DAA Interface** is a set of functions that implement control and interaction with the Modem Daughter card based on an integrated direct access arrangement (DAA) Si3044. It uses the UART1 module in modem mode (USART).

#### 2.2 Sampling Frequency

The sampling frequency used is 9600Hz. The 9600Hz sampling rate is practical for several reasons:

- It is higher than the Nyquist sampling frequency of approximately 8KHz for the telephone channel.
- It is a convenient multiple of the popular modem band rates (75, 300, 600, 1200 and 2400 band etc.)

#### 2.3 Numeric Formats

The Source code uses integer math functions only, no floating-point operation; this allows the algorithms to be optimized to the target processor for speed.

A lot of variables used in the modem are 16 bit signed fractional integers in Q14 and Q15 formats. In the case of Q14, this means that the msb is a sign bit, the bit before the msb is to the left of the decimal point and the remaining 14 bits are to the right of the decimal point. Therefore the numeric range is –32768 to +32767 corresponding to -2.000000 to 1.999939, with a quantization step of 1/32768. For example, Q14 format is used by the data pump modules to represent a PCM signal.

The header file "init.h" contains a set of useful macros for operation with fractional data:

```
. . .
```

#### 2.4 Channel

The Channel is the main control data structure, and is a mandatory parameter for most of the Soft Modem functions.

It is actually a union of the control structures that contains pointers to the Transmitter, Receiver, DTE and DAA control structures associated with this Channel.

In other words, the Channel is the root of the Soft Modem. Through the Channel data structure, it is possible to get any current parameter of the Soft Modem.

```
struct daa_control_t * daa_control;
};
```

The channel\_t structure parameter descriptions:

- **Tx\_control\_ptr** Pointer to the Transmitter control data structure (*see chapter 2.4.1. Transmitter*). Initialized in *Tx\_channel\_init(*).
- **Rx\_control\_ptr** Pointer to the Receiver control data structure (*see chapter 2.4.2. Receiver*). Initialized in *Rx\_channel\_init(*).
- **state\_machine\_ptr** Pointer to the State Machine control data structure (*see chapter 2.5. Global State Machine*). Initialized in *state\_machine\_init(*).
- **uart\_control** Pointer to the UART module control data structure (*see chapter 3.6.1. UART module*). Initialized in *mcf5407\_uart\_init()*.
- **daa\_control** Pointer to the DAA Interface control data structure (*see chapter 3.6.2. DAA Interface*). ). Initialized in *daa\_init()*.

#### 2.4.1 Transmitter

The Transmitter control structure is a constituent part of the Channel. It contains the parameters required for the Tx Data Handler, the Tx Data Pump, the Command Parser, the Command Handler and the transmitter circular data buffer operation.

```
uint32 data length;
   uint8 *uart data ptr;
   uint8 *uart data head;
   uint8 *uart data tail;
   uint8 *uart data end;
   uint32 uart data length;
   uint32 *command ptr;
   uint32 *command head;
   uint32 *command tail;
   uint32 *command end;
   uint32 command length;
   void * data pump ptr;
   call func t data pump call func;
   uint32 number samples;
  uint8 n bits;
   uint8 n bits mask;
   uint32 baud rate;
   uint32 process count;
   void * data handler ptr;
   call func t data handler call func;
   enum Tx DH state t data handler state;
  uint32 number n bits;
   void * command parser ptr;
   call func t command parser call func;
   call func t command handler call func;
};
```

The Tx\_control\_t structure parameter descriptions:

- **state** State identifier. It contains the current state of the transmitter. It is initialized in *Tx\_channel\_init()* to TX\_SILENCE\_STATE.
- **sample\_ptr** Pointer to the  $Tx_sample[]$  circular buffer. Users should not modify this pointer. It is initialized in  $Tx_channel_init()$  to the start of  $Tx_sample[]$ .

- **sample\_head** The Tx\_sample[] circular buffer write pointer. The Data Pump modules modify this pointer when writing their samples into Tx\_sample[]. It is initialized in *Tx\_channel\_init()* to the start of Tx\_sample[].
- **sample\_tail** The Tx\_sample[] circular buffer read pointer. The Tx DAA Interface modifies this pointer when reading samples from Tx\_sample[]. The Data Pump modules don't modify this pointer. It is initialized in *Tx\_channel\_init()* to the start of Tx\_sample[].
- **sample\_end** Pointer to the end of the Tx\_sample[] circular buffer. Users should not modify this pointer. It is initialized in *Tx\_channel\_init()* to the end of Tx\_sample[].
- **sample\_length** Length of the Tx\_sample[] circular buffer. Initialized in *Tx\_channel\_init()*. The user must ensure that the <code>sample\_length</code> is large enough to hold all of the samples generated per call as specified by the value of <code>number\_samples</code>.
- **data\_ptr** Pointer to the Tx\_data[] circular buffer. Users should not modify this pointer. It is initialized in *Tx\_channel\_init()* to the start of Tx\_data[].
- **data\_head** The Tx\_data[] circular buffer write pointer. The Data Handler modules modify this pointer when writing symbols to Tx\_data[]. It is initialized in *Tx\_channel\_init()* to the start of Tx\_data[].
- **data\_tail** The Tx\_data[] circular buffer read pointer. The Data Pump modules modify this pointer when reading symbols from Tx\_data[] for modulation. It is initialized in *Tx\_channel\_init()* to the start of Tx\_data[].
- **data\_end** Pointer to the end of the Tx\_data[] circular buffer. Users should not modify this pointer. It is initialized in *Tx\_channel\_init()* to the end of Tx\_data[].
- **data\_length** Length of the Tx\_data[] circular buffer. Initialized in *Tx\_channel\_init()*. Users must ensure that the <code>data\_length</code> is large enough to hold all of the symbols that will be modulated by the current Data Pump.
- **uart\_data\_ptr** Pointer to the Tx\_uart\_data[] circular buffer. Users should not modify this pointer. It is initialized in *Tx\_channel\_init()* to the start of Tx\_uart\_data[].
- **uart\_data\_head** The Tx\_uart\_data[] circular buffer write pointer. The UART interface modifies this pointer when writing characters into Tx\_uart\_data[]. It is initialized in *Tx\_channel\_init()* to the start of Tx\_uart\_data[].
- **uart\_data\_tail** The Tx\_uart\_data[] circular buffer read pointer. The Data Handler or Command Parser modifies this pointer when reading characters from Tx\_uart\_data[] for further processing. It is initialized in *Tx\_channel\_init()* to the start of Tx\_uart\_data[].
- **uart\_data\_end** Pointer to the end of the Tx\_uart\_data[] circular buffer. Users should not modify this pointer. It is initialized in *Tx\_channel\_init()* to the end of Tx\_uart\_data[].

- **uart\_data\_length** Length of the Tx\_uart\_data[] circular buffer. Initialized in *Tx\_channel\_init()*.
- **command\_ptr** Pointer to the command[] circular buffer. Users should not modify this pointer. It is initialized in *Tx\_channel\_init()* to the start of command[].
- **command\_head** The command[] circular buffer write pointer. The Command Parser modifies this pointer when writing detected sequences of user commands with parameters to the command[]. It is initialized in *Tx\_channel\_init()* to the start of command[].
- **command\_tail** The command[] circular buffer read pointer. The Command Handler module modifies this pointer when reading commands and parameters from command[] for further execution. It is initialized in *Tx\_channel\_init()* to the start of command[].
- **command\_end** Pointer to the end of the command[] circular buffer. Users should not modify this pointer. It is initialized in *Tx\_channel\_init()* to the end of command[].
- **command\_length** Length of the command[] circular buffer. Initialized in *Tx\_channel\_init()*.
- **data\_pump\_ptr** Pointer to the current Tx Data Pump control structure. Every Tx Data Pump module can have its own control data structure. It is initialized in *Tx\_channel\_init()* to NULL.
- **data\_pump\_call\_func** Pointer to the function call associated with the current Tx Data Pump. The  $Tx\_data\_pump()$  function simply calls whatever function this pointer is pointing to. It is initialized in  $Tx\_channel\_init()$  to  $Tx\_silence\_gen()$  (Generates silence).
- **number\_samples** Specifies the number of samples generated by Tx\_sample[] per call to the *Tx\_data\_pump()*. It is initialized in *Tx\_channel\_init()* to *DEFAULT\_NUMBER\_SAMPLES* (defined in "modem.h").
- **n\_bits** Number of bits per symbol for the current modulator. It is initialized in *Tx\_channel\_init()* to 1.
- **n\_bits\_mask** Bit mask corresponding to (1<< n\_bits) -1. Can be used to mask for valid bits when converting to symbols for Tx\_data[]. It is initialized in *Tx\_channel\_init()* to 1.
- **baud\_rate** Modulator baud rate. It is initialized in *Tx\_channel\_init()* to 0.
- **process\_count** Specifies the maximum number of modulator specific operations for completion of the current state. For example, the DTMF generator uses it as a number of modulated digits. It is initialized in *Tx\_channel\_init()* to -1.
- **data\_handler\_ptr** Pointer to the current Tx Data Handler control structure. Every Tx Data Handler module can have its own control structure. It is initialized in *Tx\_channel\_init()* to NULL.

- **data\_handler\_call\_func** Pointer to the function call associated with the current Tx Data Handler. The *Tx\_data\_handler()* function simply calls whatever function this pointer is pointing to. It is initialized in *Tx\_channel\_init()* to NULL.
- **data\_handler\_state** Current Tx Data Handler state identifier. It is initialized in *Tx\_channel\_init()* to TX\_DH\_STATE\_NORMAL.
- **number\_n\_bits** Number of elements that should be generated in Tx\_data[] per call to  $Tx_data_handler()$ . It is initialized in  $Tx_channel_init()$  to DEFAULT\_NUMBER\_NBITS (defined in "modem.h").
- **command\_parser\_ptr** Pointer to the current Command Parser control data structure. It is initialized in *Tx\_channel\_init()* to NULL.
- **command\_parser\_call\_func** Pointer to the function call associated with the current Command Parser. The *command\_parser()* function simply calls whatever function this pointer is pointing to. It is initialized in Tx\_channel\_init() to NULL.
- **command\_handler\_call\_func** Pointer to the function call associated with the current Command Handler. The *command\_handler()* function simply calls whatever function this pointer is pointing to. It is initialized in Tx\_channel\_init() to NULL.

#### 2.4.2 Receiver

The Receiver control structure is a constituent part of the Channel. It contains the parameters required for the Rx Data Handler, the Rx Data Pump and the receiver circular data buffer operation.

#### Channel

```
uint8 *data tail;
   uint8 *data end;
   uint32 data length;
   uint8 *uart data ptr;
   uint8 *uart data head;
   uint8 *uart data tail;
   uint8 *uart data end;
   uint32 uart data length;
   void * data pump ptr;
   call func t data pump call func;
   uint32 number samples;
  uint8 n bits;
  uint8 n bits mask;
  uint32 baud rate;
  uint32 process count;
   void * data handler ptr;
   call func t data handler call func;
   enum Rx DH state t data handler state;
   uint32 number n bits;
   uint32 connection code;
};
```

The Tx\_control\_t structure parameter descriptions:

- **state** State identifier. It contains the current state of the receiver. It is initialized in *Rx\_channel\_init()* to RX\_IDLE\_STATE.
- **sample\_ptr** Pointer to the Rx\_sample[] circular buffer. Users should not modify this pointer. It is initialized in *Rx\_channel\_init()* to the start of Rx\_sample[].
- **sample\_head** The Rx\_sample[] circular buffer write pointer. The Rx DAA Interface modifies this pointer when writing the samples into Rx\_sample[]. It is initialized in *Rx\_channel\_init()* to the start of Rx\_sample[].
- **sample\_tail** The Rx\_sample[] circular buffer read pointer. The Rx Data Pump modules modify this pointer when reading samples from Rx\_sample[]. It is initialized in *Rx\_channel\_init()* to the start of Rx\_sample[].

- **sample\_end** Pointer to the end of the Rx\_sample[] circular buffer. Users should not modify this pointer. It is initialized in *Rx\_channel\_init()* to the end of Rx\_sample[].
- **sample\_length** Length of the Rx\_sample[] circular buffer. Initialized in *Rx\_channel\_init()*.
- **data\_ptr** Pointer to the Rx\_data[] circular buffer. Users should not modify this pointer. It is initialized in Rx\_channel\_init() to the start of Rx\_data[].
- **data\_head** The Rx\_data[] circular buffer write pointer. The Data Pump modules modify this pointer when writing symbols into Rx\_data[]. It is initialized in *Rx\_channel\_init()* to the start of Rx\_data[].
- **data\_tail** The Rx\_data[] circular buffer read pointer. The Data Handler modules modify this pointer when reading symbols from Rx\_data[] for further processing. It is initialized in *Rx\_channel\_init()* to the start of Rx\_data[].
- **data\_end** Pointer to the end of the Rx\_data[] circular buffer. Users should not modify this pointer. It is initialized in *Rx\_channel\_init()* to the end of Rx\_data[].
- **data\_length** Length of the Rx\_data[] circular buffer. Initialized in *Rx\_channel\_init()*. Users must ensure that the <code>data\_length</code> is large enough to hold all of the symbols that will be demodulated by the current Data Pump.
- **uart\_data\_ptr** Pointer to the Rx\_uart\_data[] circular buffer. Users should not modify this pointer. It is initialized in *Rx\_channel\_init()* to the start of Rx\_uart\_data[].
- **uart\_data\_head** The Rx\_uart\_data[] circular buffer write pointer. The Data Handler modules modify this pointer when writing characters into Rx\_uart\_data[]. It is initialized in *Rx\_channel\_init()* to the start of Rx\_uart\_data[].
- **uart\_data\_tail** The Rx\_uart\_data[] circular buffer read pointer. The UART Interface modifies this pointer when reading characters from Rx\_uart\_data[] for sending to the DTE. It is initialized in *Rx\_channel\_init()* to the start of Rx\_uart\_data[].
- **uart\_data\_end** Pointer to the end of the Rx\_uart\_data[] circular buffer. Users should not modify this pointer. It is initialized in *Rx\_channel\_init()* to the end of Rx uart data[].
- **uart\_data\_length** Length of the Rx\_uart\_data[] circular buffer. Initialized in *Rx\_channel\_init()*.
- data\_pump\_ptr Pointer to the current Rx Data Pump control structure. Every Data Pump module can have its own control data structure. Initialized in Rx\_channel\_init() to NULL.
- **data\_pump\_call\_func** Pointer to the function call associated with the current Rx Data Pump. The *Rx\_data\_pump()* function simply calls whatever function this

- pointer is pointing to. It is initialized in  $Rx\_channel\_init()$  to  $Rx\_idle()$  (Discards samples from the  $Rx\_sample[]$ ).
- **number\_samples** Specifies the number of samples that should be in Rx\_sample[] before an *Rx\_data\_pump()* call. It is initialized in Rx\_channel\_init() to *DEFAULT\_NUMBER\_SAMPLES* (defined in "modem.h").
- **n\_bits** Number of bits per symbol for the current demodulator. Initialized in *Rx\_channel\_init()* to 1.
- **n\_bits\_mask** Bit mask corresponding to (1<< n\_bits) -1. Can be used to mask for valid bits when converting to symbols from Rx\_data[]. Initialized in Rx\_channel\_init() to 1.
- **baud\_rate** Demodulator baud rate. It is initialized in *Rx\_channel\_init()* to 0.
- **process\_count** Specifies the maximum number of demodulator specific operations for completion of the current state. For example, Ring detector uses it as number of Ring cadence to be detected. It is initialized in *Rx\_channel\_init()* to -1.
- data\_handler\_ptr Pointer to current Rx Data Handler control structure. Every Rx Data Handler module can have its own control structure. It is initialized in Rx\_channel\_init() to NULL.
- **data\_handler\_call\_func** Pointer to the function call associated with the current Rx Data Handler. The *Rx\_data\_handler()* function simply calls whatever function this pointer is pointing to. It is initialized in *Rx\_channel\_init()* to NULL.
- **data\_handler\_state** Current Rx Data Handler state identifier. It is initialized in *Rx\_channel\_init()* to RX\_DH\_STATE\_NORMAL.
- **number\_n\_bits** Specifies the number of elements that should be in Rx\_data[] before a *data\_handler\_call\_func()* call. It is initialized in *Rx\_channel\_init()* to *DEFAULT\_NUMBER\_NBITS* (defined in "modem.h").
- **connection\_code** Contains two connection responses used for printing to the DTE after the Handshake is complete. The Data Handler module modifies this value after protocol handshake completion. It is initialized in *Rx\_channel\_init()* to 0.

#### 2.5 Global State Machine

The Global State Machine determines the behavior of the Soft Modem. It implements the role of the scheduler (supervisor). The State Machine is responsible for proper initialization and calling of the Data Pump, the Data Handler, the Command Parser and Command Handler and other blocks according to the Soft Modem settings.

The description of the Global State Machine states can be found in Table 2.5. The graphical representation of it is represented in Fig 2.5.

The state\_machine\_t structure is defined in "state\_machine.h" :

### Freescale Semiconductor, Inc.

```
/************************
* States of the Global State Machine
*************************
enum sm state t {SM INITIALIZATION STATE,
               SM RESET COMMAND STATE,
               SM COMMAND STATE,
               SM INIT ON LINE COMMAND STATE,
               SM ON LINE COMMAND STATE,
               SM INIT ON LINE STATE,
               SM ON LINE STATE,
               SM INIT PRE DIAL STATE,
               SM PRE DIAL PAUSE STATE,
               SM DIALTONE DET STATE,
               SM INIT DIALING STATE,
               SM DIALING STATE,
               SM PRE PHISICAL HANDSHAKING STATE,
               SM PHISICAL HANDSHAKING STATE,
               SM PROTOCOL HANDSHAKING STATE,
               SM INIT RETRAIN STATE,
               SM RETRAIN STATE,
               SM INIT PRE ANSWERTONE PAUSE STATE,
               SM PRE ANSWERTONE PAUSE STATE,
               SM INIT AFTER ANSWERTONE PAUSE STATE,
               SM AFTER ANSWERTONE_PAUSE_STATE,
               SM INIT ANSWERTONE GEN STATE,
               SM ANSWERTONE GEN STATE,
               SM INIT ANSWERTONE DET STATE,
               SM ANSWERTONE DET STATE,
               SM INIT V8 HANDSHAKING STATE,
               SM V8 HANDSHAKING STATE,
               SM INIT_LOOPBACK_STATE);
```

### Freescale Semiconductor, Inc.

#### **Global State Machine**

The state machine t structure parameter descriptions:

• **state** - Contains the current state of the Global State Machine. It is initialized in *state\_machine\_init()* to *SM\_INITIALIZATION\_STATE*.

**Table 2-1. The Global State Machine States** 

Name of the State	Activities	Exit States
SM_INITIALIZATION_STATE	Initialises the Command Parser and the Command Handler.	SM_RESET_COMMAND_STATE
SM_RESET_COMMAND_STATE	Resets the Receiver and the Transmitter. Initialises the Ring Detector.	SM_COMMAND_STATE
SM_COMMAND_STATE	Calls the Command Parser, the Command Handler and the Rx Data Pump (Ring Detector). If a Ring is detected, it sends a "RING" response to the DTE.	1)SM_COMMAND_STATE:  • By default. 2)SM_INIT_PRE_ANSWERTONE_PAUS E_STATE:  • If S0 is less or equal to S1.  • On an AT A command. 3)SM_INIT_PRE_DIAL_STATE:  • On an AT D command. 4)SM_RESET_COMMAND_STATE:  • On an AT H command. 5)SM_INIT_LOOPBACK_STATE:  • On an AT &T command. 6)SM_INITIALIZATION_STATE:  • On an AT Z command.
SM_INIT_ON_LINE_COMMAND_STATE	Sends an "OK" response to the DTE. Sets the mode of the Rx & Tx Data Handler and the Command Parser to the "Online command mode".	SM_ON_LINE_COMMAND_STATE
SM_ON_LINE_COMMAND_STATE	Calls the Tx & Rx Data Pumps. Calls the Tx & Rx Data Handlers. Calls the Command Parser and the Command Handler. On the "break of line" event, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1)SM_ON_LINE_COMMAND_STATE:  • By default. 2) SM_ON_LINE_STATE:  • On an AT O command. 3)SM_INITIALIZATION_STATE:  • On an AT Z command.

**Table 2-1. The Global State Machine States (continued)** 

Name of the State	Activities	Exit States
SM_INIT_ON_LINE_STATE	Initialises the Rx & Tx Data Pumps, and the Rx & Tx Data Handlers (Handshaker) according to the chosen protocol and speed (V.21, V.23, V.22, V.22bis)	1)SM_PRE_PHISICAL_HANDSHAKING_ STATE: • If the Channel is in the calling mode. 2)SM_PHISICAL_HANDSHAKING_STAT E: - If the Channel is in the angularing mode.
SM_ON_LINE_STATE	Calls the Tx & Rx Data Pumps. Calls the Tx & Rx Data Handlers. Calls the Command Parser for the Escape Sequence detection. On the "break of line" event, it goes on-hook.	If the Channel is in the answering mode.  1) SM_ON_LINE_STATE: By default. 2) SM_INIT_RETRAIN_STATE: On Retrain sequence detection. 3) SM_RESET_COMMAND_STATE: On the Break of the line. 4) SM_INIT_ON_LINE_COMMAND_STATE: If the Escape sequence is detected and the "Go to On-line command mode on escape code" is set in the S18 register.
SM_INIT_PRE_DIAL_STATE	If AT X2 or X4 is set, it initialises the Tone Detector as the Dial tone detector, otherwise (No Dial tone detection or Blind Dialing) it initialises the Tone Generator as the Silence tone generator.  And it goes off-hook.	1)SM_PRE_DIAL_PAUSE_STATE:  If AT X0 was set before.  If AT X1 was set before.  If AT X3 was set before.  2)SM_DIALTONE_DET_STATE:  If AT X2 was set before.  If AT X4 was set before.
SM_PRE_DIAL_PAUSE_STATE	Calls the Rx Data Pump (Idle mode) and the Tx Data Pump (Silence Generator). On receipt of any character from the DTE, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1) SM_PRE_DIAL_PAUSE_STATE:  • By default.  2) SM_INIT_DIALING_STATE:  • On the completion of the Silence Generation.  3) SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.
SM_DIALTONE_DET_STATE	Calls the Tx Data Pump (Silence generator) and the Rx Data Pump (Dial Tone Detector).  On the receipt of any character from the DTE, it goes on-hook and sends the "NO CARRIER" response to the DTE.  If the Dial Tone was not detected during a fixed period of time, it goes on-hook and sends the "NO DIAL TONE" response to the DTE.	1) SM_DIALTONE_DET_STATE:  • By default. 2) SM_INIT_DIALING_STATE:  • If the Dial Tone is detected during a fixed time. 3) SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.  • If the Dial Tone is not detected during a fixed period of time.
SM_INIT_DIALING_STATE	Fills the Tx_data[] buffer with dial digits. initializes the DTMF dialer or the Pulse dialer (if Pulse dialing is enabled in the S[13] register)	SM_DIALING_STATE

**Table 2-1. The Global State Machine States (continued)** 

Name of the State	Activities	Exit States
SM_DIALING_STATE	Calls the Tx Data Pump (The DTMF or Pulse dialer) and the Rx Data Pump (Idle mode). On the receipt of any character from the DTE, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1) SM_DIALING_STATE:  • By default. 2) SM_INIT_ANSWERTONE_DET_STATE:  • If dialing is completed and "return to command mode after dialing" (H[2] register) is disabled. 3) SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.  • If dialing is completed and "return to command mode after dialing" (H[2] register) is enabled.
SM_PRE_PHISICAL_HANDSHAKING_STAT E	Calls the Rx Data Handler (Handshaker) and the Rx Data Pump. On the receipt of any character from the DTE and on the "break of the line" event, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1)SM_PRE_PHISICAL_HANDSHAKING_STATE:  • By default. 2)SM_PHISICAL_HANDSHAKING_STATE:  • If the Rx Data Handler (Handshaker) current state is complete. 3)SM_RESET_COMMAND_STATE: • On the reception of any character from the DTE. • On the Break of line.
SM_PHISICAL_HANDSHAKING_STATE	Calls the Tx & Rx Data Handlers (Handshaker) and the Tx & Rx Data Pumps. When the Rx Data Handler (Handshaker) current state is complete, it initializes the V.14 or V.42 Data Handler, and mutes the speaker on the modem daughter card. On receiving any character from the DTE and on the "break of the line" event, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1)SM_PHISICAL_HANDSHAKING_STAT E:  • By default. 2)SM_PROTOCOL_HANDSHAKING_ST ATE:  • When the Rx Data Handler (Handshaker) current state is complete. 3)SM_INIT_RETRAIN_STATE:  • On the Retrain sequence detection. 4)SM_RESET_COMMAND_STATE: • On the reception of any character from the DTE. • On the Break of the line.
SM_PROTOCOL_HANDSHAKING_STATE	Calls the Tx & Rx Data Handlers (Handshaker) and the Tx & Rx Data Pumps. When the Rx Data Handler (Handshaker) current state is complete, it sends the "CONNECT XXX" response to the DTE. On the receipt of any character from the DTE or on the "break of the line" event, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1)SM_PROTOCOL_HANDSHAKING_ST ATE:  • By default. 2) SM_ON_LINE_STATE:  • When the Rx Data Handler (Handshaker) current state is complete. 3)SM_INIT_RETRAIN_STATE:  • On the Retrain sequence detection. 4)SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.  • On the Break of the line.

**Table 2-1. The Global State Machine States (continued)** 

Name of the State	Activities	Exit States
SM_INIT_RETRAIN_STATE	If the current Data Pump is in the 2400bps V.22bis mode, it initializes V.22bis Handshake in the Retrain mode.	1)SM_RETRAIN_STATE:  2 By default. 2)SM_PHISICAL_HANDSHAKING_STAT E: 2 If the variable return_sm_state is equal to SM_PHISICAL_HANDSHAKING_STATE. 3) The state is equal to the value of the return_sm_state variable: 2 If the current Data Pump is not in the 2400bps V.22bis mode.
SM_RETRAIN_STATE	Calls the Tx & Rx Data Handlers (Handshaker) and the Tx & Rx Data Pumps.  When the Rx Data Handler (Handshaker) current state is complete, it loads the saved parameters of the old Data Handler.  If the variable return_sm_state is equal to SM_ON_LINE_STATE, it calls the Command Parser for Escape Sequence detection, otherwise on the receipt of any character from the DTE it goes on-hook and sends the "NO CARRIER" response to the DTE.  On the "break of the line" event, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1) SM_RETRAIN_STATE:  • By default.  2) The state equals to the value of the return_sm_state variable:  • When the Rx Data Handler (Handshaker) current state is complete.  3) SM_INIT_RETRAIN_STATE:  • On the Retrain sequence detection.  4) SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE and when the return_sm_state variable is not equal to SM_ON_LINE_STATE.  • On the Break of the line.  4) SM_INIT_ON_LINE_COMMAND_STATE:  • On Escape sequence is detected, "Go to On-line command mode on escape code" is set in the S18 register, and the return_sm_state variable is equal to SM_ON_LINE_STAT.
SM_INIT_PRE_ANSWERTONE_PAUSE_STA TE	It initializes the Tone Generator as the Silence tone generator. And it goes off-hook.	SM_PRE_ANSWERTONE_PAUSE_STAT E
SM_PRE_ANSWERTONE_PAUSE_STATE	Calls the Rx Data Pump (Idle mode) and the Tx Data Pump (Silence Generator). On the receipt of any character from the DTE, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1)SM_PRE_ANSWERTONE_PAUSE_ST ATE:  • By default. 2)SM_INIT_ANSWERTONE_GEN_STAT E:  • If Silence Generation is completed. 3)SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.
SM_INIT_AFTER_ANSWERTONE_PAUSE_S TATE	It initializes the Tone Generator as the Silence tone generator.	SM_AFTER_ANSWERTONE_PAUSE_S TATE

**Table 2-1. The Global State Machine States (continued)** 

Name of the State	Activities	Exit States
SM_AFTER_ANSWERTONE_PAUSE_STATE	Calls the Rx Data Pump (Idle mode) and the Tx Data Pump (Silence Generator). On the receipt of any character from the DTE, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1)SM_AFTER_ANSWERTONE_PAUSE_STATE:  • By default. 2)SM_INIT_ON_LINE_STATE:  • If the Silence Generation is completed. 3)SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.
SM_INIT_ANSWERTONE_GEN_STATE	If V.8 is disabled (S[19] register), it initializes the Tone Generator as the Answer (ANS) tone generator, otherwise as the ANSam tone generator.	1)SM_ANSWERTONE_GEN_STATE:  • If V.8 is disabled.  2)SM_INIT_V8_HANDSHAKING_STATE:  • If V.8 is enabled.
SM_ANSWERTONE_GEN_STATE	Calls the Tx Data Pump (ANS or ANSam generator).  If V.8 is enabled it calls the Rx Data Handler (V.8) and the Rx Data Pump (V.21), otherwise it calls only the Rx Data Pump (Idle mode).  If V.8 is enabled, on the end of an Answer tone generation, it initializes the V.21 Tx Data Pump.  On the receipt of any character from the DTE, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1) SM_ANSWERTONE_GEN_STATE:  • By default. 2) SM_V8_HANDSHAKING_STATE:  • At the end of Answer Tone generation if V.8 is enabled. 3) SM_INIT_AFTER_ANSWERTONE_PAUSE_STATE:  • At the end of Answer Tone generation, if V.8 is disabled 4) SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.
SM_INIT_ANSWERTONE_DET_STATE	If AT X3 or X4 is set, it adds the Busy tone detection to the Tone Detector. It also adds the End of Answer (ANS) Tone detection to the Tone Detector. If V8 protocol is enabled (S[19] register), it adds the ANSam tone detection to the Tone Detector.	SM_ANSWERTONE_DET_STATE
SM_ANSWERTONE_DET_STATE	Calls the Tx Data Pump (Silence generator) and the Rx Data Pump (Tone Detector).  On the receipt of any character from the DTE, it goes on-hook and sends the "NO CARRIER" response to the DTE.  If the Busy Tone was detected, it goes on-hook and sends the "BUSY" response to the DTE.  If no tones were detected during a fixed period of time, it goes on-hook and sends the "NO ANSWER" response to the DTE.	1) SM_ANSWERTONE_DET_STATE:  • By default. 2) SM_INIT_ON_LINE_STATE:  • When the End of Answer (ANS) Tone is detected. 3) SM_INIT_V8_HANDSHAKING_STATE:  • When the ANSam Tone is detected. 3) SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.  • When no tones were detected during a fixed period of time.  • When the Busy tone is detected.
SM_INIT_V8_HANDSHAKING_STATE	Initialises the V.8 Rx and Tx Data Handlers. Initialises the V.21 Rx Data Pump. If the Channel is in calling mode (S[13] register), it initializes the V.21 Tx Data Pump.	1) SM_V8_HANDSHAKING_STATE:  If the Channel is in calling mode.  2) SM_ANSWERTONE_GEN_STATE:  If the Channel is in answering mode.

**Table 2-1. The Global State Machine States (continued)** 

Name of the State	Activities	Exit States
SM_V8_HANDSHAKING_STATE	Calls the Tx & Rx Data Handlers (V.8 Handshaker) and the Tx & Rx Data Pumps. On the receipt of any character from the DTE and on the "break of the line" event, it goes on-hook and sends the "NO CARRIER" response to the DTE.	1) SM_V8_HANDSHAKING_STATE:  • By default.  2) SM_INIT_AFTER_ANSWERTONE_PAUSE_STATE:  • When the Rx Data Handler (Handshaker) current state is complete.  3) SM_RESET_COMMAND_STATE:  • On the reception of any character from the DTE.  • On the Break of the line.
SM_INIT_LOOPBACK_STATE	Sets UART1 to local loop back mode. Initialises the Rx & Tx Data Pumps, and the Rx & Tx Data Handlers (Handshaker) according to the chosen protocol and speed ( V.21, V.23, V.22, V.22bis).	SM_PHISICAL_HANDSHAKING_STATE

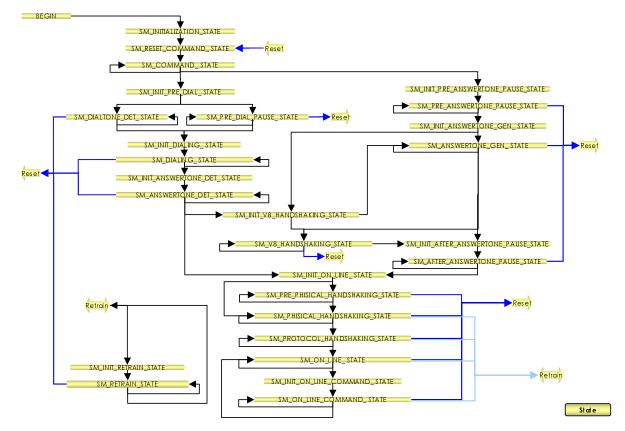


Figure 2-3. Global State Machine

# state\_machine\_init

### Call(s):

#### **Arguments:**

### Table 2-2. state\_machine\_init arguments

channel	in	Pointer to the Channel control data structure
sm	in	Pointer to the State Machine control data structure

### **Description:**

This function initializes the State Machine control data structure. It must be called before calling the state\_machine() function.

This function is called once in *main()*.

Returns: None.

```
struct channel_t channel;
struct state_machine_t sm;
...
state_machine_init(&channel, &sm);
...
```

# state\_machine

### Call(s):

```
void state_machine(struct channel_t * channel);
```

### **Arguments:**

### Table 2-3. state\_machine arguments

channel in Pointer to the Channel control data structure	el in	Pointer to the Channel control data structure
--	-------	---

### **Description:**

This function implements the Global State Machine of the Soft Modem. The description of each state is in Table 2.5.

This function is called in *main()*.

Returns: None.

```
struct channel_t channel;
struct state_machine_t sm;
...
state_machine_init(&channel, &sm);
state_machine(&channel);
...
```

### 2.6 Interface

This section describes the common functions and some data structures used in the Soft Modem.

### 2.6.1 Global Data Structures

This section describes the global data structures that are used in the Soft Modem. They are declared in the "modem. c" file.

Table 2-4. Global Data Structures

struct channel_t* CURRENT_CHANNEL	Pointer to the Current Channel control structure used by the interrupt routines. It is initialized by the global_structure_init() function.
const uint8 SUPPORT_COUNTRIES[NUMBER_SUPPORT_COUNTRIES]	Array of countries that are supported by the current version of the Soft Modem.
uint8 S[S_NUMBER]	Array of "S Registers". It is initialized by the global_structure_init() function. These registers are accessed via AT commands for setting the modem configuration parameters and establishing modem default behavior.  The S registers are used to provide adjustability while avoiding unnecessary coding changes.
const uint8 S_FACTORY_DEFAULT [S_NUMBER]	Factory-defined Configuration of the S-registers, contains the default characters and values used by the Soft Modem modules, these are "firmware", and can be modified at compile time.
uint8 H[H_NUMBER]	Array of "Hidden Registers". It is initialized by the global_structure_init() function. This array of registers is not accessed via AT commands directly and is internally used by the Soft Modem modules.
const uint8 H_FACTORY_DEFAULT [H_NUMBER]	Factory-defined Configuration of H-registers, contains the default values used by the Soft Modem modules, these are "firmware", and can be modified at compile time.

### 2.6.2 Circular Buffer Inline Functions

These functions introduce convenient access to the elements of the Soft Modem circular buffers: Tx\_sample[], Rx\_sample[], Tx\_data[], Rx\_data[], Tx\_uart\_data[], Rx\_uart\_data[] and command[].

# Tx\_sample\_write

Call(s):

### **Arguments:**

### Table 2-5. Tx\_sample\_write arguments

Tx_control	in	Pointer to the Transmitter control data structure
sample	in	Sample for writing to the Tx_sample[] circular buffer

### **Description:**

This function writes the <code>sample</code> into the <code>Tx\_sample[]</code> circular buffer to the position pointed to by <code>Tx\_control->sample\_head</code>, then updates the <code>Tx\_control->sample\_head</code> to the next position in the buffer.

Returns: None.

```
struct Tx_control_t Tx_control;
int16 sample=0;
...
Tx_sample_write (&Tx_control, sample);
...
```

## Tx\_sample\_read

### Call(s):

```
inline int16 Tx_sample_read (struct Tx_control_t *Tx_control);
```

### **Arguments:**

#### Table 2-6. Tx\_sample\_read arguments

Tx_control	in	Pointer to the Transmitter control data structure
------------	----	---

**Description:** This function reads the value of the current element in the

Tx\_sample[] circular buffer pointed to by  $Tx\_control->sample\_tail$ , then updates  $Tx\_control->sample\_tail$  to the next position in the

buffer.

**Returns:** Value of the element pointed to by *Tx\_control->sample\_tail*.

```
...
struct Tx_control_t Tx_control;
int16 sample;
...
sample=Tx_sample_read (&Tx_control);
...
```

# Rx\_sample\_write

### Call(s):

### **Arguments:**

### Table 2-7. Rx\_sample\_write arguments

Rx_control	in	Pointer to the Receiver control data structure
sample	in	Sample for writing to the Rx_sample[] circular buffer

### **Description:**

This function writes the <code>sample</code> into the Rx\_sample[] circular buffer to the position pointed to by <code>Rx\_control->sample\_head</code>, then updates <code>Rx\_control->sample\_head</code> to the next position in the buffer.

**Returns:** None.

```
struct Rx_control_t Rx_control;
int16 sample=0;
...
Rx_sample_write (&Rx_control, sample);
...
```

## Rx\_sample\_read

### Call(s):

inline int16 Rx\_sample\_read (struct Rx\_control\_t \*Rx\_control);

#### **Arguments:**

### Table 2-8. Rx\_sample\_read arguments

Rx_control	in	Pointer to the Receiver control data structure

**Description**: This function reads the value of the current element in the

Rx\_sample[] circular buffer pointed to by

Rx control->sample tail, then updates Rx control->sample tail

to the next position in the buffer.

**Returns**: Value of the element pointed to by  $Rx\_control->sample\_tail$ .

```
...
struct Rx_control_t Rx_control;
int16 sample;
...
sample=Rx_sample_read (&Rx_control);
...
```

# Tx\_data\_write

### Call(s):

### **Arguments:**

### Table 2-9. Tx\_data\_write arguments

Tx_control	in	Pointer to the Transmitter control data structure
data	in	Symbol for writing to the Tx_data[] circular buffer

**Description**:

This function writes <code>data</code> into the <code>Tx\_data[]</code> circular buffer to the position pointed to by <code>Tx\_control->data\_head</code>, then updates-<code>Tx\_control->data\_head</code> to the next position in the buffer.

**Returns:** None.

```
struct Tx_control_t Tx_control;
uint8 data=0;
...
Tx_data_write (&Tx_control, data);
...
```

# Tx\_data\_read

### Call(s):

inline uint8 Tx\_data\_read (struct Tx\_control\_t \*Tx\_control);

#### **Arguments:**

### Table 2-10. Tx\_data\_read arguments

	Tx_control	in	Pointer to the Transmitter control data structure
--	------------	----	---

**Description:** This function reads the value of the current element in the Tx\_data[]

circular buffer pointed to by Tx control->data tail, then updates

Tx\_control->data\_tail to the next position in the buffer.

**Returns:** Value of the element pointed to by Tx\_control->data\_tail

```
struct Tx_control_t Tx_control;
uint8 data;
...
data=Tx_data_read (&Tx_control);
...
```

# Rx\_data\_write

Call(s):

### **Arguments:**

### Table 2-11. Rx\_data\_write arguments

Rx_control	in	Pointer to the Receiver control data structure
data	in	Symbol for writing to the Rx_data[] circular buffer

**Description:** This function writes the data into the Rx\_data[] circular buffer to the

position pointed to by Rx\_control->data\_head, then updates Rx control->data head to the next position in the buffer.

**Returns:** None.

```
...
struct Rx_control_t Rx_control;
uint8 data=0;
...
Rx_data_write (&Rx_control, data);
...
```

### Rx\_data\_read

### Call(s):

```
inline uint8
Rx data read (struct Rx control t *Rx control);
```

#### **Arguments:**

### Table 2-12. Rx\_data\_read arguments

Rx_control	in	Pointer to the Receiver control data structure
------------	----	--

**Description:** This function reads the value of the current element in the Rx\_data[]

circular buffer pointed to by Rx\_control->data\_tail, then updates

Rx control->data tail to the next position in the buffer.

**Returns:** Value of the element pointed to by Rx\_control->data\_tail.

```
...
struct Rx_control_t Rx_control;
uint8 data;
...
data=Rx_data_read (&Rx_control);
...
```

# Tx\_uart\_data\_write

### Call(s):

### **Arguments:**

### Table 2-13. Tx\_uart\_data\_write arguments

Tx_control	in	Pointer to the Transmitter control data structure
uart_data	in	Character for writing to the Tx_uart_data[] circular buffer

**Description:** This function writes the uart\_data into the Tx\_uart\_data[] circular

buffer into the position pointed to by Tx\_control->uart\_data\_head, then updates Tx control->uart data head to the next position in

the buffer.

**Returns:** None.

```
struct Tx_control_t Tx_control;
uint8 character='a';
...
Tx_uart_data_write (&Tx_control, character);
...
```

# Tx\_uart\_data\_read

### Call(s):

inline uint8 Tx\_uart\_data\_read (struct Tx\_control\_t \*Tx\_control);

#### **Arguments:**

### Table 2-14. Tx\_uart\_data\_read arguments

Tx_contro	ol	in	Pointer to the Transmitter control data structure
-----------	----	----	---

**Description:** This function reads the value of the current element in the

Tx\_uart\_data[] circular buffer pointed to by Tx\_control->uart\_data\_tail, then updates

Tx control->uart data tail to the next position in the buffer.

**Returns:** Value of the element pointed to by Tx\_control->uart\_data\_tail.

```
struct Tx_control_t Tx_control;
uint8 character;
...
character=Tx_uart_data_read (&Tx_control);
...
```

# Rx\_uart\_data\_write

### Call(s):

### **Arguments:**

Table 1: Rx\_uart\_data\_write arguments

Rx_control	in	Pointer to the Receiver control data structure
uart_data	in	Symbol for writing to the Rx_uart_data[] circular buffer

**Description:** This function writes the uart\_data into the Rx\_data[] circular buffer

to the position pointed to by Rx\_control->uart\_data\_head, then updates Rx\_control->uart\_data\_head to the next position in the

buffer.

**Returns:** None.

```
...
struct Rx_control_t Rx_control;
uint8 character='a';
...
Rx_uart_data_write (&Rx_control, character);
...
```

## Rx\_uart\_data\_read

Call(s):

inline uint8 Rx\_uart\_data\_read (struct Rx\_control\_t \*Rx\_control);

### **Arguments:**

### Table 2-15. Rx\_uart\_data\_read arguments

Rx_control	in	Pointer to the Receiver control data structure
------------	----	--

**Description:** This function reads the value of the current element of the

Rx\_uart\_data[] circular buffer pointed to by Rx\_control->uart\_data\_tail, then updates

Rx\_control->uart\_data\_tail to the next position in the buffer.

**Returns:** Value of the element pointed to by Rx\_control->uart\_data\_tail

```
struct Rx_control_t Rx_control;
uint8 character;
...
character =Rx_uart_data_read (&Rx_control);
...
```

# command\_write

### Call(s):

#### **Arguments:**

### Table 2-16. command\_write arguments

Tx_control	in	Pointer to the Transmitter control data structure
command	in	Command number and its three parameters for writing to the command[] circular buffer.  4-th byte is the command number;  3-rd byte is the 1-st parameter;  2-nd byte is the 2-nd parameter;  1-st byte is the 3-rd parameter;

Description: This function writes the command into the command[] circular buffer to the position pointed to by

 $Tx\_control -> command\_head$ , then updates  $Tx\_control -> command\_head$  to the next

position in the buffer.

Returns: None.

```
#define COMMANDAT 1

...

struct Tx_control_t Tx_control;

uint32 command=COMMANDAT;

...

command_write (&Tx_control, command);

...
```

## command\_read

### Call(s):

inline uint32 command\_read (struct Tx\_control\_t \*Tx\_control);

#### **Arguments:**

### Table 2-17. command\_read arguments

Tx_control	in	Pointer to the Transmitter control data structure
------------	----	---

**Description:** This function

This function reads the value of the current element in the command[] circular buffer pointed to by  $Tx\_control->command\_tail$ , then updates  $Tx\_control->command\_tail$  to the next action in the huffer.

position in the buffer.

**Returns:** Value of the element pointed to by  $Tx\_control -> command\_tail$ .

```
struct Tx_control_t Tx_control;
uint32 command;
...
command=command_read (&Tx_control);
...
```

# global\_structure\_init

### Call(s):

```
void global structure init (struct channel t * channel);
```

#### **Arguments:**

### Table 2-18. global\_structure\_init arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This function initializes the S[] and H[] global data arrays (registers) to the default values, and assigns the CURRENT\_CHANNEL global pointer (pointer to the current Channel used by the Interrupt Service Routines) to the input argument *channel*.

This function is called once in main().

Returns: None.

```
struct channel_t channel;
...
global_structure_init(&channel);
...
```

## Tx\_channel\_init

### Call(s):

### **Arguments:**

#### Table 2-19. Tx\_channel\_init arguments

channel	in	Pointer to the Channel control data structure
Tx_control	in	Pointer to the Transmitter control data structure
data	in	Pointer to the Tx_data[] circular buffer
data_length	in	Length of the Tx_data[] circular buffer
sample	in	Pointer to the Tx_sample[] circular buffer
sample_length	in	Length of the Tx_sample[] circular buffer
uart_data	in	Pointer to the Tx_uart_data[] circular buffer
uart_data_length	in	Length of the Tx_uart_data[] circular buffer
command	in	Pointer to the command[] circular buffer
command_length	in	Length of the command[] circular buffer

**Description:** 

This function initializes all the fields of the Transmitter control data structure, to the default values. This function should be called before calling Tx\_data\_pump(), Tx\_data\_handler(), command\_parser(), command\_handler() and Transmitter Protocols initialization functions.

This function is called once in *main()*.

**Returns:** None.

#### Interface

```
#define DATA LENGTH
                          (60)
#define SAMPLE_LENGTH
                          (400)
#define COMMAND LENGTH
                          (40)
#define UART_DATA_LENGTH (300)
struct channel t channel;
struct Tx control t Tx control;
uint8 Tx data[DATA LENGTH];
int16 Tx sample[SAMPLE LENGTH];
uint8 Tx_uart_data[UART_DATA_LENGTH];
uint32 command[COMMAND LENGTH];
. . .
Tx channel init (&channel, &Tx control,
           Tx data, DATA LENGTH,
           Tx sample, SAMPLE LENGTH,
           Tx uart data, UART DATA LENGTH,
           command, COMMAND LENGTH);
```

## Rx\_channel\_init

### Call(s):

### **Arguments:**

### Table 2-20. Rx channel init arguments

channel	in	Pointer to the Channel control data structure
Rx_control	in	Pointer to the Receiver control data structure
data	in	Pointer to the Rx_data[] circular buffer
data_length	in	Length of the Rx_data[] circular buffer
sample	in	Pointer to the Rx_sample[] circular buffer
sample_length	in	Length of the Rx_sample[] circular buffer
uart_data	in	Pointer to the Rx_uart_data[] circular buffer
uart_data_length	in	Length of the Rx_uart_data[] circular buffer

### **Description:**

This function initializes the Receiver control data structure. This function should be called before calling Rx\_data\_pump(), Rx\_data\_handler() and Receiver Protocol initialization functions.

This function is called once in *main()*.

**Returns:** 

None.

#### **Code example:**

. . .

```
#define DATA_LENGTH (60)
#define SAMPLE_LENGTH (400)
#define COMMAND_LENGTH (40)
```

### Freescale Semiconductor, Inc.

#### Interface

# **S\_registers\_init**

### Call(s):

void S\_registers\_init(void);

Arguments: None.

**Description:** This function initializes the global S[] (registers) data array to the default values (profile) that are contained in the S\_FACTORY\_DEFAULT[] data array (Factory-defined Configuration of S registers). This function is called by the *global\_structure\_init()* function and by the Command Handler on the ATZ or AT&F commands.

Returns: None.

```
...
S_registers_init();
...
```

# **H\_registers\_init**

### Call(s):

void H registers init(void);

Arguments: None.

**Description:** This function initializes the global H[] (hidden registers, that are not directly accessible to the user via AT commands) data array to the default values (profile) that are contained in the H\_FACTORY\_DEFAULT[] data array (Factory-defined Configuration of H registers). This function is called by the *global\_structure\_init()* function and the Command Handler on ATZ or AT&F commands.

Returns: None.

```
...
H_registers_init();
...
```

## Tx\_reset

### Call(s):

```
void Tx reset(struct channel t * channel);
```

### **Arguments:**

#### Table 2-21. Tx\_reset arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This function resets most of the used fields of the Transmitter control data structure (sample\_head, sample\_tail, data\_head, data\_tail, command\_head, command\_tail, data\_pump\_ptr, data\_pump\_call\_func\_state\_number\_samples\_baud\_rate

data\_pump\_call\_func, state, number\_samples, baud\_rate, process\_count, n\_bits, n\_bits\_mask) pointed to by the channel->Tx control ptr to the default values.

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
...
Tx_reset (&channel);
...
```

## Rx\_reset

### Call(s):

void Rx reset(struct channel t \* channel);

### **Arguments:**

### Table 2-22. Rx\_reset arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This function resets most of the used fields in the Receiver control data structure (sample\_head, sample\_tail, data\_head, data\_tail, data\_pump\_ptr, data\_pump\_call\_func, state, number\_samples, baud\_rate, process\_count, n\_bits, n\_bits\_mask, connection\_code) pointed to by the channel->Rx control ptr to the default values.

This function is called in *state\_machine()*.

**Returns:** None.

```
...
struct channel_t channel;
...
Rx_reset (&channel);
...
```

## Tx\_data\_pump

### Call(s):

```
enum Tx_result_t {TX_OK, TX_SAMPLE_BUF_FULL, TX_STATE_COMPLETED};
enum Tx_result_t Tx_data_pump (struct channel_t * channel);
```

### **Arguments:**

### Table 2-23. Tx\_data\_pump arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This is the Transmitter (Tx) Data Pump entry point. A call to this function will execute, if possible, the function

Tx\_control->data\_pump\_call\_func associated with the current Tx Data Pump. The Tx\_data\_pump function contains logic to check the condition of Tx control->sample head and

Tx\_control->sample\_tail to determine if the number of samples specified by Tx\_control->number\_samples can be placed into the Tx\_sample[] buffer yet. If the distance between the two pointers is equal to or less than the value Tx\_control->number\_samples, it calls the Tx control->data pump call func.

This function is called in the *state\_machine()*.

#### **Returns:**

- TX\_OK The Tx\_control->data\_pump\_call\_func was called.
- TX\_SAMPLE\_BUF\_FULL The Tx\_control->data\_pump\_call\_func was not called, because Tx\_sample[] is already full.
- TX\_STATE\_COMPLETED The current state of Tx Data Pump is complete, this means that Tx\_control->process\_count is equal to 0.

```
struct channel_t channel;
...
if(Tx_data_pump (&channel) == TX_STATE_COMPLETED)
{
    ...
}
```

### Rx\_data\_pump

### Call(s):

```
enum Rx_result_t {RX_OK, RX_SAMPLE_BUF_NOT_FILED, RX_STATE_COMPLETED};
enum Rx_result_t Rx_data_pump (struct channel_t * channel)
```

### **Arguments:**

### Table 2-24. Rx\_data\_pump arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This is the Receiver (Rx) Data Pump entry point. A call to this function will execute, if possible, the function

Rx\_control->data\_pump\_call\_func associated with the current Rx Data Pump. The Rx\_data\_pump function contains logic to check the condition of Rx control->sample head and

Rx\_control->sample\_tail to determine if the number of samples specified by Rx\_control->number\_samples are present in the Rx\_sample[] buffer yet. If the distance between the two pointers is equal to or more than the value Rx\_control->number\_samples, it calls the Rx\_control->data\_pump\_call\_func.

This function is called in *state\_machine()*.

#### **Returns:**

- RX\_OK The Rx\_control->data\_pump\_call\_func was called normally.
- RX\_SAMPLE\_BUF\_NOT\_FILED The Rx\_control->data\_pump\_call\_func was not called, because Rx\_sample[] is not full enough.
- RX\_STATE\_COMPLETED The current state of the Rx Data Pump is complete, this means that Rx\_control->process\_count is equal to 0.

```
struct channel_t channel;
...
if(Rx_data_pump (&channel) == RX_STATE_COMPLETED)
{
    ...
}
```

## Tx\_data\_handler

### Call(s):

```
enum Tx_DH_result_t {TX_DH_OK, TX_DATA_BUF_FULL};
enum Tx DH result t Tx data handler (struct channel t * channel);
```

### **Arguments:**

### Table 2-25. Tx\_data\_handler arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This is the Transmitter Data Handler entry point. A call to this function will execute, if possible, the function

Tx\_control->data\_handler\_call\_func associated with the current Tx Data Handler. The Tx\_data\_handler function contains logic to check the condition of Tx control->data head and

Tx\_control->data\_tail to determine if the number of symbols specified by Tx\_control->number\_n\_bits can be placed into the Tx\_data[] buffer yet. If the distance between the two pointers is equal to or less than the value Tx\_control->number\_n\_bits, it calls the Tx control->data handler call func.

This function is called in *state\_machine()*.

#### **Returns:**

- TX\_DH\_OK The Tx\_control->data\_handler\_call\_func was called.
- TX\_DATA\_BUF\_FULL The Tx\_control->data\_handler\_call\_func was not called, because Tx\_data[] is already full.

```
struct channel_t channel;
...

Tx_data_handler (&channel);
...
```

### Rx\_data\_handler

### Call(s):

```
enum Rx_DH_result_t {RX_DH_OK, RX_DH_NO_DATA};
enum Rx DH result t Rx data handler (struct channel t * channel);
```

### **Arguments:**

### Table 2-26. Rx\_data\_handler arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This is the Receiver Data Handler entry point. A call to this function will execute, if possible, the function <code>Rx\_control->data\_handler\_call\_func</code> associated with the current Rx Data Handler. The <code>Rx\_data\_handler</code> function contains logic to check the condition of <code>Rx\_control->data\_head</code> and <code>Rx\_control->data\_tail</code> to determine if the number of symbols specified by <code>Rx\_control->number\_n\_bits</code> are present in the <code>Rx\_data[]</code> buffer yet. If the distance between the two pointers is equal to or more than the value <code>Rx\_control->number\_n\_bits</code>, it calls the <code>Rx\_control->data\_handler\_call\_func</code>.

This function is called in *state\_machine()*.

#### **Returns:**

- RX\_DH\_OK The Rx control->data handler call func was called normally.
- RX\_DH\_NO\_DATA The Rx\_control->data\_handler\_call\_func was not called, because Rx\_data[] is not full enough.

```
struct channel_t channel;
...
Rx_data_handler (&channel);
```

### command\_parser

### Call(s):

```
void command_parser (struct channel_t * channel);
```

#### **Arguments:**

### Table 2-27. command\_parser arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This is the Command Parser entry point. This function calls the

function Tx\_control->command\_parser\_call\_func associated with

the current Command Parser only.

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
...
command_parser(&channel);
...
```

### command\_handler

### Call(s):

### **Arguments:**

### Table 2-28. command\_handler arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This is the Command Handler entry point. A call to this function will execute, if possible, the function

Tx\_control->command\_handler\_call\_func associated with the current Command Handler. The command\_handler function contains logic to check the condition of Tx\_control->command\_head and Tx\_control->command\_tail to determine the existence of elements in the command[] buffer. If the command[] buffer is not empty, it calls the Tx\_control->command\_handler\_call\_func.

This function is called in *state\_machine()*.

#### **Returns:**

- TX\_COM\_HANDLER\_OK The Tx\_control->command\_handler\_call\_func was called.
- TX\_COM\_HANDLER\_NO\_COMMAND The Tx\_control->command\_handler\_call\_func was not called, because command[] is empty.

```
struct channel_t channel;
...
command_handler (&channel);
```

## Tx\_silence\_gen

### Call(s):

```
void Tx silence gen (struct channel t * channel);
```

### **Arguments:**

### Table 2-29. Tx\_silence\_gen arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This function generates silence (samples of 0) into the Tx\_sample[] buffer. The number of generated samples, per call is defined by  $Tx\_control->number\_samples$ . This function is the default value for the  $channel->Tx\_control\_ptr->data\_pump\_call\_func$ .

This function is called by the  $Tx\_data\_pump()$  function via the  $channel->Tx\_control\_ptr-> data\_pump\_call\_func()$ .

**Returns:** None.

```
struct channel_t channel;
...
Tx_silence_gen (&channel);
...
```

### Rx\_idle

### Call(s):

```
void Rx idle (struct channel t * channel);
```

### **Arguments:**

### Table 2-30. Rx\_idle arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** 

This function reads and discards samples from Rx\_sample[]. The

number of samples is defined by

*channel->Rx\_control\_ptr->number\_samples*. This function is the

default value for Rx\_control->data\_pump\_call\_func.

This function is called by the  $Rx\_data\_pump()$  function via the

channel->Rx\_control\_ptr-> data\_pump\_call\_func ().

**Returns:** None.

```
struct channel_t channel;
...
Rx_idle (&channel);
...
```

### save\_data\_handler\_parameters

### Call(s):

```
void save_data_handler_parameters (struct channel_t * channel,
struct data_handler_parameters_t * data_handler_parameters);
```

### **Arguments:**

Table 2-31. save\_data\_handler\_parameters arguments

channel	in	Pointer to the Channel control data structure
data_handler_parameter s	in	Pointer to the Data handler parameters structure

**Description:** This function copies the parameters of channel->Tx\_control\_ptr

and channel->Rx\_control\_ptr, concerning the current Data

Handler (data\_handler\_ptr, data\_handler\_call\_func,

number\_n\_bits), to the data handler parameters structure. This

function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
struct data_handler_parameters_t data_handler_parameters;
...
save_data_handler_parameters (&channel, &data_handler_parameters);
...
```

### load\_data\_handler\_parameters

### Call(s):

```
void load_data_handler_parameters (struct channel_t * channel,
struct data handler parameters t * data handler parameters);
```

### **Arguments:**

Table 2-32. load\_data\_handler\_parameters arguments

channel	in	Pointer to the Channel control data structure
data_handler_parameters	in	Pointer to the Data handler parameters structure

**Description:** This function loads the parameters of

channel->Tx control ptr and

channel->Rx\_control\_ptr, concerning the Data Handler (data\_handler\_ptr, data\_handler\_call\_func, number\_n\_bits), from

the data handler parameters Structure.

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
struct data_handler_parameters_t data_handler_parameters;
...
save_data_handler_parameters (&channel, &data_handler_parameters);
...
load_data_handler_parameters (&channel, &data_handler_parameters);
```

# **Chapter 3 Module Descriptions**

The following sections describe each module associated with the Soft Modem including ITU-T V-series modems, asynchronous-synchronous converters, general telephony call progress signal generators/detectors and other support modules.

### 3.1 Data Pump

The data pump is the heart of the modem, which converts data into a format suitable for transmission over an analogue line. Likewise, it converts the analogue signal from the line back into digital data.

Modulation protocols determine how the modem converts digital data into analog signals that can be sent over a phone line. The protocol standards implemented in the LDR Soft Modem are summarized in Table 1-1. The data rates of the LDR Soft Modem specified by the standards vary from 300bps(V.21) to 2400bps(V.22bis).

### 3.1.1 V.21

The V.21 module implements the ITU V.21 recommendation for a speed of up to 300bps for use in the GSTN. The V.21 implementation uses Frequency Shift Keying modulation with the symbol (baud) rate equal to the bit rate.

It is a 2-channel modem:

- For channel No.1, the mean frequency is 1080 Hz (low channel)
- For channel No.2, the mean frequency is 1750 Hz (high channel)

The frequency deviation is  $\pm 100$  Hz. In each channel, the higher characteristic frequency corresponds to a binary 0. Channel No.1 is used for transmission of the caller's data (i.e. the person making the telephone call) towards the called station, while channel No.2 is used for transmission in the other direction.

The V.21 module initializes the FSK Data Pump module, and also includes a Handshake Data Handler for the process of negotiating a connection.

The V.21 Data Pump is implemented in the FSK module (see chapter 3.1.5).

### 3.1.1.1 Establishment of connection

The V.21 recommendation does not describe how a connection is established.

The procedure used by the LDR Soft Modem is described below and illustrated in Figure 3-1.

- Calling modem:
  - On connection to the line, the calling modem shall be conditioned to receive signals in the high channel and transmit signals in the low channel.
  - After 155±10 ms of binary 1's have been detected, the modem shall remain silent for a further 456±10 ms then it shall transmit binary 1's.
  - 755±10 ms (defined by 155ms+S[9]\*100 ms) later, the modem shall be ready to transmit and receive data.
- Answering modem:
  - On connection to the line, the answering modem shall be conditioned to transmit signals in the high channel and receive signals in the low channel. Following transmission of the answer sequence, the modem shall transmit binary 1's.
  - After 155±10 ms of binary 1's have been detected, the modem shall be ready to transmit and receive data, after waiting a further 600±10 ms (defined by S[9]\*100 ms).

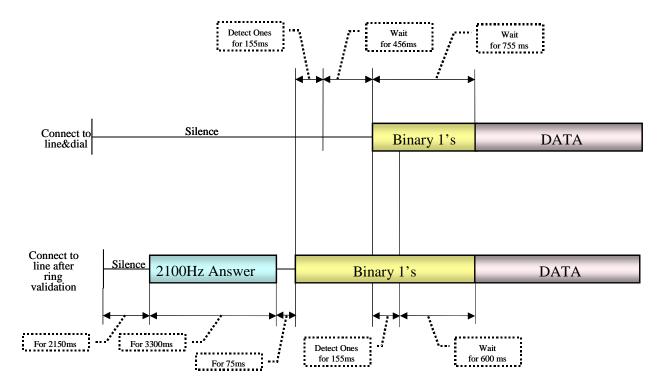


Figure 3-1. Timing diagram of the V.21 Handshake sequence

The V.21 Handshake routines can be found in the "v21.c" file.

The Rx V.21 Handshake control data structure  $Rx_V21_DH_handshake_t$  is defined in the "v21.h" file:

```
/*************************
* States of Rx V21 Handshake Data Handler
*********************
                                       Rx V21 hadshake state t
{RX V21 DH DETECT ONES,RX V21 DH WAIT SEND,RX V21 DH WAIT CARRIER};
/****************************
   Rx V.21 Handshake control data Structure
************************
struct Rx V21 DH handshake t
{
  enum Rx V21 hadshake state t state;
  uint32 det ones ms;
  uint32 wait send ms;
  uint32 wait carrier ms;
  uint32 current counter;
  uint32 global counter
};
```

The Rx\_V21\_DH\_handshake\_t structure parameter descriptions:

- **state** Current state of the Rx V21 Handshake Data Handler. It can be equal to:
  - RX\_V21\_DH\_DETECT\_ONES = The Handshaker is trying to detect <det\_ones\_ms> ms of binary 1's.
  - RX\_V21\_DH\_WAIT\_SEND = The Handshaker remains silent for a <wait\_send\_ms> ms, before transmitting binary 1's.
  - RX\_V21\_DH\_WAIT\_CARRIER = The Handshaker remains in this state during <wait\_carrier\_ms> ms, on the completion of the state, the modem shall be ready to transmit and receive actual data.

It is initialized in Rx\_V21\_handshake\_init() to RX\_V21\_DH\_DETECT\_ONES.

#### **Data Pump**

- **det\_ones\_ms** Duration of binary 1 detection, in ms. It is initialized in *Rx\_V21\_handshake\_init()* to 155 ms.
- wait\_send\_ms Time before starting to transmit binary 1's, in ms. It is initialized in *Rx\_V21\_handshake\_init()* to 456 ms, if the receiver is in calling mode, otherwise it is initialized to 0.
- wait\_carrier\_ms Time before the modem goes into the data mode (ready to transmit and receive actual data) in ms. It is initialized in *Rx\_V21\_handshake\_init()* to 755 ms, if the receiver is in calling mode, otherwise it is initialized to 600 ms.
- **current\_counter** Counter used in the current state for the calculation of binaries conducive to completion of the current state (For example, number of detected binary 1's in the RX\_V21\_DH\_DETECT\_ONES state). It is initialized in *Rx\_V21\_handshake\_init()* to 0.
- **global\_counter** Counter of bits received throughout the handshaking process. It is initialized in *Rx\_V21\_handshake\_init()* to 0.

### Tx\_V21\_init

### Call(s):

```
void Tx_V21_init (struct channel_t * channel,
struct Tx_control_FSK_t *Tx_control_FSK,
bool calling);
```

### **Arguments:**

### Table 3-1. Tx\_V21\_init arguments

channel	in	Pointer to the Channel control data structure
Tx_control_FSK	in	Pointer to the FSK transmitter control data structure
calling	in	Contains TRUE if the transmitter is in calling mode, and FALSE if the transmitter is in answering mode.

### **Description:**

This function sets the fields of the FSK transmitter control data structure, pointed to by  $Tx\_control\_FSK$  ( $f\_space, f\_mark$ , amplitude), and calls  $FSK\_modulator\_init()$ . It also initializes the fields of the Channel control data structure, pointed to by the channel, that are responsible for the Tx Data Pump ( $data\_pump\_ptr$ ,  $baud\_rate$ ,  $number\_samples$ ,  $data\_pump\_call\_func$ , state,  $n\_bits$ ,  $n\_bits\_mask$ ,  $process\_count$ ).

After calling this function, the Tx Data Pump is initialized to work according to V.21.

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
struct Tx_control_FSK_t Tx_control_FSK;
bool calling=TRUE;
...
Tx_V21_init (&channel, &Tx_control_FSK, calling);
...
```

### Rx\_V21\_init

### Call(s):

### **Arguments:**

### Table 3-2. Rx\_V21\_init arguments

channel	in	Pointer to the Channel control data structure
Rx_control_FSK	in	Pointer to the FSK receiver control data structure
calling	in	Contains TRUE if the receiver is in calling mode, and FALSE if the receiver is in answering mode.

### **Description:**

This function sets the fields of the FSK receiver control data structure, pointed to by the  $Rx\_control\_FSK$  ( $f\_space, f\_mark, lpf\_coef, filter\_size$ ,), and calls  $FSK\_demodulator\_init()$ . It also initializes the fields of the Channel control data structure, pointed to by the channel, that are responsible for the Rx Data Pump ( $data\_pump\_ptr, baud\_rate, number\_samples, data\_pump\_call\_func, state, n\_bits, n\_bits\_mask, process\_count$ ).

After calling this function, the Rx Data Pump is initialized to work according to V.21.

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
struct Rx_control_FSK_t Rx_control_FSK;
bool calling=TRUE;
...
Rx_V21_init (&channel, &Rx_control_FSK, calling);
...
```

# Tx\_V21\_handshake\_init

### Call(s):

void Tx V21 handshake init (struct channel t \* channel);

Handshake sequence.

#### **Arguments:**

### Table 3-3. Tx\_V21\_handshake\_init arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This function initializes the fields of the Channel control data structure, pointed to by the *channel*, that are responsible for the Tx Data Handler (*data\_handler\_ptr*, *data\_handler\_state*, *data\_handler\_call\_func*, *number\_n\_bits*). After calling this function, the Tx Data Handler is initialized to perform the Tx V.21

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
...

Tx_V21_handshake_init (&channel);
...
```

# Tx\_V21\_handshake\_routine

### Call(s):

void Tx V21 handshake routine(struct channel t \* channel);

### **Arguments:**

### Table 3-4. Tx\_V21\_handshake\_routine arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This is the Tx V.21 Handshake Data Handler routine. This function

places 0xFF values (sequence of 1's) into the Tx\_data[] buffer.

This function is called by the  $Tx\_data\_handler()$  function via the

 $channel-> Tx\_control\_ptr-> data\_handler\_call\_func~().$ 

**Returns:** None.

### **Code example:**

```
...
struct channel_t channel;
...
Tx_V21_handshake_routine(&channel);
...
```

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### Rx\_V21\_handshake\_init

### Call(s):

### **Arguments:**

### Table 3-5. Rx\_V21\_handshake\_init arguments

channel	in	Pointer to the Channel control data structure
Rx_V21_handshake	in	Pointer to the Rx V.21 Handshake control data structure
calling	in	Contains TRUE if the receiver is in calling mode, and FALSE if the receiver is in answering mode.

### **Description:**

This function sets the fields of the Rx V.21 Handshake control data structure, pointed to by  $Rx\_V21\_handshake$  ( $state, det\_ones\_ms, wait\_send\_ms, wait\_carrier\_ms, current\_counter, global\_counter$ ). It also initializes the fields of the Channel control data structure, pointed to by the channel, that are responsible for the Rx Data Handler ( $data\_handler\_ptr, data\_handler\_call\_func, data\_handler\_state, number\_n\_bits$ ). After calling this function, the Rx Data Handler is initialized to perform the Rx V.21 Handshake sequence.

This function is called in *state machine()*.

**Returns:** None.

```
struct channel_t channel;
struct Rx_V21_DH_handshake_t Rx_V21_handshake;
bool calling=TRUE;
...
Rx_V21_handshake_init (&channel, &Rx_V21_handshake, calling);
...
```

### Rx\_V21\_handshake\_routine

### Call(s):

void Rx V21 handshake routine(struct channel t \* channel);

#### **Arguments:**

### Table 3-6. Rx\_V21\_handshake\_routine arguments

channel	in	Pointer to the Channel control data structure

**Description:** This is the Rx V.21 Handshake Data Handler routine. This function

reads data (symbols) from the Rx\_data[] buffer. The function performs and controls the V.21 Handshake sequence (see 3.1.1.2.).

After successful completion of the Handshake sequence, it sets

*channel->Rx\_control\_ptr->data\_handler\_state* to RX\_DH\_STATE\_COMPLETED, otherwise to

RX\_DH\_STATE\_FAILED.

This function is called by the  $Rx\_data\_handler()$  function via the

channel->Rx\_control\_ptr-> data\_handler\_call\_func ().

**Returns:** None.

```
struct channel_t channel;

...

Rx_V21_handshake_routine(&channel);
...
```

### 3.1.2 V.23

The V.23 module implements the ITU V.23 recommendation for speeds of 600bps and 1200bps (it also supports a backward channel of 75bps) for use in the GSTN. The V.23 implementation uses Frequency Shift Keying modulation with the symbol (baud) rate equal to the bit rate.

It is a 2-channel modem:

- For channel No.1 (backward): The modulation rate is 75 bauds; the mean frequency is 420 Hz; the frequency deviation is  $\pm 30$  Hz
- For channel No.2 (forward):
  - Mode 1: The modulation rate is 600 bauds; the mean frequency is 1500 Hz; the Frequency deviation is  $\pm 200$  Hz
  - Mode 2: The modulation rate is 1200 bauds; the mean frequency is 1700 Hz; the frequency deviation is  $\pm 400$  Hz

In each channel, the higher characteristic frequency corresponds to a binary 0. Channel No.1 is used for transmitting the caller's data (i.e. the person making the telephone call) towards the called station, while channel No.2 is used for transmission in the other direction.

Most hardware modems support Mode 2 only (1200bps) for channel No.2.

The V.23 module initializes the FSK Data Pump module, and also includes a Handshake Data Handler for the process of negotiating a connection.

The V.23 Data Pump is implemented in the FSK module (see chapter 3.1.5).

#### 3.1.2.1 Establishment of connection

The V.23 recommendation does not describe how a connection is established. The procedure used by the LDR Soft Modem is described below and illustrated in Fig 3.1.2.2.

- Calling modem:
  - On connection to the line, the calling modem shall be conditioned to receive signals in the high channel and transmit signals in the low channel.
  - After 155±10 ms of binary 1's have been detected, the modem shall remain silent for a further 456±10 ms then it shall transmit binary 1's.
  - 755±10 ms (defined by 155ms+S[9]\*100 ms) later, the modem shall be ready to transmit and receive data.
- Answering modem:
  - On connection to the line, the answering modem shall be conditioned to transmit signals in the high channel and receive signals in the low channel. Following transmission of the answer sequence the modem shall transmit binary 1's.

— After 155±10 ms of binary 1's have been detected, the modem shall be ready to transmit and receive data after waiting a further 600±10 ms (defined by S[9]\*100 ms).

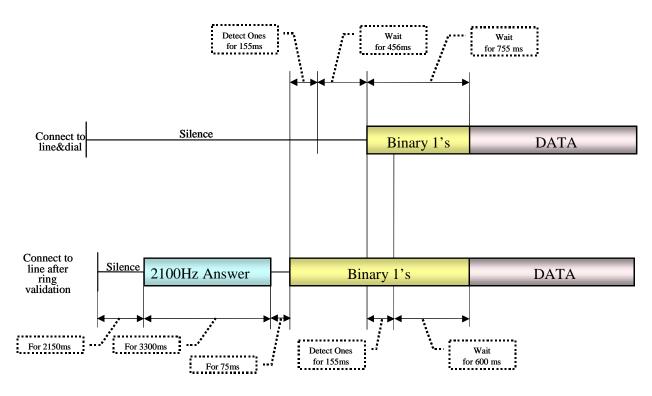


Figure 3-2. Timing diagram of the V.23 Handshake sequence

The V.23 Handshake routines can be found in the "v23.c" file.

The Rx V.23 Handshake control data structure  $Rx_V23_DH_handshake_t$  is defined in the "v23.h" file:

```
enum Rx_V23_hadshake_state_t state;
uint32 det_ones_ms;
uint32 wait_send_ms;
uint32 wait_carrier_ms;

uint32 current_counter;
uint32 global_counter
};
```

The Rx\_V23\_DH\_handshake\_t structure parameter descriptions:

- **state** Current state of the Rx V23 Handshake Data Handler. It can be equal to:
  - RX\_V23\_DH\_DETECT\_ONES = The Handshaker is trying to detect <det\_ones\_ms> ms of binary 1's.
  - RX\_V23\_DH\_WAIT\_SEND = The Handshaker remains silent for a period of <wait\_send\_ms> ms, before transmitting binary 1's.
  - RX\_V23\_DH\_WAIT\_CARRIER = The Handshaker remains in this state during <wait\_carrier\_ms> ms, on the completion of the state, the modem shall be ready to transmit and receive actual data.

It is initialized in *Rx\_V23\_handshake\_init()* to RX\_V23\_DH\_DETECT\_ONES.

- **det\_ones\_ms** Duration of binary 1's detection, in ms. It is initialized in *Rx\_V23\_handshake\_init()* to 155 ms.
- wait\_send\_ms Time before starting to transmit binary 1's, in ms. It is initialized in Rx\_V23\_handshake\_init() to 456 ms, if the receiver is in calling mode, otherwise it is initialized to 0.
- wait\_carrier\_ms Time before the modem goes into the data mode (ready to transmit and receive actual data) in ms. It is initialized in *Rx\_V23\_handshake\_init()* to 755 ms, if the receiver is in calling mode, otherwise it is initialized to 600 ms.
- **current\_counter** Counter used in the current state for the calculation of binaries conducive to completion of the current state (For example, number of binary 1's in the RX\_V23\_DH\_DETECT\_ONES state). It is initialized in *Rx\_V23\_handshake\_init()* to 0.
- **global\_counter** Counter of bits received throughout the handshaking process. It is initialized in  $Rx_V23_handshake_init()$  to 0.

### Tx\_V23\_init

### Call(s):

### **Arguments:**

### Table 3-7. Tx\_V23\_init arguments

channel	in	Pointer to the Channel control data structure	
Tx_control_FSK	in	Pointer to the FSK transmitter control data structure	
calling	in	Contains TRUE if the transmitter is in calling mode, and FALSE if the transmitter is in answering mode.	
mode	in	Defines mode for channel No.2. It can be equal to:  • V23_MODE_600 = Mode 1 (600 bauds).  • V23_MODE_1200 = Mode 2 (1200 bauds).	

### **Description:**

This function sets the fields of the FSK transmitter control data structure, pointed by  $Tx\_control\_FSK$  ( $f\_space$ ,  $f\_mark$ , amplitude), and calls  $FSK\_modulator\_init()$ . It also initializes the fields of the Channel control data structure, pointed to by the *channel*, that are responsible for the Tx Data Pump ( $data\_pump\_ptr$ ,  $baud\_rate$ ,  $number\_samples$ ,  $data\_pump\_call\_func$ , state,  $n\_bits$ ,  $n\_bits\_mask$ ,  $process\_count$ ).

After calling this function, the Tx Data Pump is initialized to work according to V.23.

This function is called in *state\_machine()*.

#### **Returns:**

None.

```
struct channel_t channel;
struct Tx_control_FSK_t Tx_control_FSK;
bool calling=TRUE;
...
Tx_V23_init (&channel, &Tx_control_FSK, calling, V23_MODE_600);
...
```

### Rx\_V23\_init

### Call(s):

### **Arguments:**

### Table 3-8. Rx\_V23\_init arguments

channel	in	Pointer to the Channel control data structure	
Rx_control_FSK	in	Pointer to the FSK receiver control data structure	
calling	in	Contains TRUE if the receiver is in calling mode, and FALSE if the receiver is in answering mode.	
mode	in	Defines mode for the channel No.2. It can be equal to:  • V23_MODE_600 = Mode 1 (600 bauds).  • V23_MODE_1200 = Mode 2 (1200 bauds).	

### **Description:**

This function sets the fields of the FSK receiver control data structure, pointed to by the  $Rx\_control\_FSK$  ( $f\_space, f\_mark$ ,  $lpf\_coef$ ,  $filter\_size$ ), and calls  $FSK\_demodulator\_init()$ . It also initializes the fields of the Channel control data structure, pointed to by the *channel*, that are responsible for the Rx Data Pump

(data\_pump\_ptr, baud\_rate, number\_samples,

data\_pump\_call\_func, state, n\_bits, n\_bits\_mask, process\_count).

After calling this function, the Rx Data Pump is initialized to work according to V.23.

This function is called in *state\_machine()*.

#### **Returns:**

None.

```
...
struct channel_t channel;
struct Rx_control_FSK_t Rx_control_FSK;
bool calling=TRUE;
...
Rx_V23_init (&channel, &Rx_control_FSK, calling, V23_MODE_600);
...
```

# Tx\_V23\_handshake\_init

### Call(s):

void Tx V23 handshake init (struct channel t \* channel);

### **Arguments:**

### Table 3-9. Tx\_V23\_handshake\_init arguments

	1	
channel	in	Pointer to the Channel control data structure

**Description:** This function initializes the fields of the Channel control data

structure, pointed to by the *channel*, that are responsible for the Tx

Data Handler (data\_handler\_ptr, data\_handler\_state,

data\_handler\_call\_func, number\_n\_bits). After calling this

function, the Tx Data Handler is initialized to perform the Tx V.23

Handshake sequence.

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
...
Tx_V23_handshake_init (&channel);
...
```

# Tx\_V23\_handshake\_routine

### Call(s):

void Tx V23 handshake routine(struct channel t \* channel);

### **Arguments:**

### Table 3-10. Tx\_V23\_handshake\_routine arguments

channel in	Pointer to the Channel control data structure
------------	---

**Description:** This is the Tx V.23 Handshake Data Handler routine. This function

places 0xFF values (sequence of 1's) into the Tx\_data[] buffer.

This function is called by the  $Tx\_data\_handler()$  function via the

channel->Tx\_control\_ptr-> data\_handler\_call\_func().

**Returns:** None.

```
struct channel_t channel;
...

Tx_V23_handshake_routine(&channel);
...
```

### Rx\_V23\_handshake\_init

### Call(s):

### **Arguments:**

#### Table 3-11. Rx\_V23\_handshake\_init arguments

channel	in	Pointer to the Channel control data structure	
Rx_V23_handshake	in	Pointer to the Rx V.23 Handshake control data structure	
calling	in	Contains TRUE if the receiver is in calling mode, and FALSE if the receiver is in answering mode.	

### **Description:**

This function sets the fields of the Rx V.23 Handshake control data structure, pointed to by  $Rx\_V23\_handshake$  (state,  $det\_ones\_ms$ ,  $wait\_send\_ms$ ,  $wait\_carrier\_ms$ ,  $current\_counter$ ,  $global\_counter$ ). It also initializes the fields of the Channel control data structure, pointed to by the channel, that are responsible for the Rx Data Handler ( $data\_handler\_ptr$ ,  $data\_handler\_call\_func$ ,  $data\_handler\_state$ ,  $number\_n\_bits$ ). After calling this function, the Rx Data Handler is initialized to perform the Rx V.23 Handshake sequence.

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
struct Rx_V23_DH_handshake_t Rx_V23_handshake;
bool calling=TRUE;
...
Rx_V23_handshake_init (&channel, &Rx_V23_handshake, calling);
...
```

### Rx\_V23\_handshake\_routine

### Call(s):

void Rx V23 handshake routine(struct channel t \* channel);

### **Arguments:**

### Table 3-12. Rx\_V23\_handshake\_routine arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This is the Rx V.23 Handshake Data Handler routine. This function reads data (symbols) from the Rx\_data[] buffer. The function performs and controls the V.23 Handshake sequence (see 3.1.2.2.).

After successful completion of the Handshake sequence, it sets

*channel->Rx\_control\_ptr->data\_handler\_state* to RX\_DH\_STATE\_COMPLETED, otherwise to

RX\_DH\_STATE\_FAILED.

This function is called by the *Rx\_data\_handler()* function via the

channel->Rx\_control\_ptr-> data\_handler\_call\_func ().

#### **Returns:**

None.

```
struct channel_t channel;
...

Rx_V23_handshake_routine(&channel);
...
```

### 3.1.3 V.22

The V.22 modem is intended for use on connections on the General Switched Telephone Network (GSTN), and on point-to-point circuits when suitably conditioned. This is a full-duplex modem with the receiver and transmitter sharing the available bandwidth of the communication channel. This modem can operate in either originate or answer mode. In the originate mode, it initiates the communication process, transmits with a carrier frequency of 1200 Hz, and receives at the frequency of 2400 Hz. At the other end of the communications channel there is a remote modem in answer mode. This remote modem receives at 1200 Hz and transmits at 2400 Hz.

The Differential Phase Shift Keying (DPSK) modulation technique is used for each channel with synchronous line transmission at 600 baud (implemented in the *dpsk.c* file). The V.22, with a 600-baud rate, accomplishes the transmission of 1200 bps by encoding two incoming bits (a dibit) in a single baud or 600 bps by encoding one incoming bit in a single baud. Since there are four possible values for every dibit (00, 01, 10, 11), the constellation diagram for the V.22 contains four points. Figure 3.1.3.1 shows the constellation diagram for the V.22. The four constellation points notated A, B, C, and D lie on a circle. Since there is no amplitude information transmitted, the radius of this circle is normalized to unity.

A scrambler is included in the input to the transmitter and a descrambler at the output of the receiver. A guard tone of  $1800 \pm 20$  Hz or  $550 \pm 20$  Hz may be used while transmitting only in the high channel (transmitter of the answering modem). The guard tone indicates to an automatic telephone system that the line is occupied by the modem. Fixed compromise equalization (filtering the input/output signal with a fixed filter) shall be incorporated in the modem. Such equalization shall be equally shared between the transmitter and the receiver.

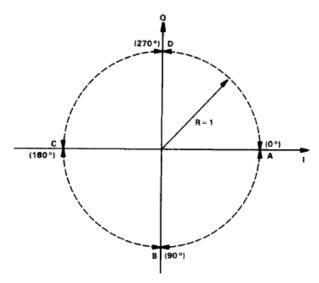


Figure 3-3. Signal Constellation for V.22

For 1200 bits per second the data stream to be transmitted shall be divided into groups of 2 consecutive bits (dibits). Each dibit shall be encoded as a phase change relative to the phase

of the preceding signal element (see table 3.1.3.1). At the receiver, the dibits shall be decoded and the bits reassembled in the correct order. The left-hand digit of the dibit is the one occurring first in the data stream as it enters the modulator portion of the modem after the scrambler.

Table 3-13. Bits encoding in V.22 modem

Dibit values (1200 bit/s)	Bit values (600 bit/s)	Phase change
00	0	90
01	-	0
11	1	270
10	-	180

The phase change is the actual on-line phase shift in the transition region from the center of one signaling element to the center of the following signaling element.

For 600 bits/s each bit shall be encoded as a phase change relative to the phase of the preceding signal element.

### 3.1.3.1 Scrambler

A self-synchronizing scrambler, which has the generating polynomial  $1 \oplus x$  -14  $\oplus x$  -17, is included in the modem transmitter. The purpose of the scrambler is to randomize the input binary data sequence, which means that it converts this sequence into a pseudo-random binary sequence. For example, if the DTE sends a series of 01 dibits, from table 3.1.3.1 it can be seen that each dibit corresponds to a 0-degree phase change. Therefore the total phase transmitted is the same (or the same constellation point). However a phase change is required for correct clock recovery (otherwise the energy of the samples remains constant). To avoid this, the scrambler is introduced to minimize the probability that such 'ill-conditioned' dibits occur. In other words it makes the data stream look like a random stream of ones and zeros regardless of the data being transmitted. This is required for the clock recovery module.

Considering d(nT) is the input to the scrambler, the output  $d_s(nT)$  is given by:

$$d_s(nT) = d(nT) \text{ XOR } d_s((n-14)T) \text{ XOR } d_s((n-17)T)$$

Where T is the data period. The signal flowchart of the modem scrambler is shown in Figure 3-4.

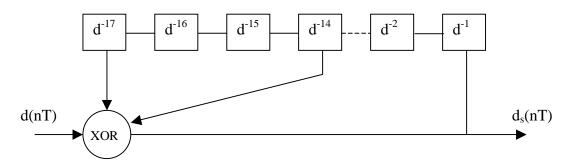


Figure 3-4. V.22 scrambler

#### 3.1.3.2 Descrambler

The descrambler is intended to recover the originally transmitted dibit. The output of the descrambler is described by:

$$d(nT) = d_s(nT) \text{ XOR } d_s((n-14)T) \text{ XOR } d_s((n-17)T)$$

Where T is the data period. The signal flowchart of the modem descrambler is shown in Figure 3-5.

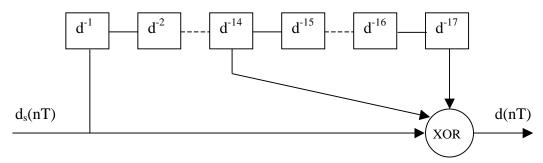


Figure 3-5. V.22 descrambler

### 3.1.3.3 Handshake

The handshake sequence for achieving synchronization between calling and answering modems is shown in Figure 3-6.

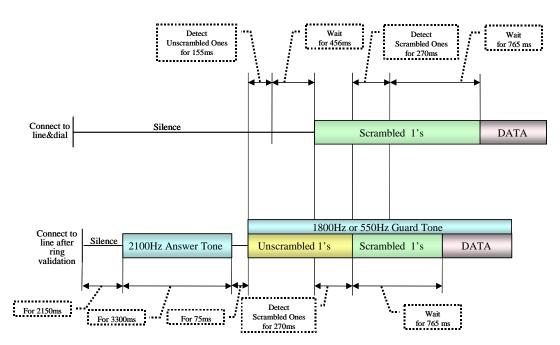


Figure 3-6. Handshake sequence

Handshake sequence for calling modem

- a) Once the calling modem has connected to the line, it shall be conditioned to receive signals in the high channel.
- b) The modem shall remain silent until unscrambled binary 1's are detected for a period of  $155 \pm 50$  ms. After waiting for a further  $456 \pm 10$  ms it shall transmit scrambled binary 1's in the low channel.
- c) Upon detecting scrambled binary 1's in the high channel for a period of  $270 \pm 40$  ms, the modem shall wait a further  $765 \pm 10$  ms, then it is ready to transmit and receive data.

Handshake sequence for answering modem

- a) Once the answering modem has connected to the line, immediately following the V.25 answer sequence, the modem shall be conditioned to receive signals in the low channel. It shall then transmit unscrambled binary 1's.
- b) Upon detecting scrambled binary 1's in the low channel for a period of  $270 \pm 40$  ms, the modem shall transmit scrambled binary 1's in the high channel, and wait for a further  $765 \pm 10$  ms. The modem is then ready to transmit and receive data.
- c) After completion of the handshake sequence, any inadvertent loss and reappearance of the received line signal should not cause another handshake sequence to be generated.

### 3.1.3.4 Implementation

The requirements of V.22 are implemented in the v22.c file. This module is responsible for the initialization of the V.22 handshake control structures ( $Rx(Tx)_{V22}DH_{handshake_t}$ ), the initialization of the DPSK control structures ( $Rx(Tx)_{control}DPSK_t$ , for details about these structures refer to Sections 3.1.6.4/3.1.6.5), scrambling, descrambling, and controlling the handshake process ( $Rx_{V22}_{handshake_routine}$ ).

A description of the main functions and control structures of v22.c are given below.

### 3.1.3.5 RX V.22 Handshake data handler structure

```
struct Rx_V22_DH_handshake_t
{
    enum Rx_V22_hadshake_state_t state;
    uint32 det_unscr_ones_ms;
    uint32 det_scr_ones_ms;
    uint32 wait_send_ms;
    uint32 wait_carrier_ms;

    uint32 current_counter;
    uint32 global_counter;
};
```

- **state** Current state of the Rx V22 Handshake Data Handler. The list of possible states can be seen in Fig 3.1.3.12.
- **det\_unscr\_ones\_ms** Duration of the unscrambled binary 1's detection in ms. This field is initialized in function *Rx\_V22\_handshake\_init*.
- **det\_scr\_ones\_ms** Duration of the scrambled binary 1's detection in ms. This field is initialized in function *Rx\_V22\_handshake\_init*.
- wait\_send\_ms Specified time before starting to transmit scrambled binary 1's in ms. This field is initialized in function *Rx\_V22\_handshake\_init*.
- wait\_carrier\_ms Specified time before Carrier Detect Response in ms. This field is initialized in function *Rx\_V22\_handshake\_init*.
- **current\_counter** The counter that is used for counting the different data sequences (e.g. for detecting consecutive 1's). This field is initialized in function *Rx\_V22\_handshake\_init* to 0.
- **global\_counter** Counter of the bits received throughout the handshake phase. This field is initialized in function  $Rx_V22_handshake_init$  to 0.

## V22\_scramble\_bit

### Call(s):

uint8 V22\_scramble\_bit(struct channel\_t \* channel, uint8 bit)

### **Arguments:**

#### Table 3-14. V22\_scramble\_bit arguments

channel	in	Pointer to the channel control data structure
bit	in	Bit to scramble

**Description:** 

Scrambles input bits according to the V.22 recommendation.

$$d_s(nT) = d(nT) \text{ XOR } d_s((n-14)T) \text{ XOR } d_s((n-17)T).$$

Scrambler register is stored in

channel->Tx\_control\_ptr->data\_pump\_ptr->scrambler\_register.

**Returns:** 

Scrambled bit.

```
uint8 temp_current_nbits, current_nbits;
struct channel_t * channel;
...
temp_current_nbits = V22_scramble_bit(channel, current_nbits);
...
```

# V22\_descramble\_bit

### Call(s):

uint8 V22\_descramble\_bit(struct channel\_t \* channel, uint8 bit)

### **Arguments:**

Table 3-15. V22\_descramble\_bit arguments

channel	in	Pointer to the channel control data structure
bit	in	Bit to descramble

**Description:** 

Descrambles input bits according to the V.22 recommendation.

$$d(nT) = d_s(nT) \text{ XOR } d_s((n-14)T) \text{ XOR } d_s((n-17)T).$$

Descrambler register is stored in

channel->Rx\_control\_ptr->data\_pump\_ptr->descrambler\_register.

**Returns:** 

Descrambled bit.

```
uint8 temp_current_nbits, current_nbits;
struct channel_t * channel;
...
temp_current_nbits = V22_descramble_bit(channel, current_nbits);
...
```

### Tx\_V22\_init

### Call(s):

void Tx\_V22\_init (struct channel\_t \* channel,struct Tx\_control\_DPSK\_t
\*Tx control DPSK, bool calling,enum v22 mode t mode)

### **Arguments:**

### Table 3-16. Tx\_V22\_init arguments

channel	in	Pointer to the channel control data structure	
Tx_control_DPSK	in	Pointer to the DPSK transmitter control data structure	
calling	in	Indicates either modem in call or answer mode	
mode	in	Specifies 600 bit/s (V22_MODE_600) or 1200 bit/s (V22_MODE_1200) mode	

### **Description:**

This function initializes the Transmitter control Structures pointed to by  $Tx\_control\_DPSK$  and channel according to V.22 and calls  $DPSK\_modulator\_init()$ .

It fills the appropriate fields of the  $Tx\_control\_t$  structure, pointed to by  $channel->Tx\_control\_ptr$  (fields:  $n\_bits$ ,  $n\_bits\_mask$ ,  $baud\_rate$ ,  $number\_samples$ ,  $data\_pump\_call\_func$ , state,  $process\_count$ ) and the  $Tx\_control\_DPSK$  structure (fields:  $f\_carrier$ ,  $f\_guard$ ,  $amplitude\_carrier$ ,  $amplitude\_guard$ ,  $omega\_ptr$ , scrambler).

This function is called from *state machine()*.

**Returns:** None

#### \_\_\_\_\_

```
struct channel_t * channel;
struct Tx_control_DPSK_t *Tx_control_DPSK;
...
Tx_V22_init (channel, Tx_control_DPSK, TRUE, V22_MODE_1200);
...
```

### Rx\_V22\_init

### Call(s):

void Rx\_V22\_init (struct channel\_t \* channel,struct Rx\_control\_DPSK\_t
\*Rx control DPSK, bool calling,enum v22 mode t mode)

### **Arguments:**

### Table 3-17. Rx\_V22\_init arguments

channel	in	Pointer to the channel control data structure	
Rx_control_DPSK	in	Pointer to the DPSK receiver control data structure	
calling	in	Indicates either modem in call or answer mode	
mode	in	Specifies 600 bit/s (V22_MODE_600) or 1200 bit/s (V22_MODE_1200) mode	

### **Description:**

This function initializes the Receiver control Structure according to V.22.

It fills the appropriate fields of the  $Rx\_control\_t$  structure, pointed to by  $channel->Rx\_control\_ptr$  (fields:  $n\_bits$ ,  $n\_bits\_mask$ ,  $baud\_rate$ ,  $number\_samples$ ,  $data\_pump\_call\_func$ , state,  $process\_count$ ) and the  $Rx\_control\_DPSK$  structure (fields:  $f\_carrier$ ,  $omega\_ptr$ ,  $noise\_threshold$ , descrambler).

This function is called from *state\_machine()*.

**Returns:** None

```
struct channel_t * channel;
struct Rx_control_DPSK_t *Rx_control_DPSK;
...
Rx_V22_init (channel, Rx_control_DPSK, TRUE, V22_MODE_1200);
```

# Rx\_V22\_handshake\_init

## Call(s):

void Rx\_V22\_handshake\_init (struct channel\_t \* channel, struct
Rx V22 DH handshake t \*Rx V22 handshake, bool calling)

## **Arguments:**

### Table 3-18. Rx\_V22\_handshake\_init arguments

channel	in	Pointer to the channel control data structure
Rx_V22_handshak e	in	Pointer to the RX handshake data handler structure
calling	in	Indicates either modem in call or answer mode

#### **Description:**

This function initializes the receiver V.22 handshake data handler control structure. It fills the appropriate fields of  $Rx\_control\_t$  structure, pointed to by  $channel->Rx\_control\_ptr$  (fields:  $data\_handler\_call\_func$ ,  $data\_handler\_state$ ,  $number\_n\_bits$ ) and the  $Rx\_V22\_handshake$  structure (fields: state,  $det\_scr\_ones\_ms$ ,  $det\_unscr\_ones\_ms$ ,  $wait\_send\_ms$ ,  $wait\_carrier\_ms$ ,  $current\_counter$ ,  $global\_counter$ ). It sets up the data handler for the receiver to perform the  $Rx\_V22\_handshake\_routine$ .

This function is called from *state\_machine()*.

#### **Returns:**

None

```
struct channel_t * channel;
struct Rx_V22_DH_handshake_t Rx_V22_DH_handshake;
...
Rx_V22_handshake_init (channel, &Rx_V22_DH_handshake, TRUE);
...
```

# Rx\_V22\_handshake\_routine

### Call(s):

void Rx V22 handshake routine(struct channel t \* channel)

#### **Arguments:**

Table 3-19. Rx\_V22\_handshake\_routine arguments

d			
	channel	in	Pointer to the channel control data structure

#### **Description:**

This is the RX V.22 Handshake Data Handler routine. This function reads data (symbols) from the Rx\_data[] buffer. It controls the handshake process. The inner state diagram for this routine during the handshake is shown in Figure 3-7. If handshake is completed successfully *Rx\_control->data\_handler\_state* is set to *RX\_DH\_STATE\_COMPLETED*, else to *RX\_DH\_STATE\_FAILED*.

This function is called by the  $Rx\_data\_handler()$  function via the  $channel->Rx\_control\_ptr-> data\_handler\_call\_func()$ .

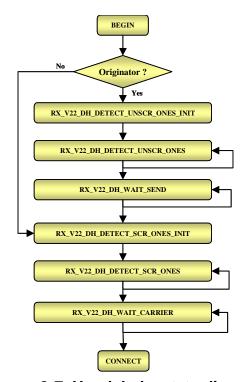


Figure 3-7. Handshake state diagram

**Returns:** None

```
struct channel_t * channel;
...
Rx_V22_handshake_routine (channel);
```

# Tx\_V22\_handshake\_init

## Call(s):

void Tx V22 handshake init (struct channel t \* channel)

None

#### **Arguments:**

#### Table 3-20. Tx\_V22\_handshake\_init arguments

channel	in	Pointer to the channel control data structure
---------	----	---

# **Description:**

This function initializes the transmitter to the V.22 handshake data handler control structure.

It fills the appropriate fields of  $Tx\_control\_t$  structure, pointed to by  $channel->Tx\_control\_ptr$  (fields:  $data\_handler\_call\_func$ ,  $data\_handler\_state$ ,  $number\_n\_bits$ ). It sets up the data handler for the transmitter to perform the  $Tx\_V22\_handshake\_routine$ .

This function is called from *state\_machine()*.

**Returns:** 

# Tx\_V22\_handshake\_routine

# Call(s):

void Tx V22 handshake routine(struct channel t \* channel)

#### **Arguments:**

Table 3-21. Tx\_V22\_handshake\_routine arguments

channel	in	Pointer to the channel control data structure
---------	----	---

**Description:** This is the TX V.22 Handshake Data Handler routine. This function

places 0xFF values (sequence of 1's) to the Tx\_data[] buffer.

This function is called by the  $Tx\_data\_handler()$  function via the

channel->Tx\_control\_ptr-> data\_handler\_call\_func ().

**Returns:** None

```
struct channel_t * channel;
...
Tx_V22_handshake_routine (channel);
```

### 3.1.4 V.22bis

The V.22bis modem is intended for use on connections on the General Switched Telephone Network (GSTN), and on point-to-point 2-wire leased telephone-type circuits. This is a full-duplex modem with the receiver and transmitter sharing the available bandwidth of the communication channel. This modem can operate in either originate or answer mode. In the originate mode, it initiates the communication process, transmits with a carrier frequency of 1200Hz, and receives at the frequency of 2400Hz. At the other end of the communications channel there is a remote modem in answer mode. This remote modem receives at 1200Hz and transmits at 2400Hz.

The Quadrature Amplitude modulation (QAM) technique is used for each channel with synchronous line transmission at 600 baud (implemented in the qam.c file). The constellation could be either 16 points, 4 bits/baud supporting an input bit rate of 2400 bit/s (See Fig 3.1.4.1), or 4 point, 2 bit/baud supporting 1200 bit/s. It is compatible with a V.22 modem at the 1200 bit/s signalling rate and includes automatic bit rate recognition. A scrambler is included in the input to the transmitter and a descrambler at the output of the receiver. A guard tone of  $1800 \pm 20$  Hz or  $550 \pm 20$  Hz may be used while transmitting only in the high channel (transmitter of the answering modem). Fixed compromise equalization and adaptive equalization shall be incorporated in the modem.

For 2400 bits per second the data stream to be transmitted shall be divided into groups of four consecutive bits (quadbits). The first two bits of a quadbit shall be encoded as a phase quadrant change relative to the quadrant occupied by the preceding signal element. (See Fig 3.1.4.1 and Table 3.1.4.1).

The last two bits of each quadbit define one of four signalling elements associated with the new quadrant (see Fig 3.1.4.1).

Example: The first dibit is encoded as a phase change according to table 3.1.4.1 (as in V.22). The second dibit defines the signalling element (the point in a new quadrant). For example, when receiving the current baud, the phase change is 90 degrees from the previous baud (i.e. the current point  $\{I, Q\}$  is in the first quadrant and the previous one was in the fourth, then, the first dibit is 00 (according to Table 3-22). If the I and Q values retrieved point to the top-right point of the top-right quadrant (i.e I=3 and Q=3, see fig. 3.1.4.1) then the second dibit is 11. Therefore the whole quad bit = 0011.

Table 3-22. Bits encoding in V.22bis modem

First two bits in quadbit (2400 bit/s) or dibit values (1200 bit/s)	Phase change
00	90
01	0
11	270
10	180

At the receiver, the dibits are decoded and the bits reassembled in the correct order. The left hand bits in Table 3-22 and Figure 3-8 are the first of each pair in the data stream as it enters the modulator portion of the modem after the scrambler.

For 1200 bits/s the data stream to be transmitted shall be divided into groups of 2 consecutive bits (dibits). The dibits shall be encoded as a phase quadrant change relative to the quadrant occupied by the preceding signal element (see Table 3.12.34). The signalling elements corresponding to 01 in the signal constellation (Figure 3-8) shall be transmitted irrespective of the quadrant concerned. This ensures compatibility with Recommendation V.22.

The phase change is the actual on-line phase shift in the transition region from the center of one signaling element to the center of the following signaling element.

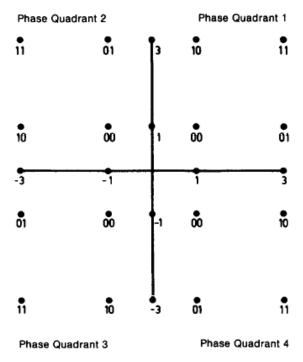


Figure 3-8. V.22bis 16-point constellation diagram

#### 3.1.4.1 Scrambler

A self-synchronizing scrambler, which has the generating polynomial  $1 \oplus x$  -14  $\oplus x$ -17 shall be included in the modern transmitter. The purpose of the scrambler is to randomize input binary data sequence, which means that it converts this sequence in a pseudo-random binary sequence. This is required for the clock recovery module.

Considering d(nT) is the input to the scrambler, the output  $d_s(nT)$  is given by:

$$d_s(nT) = d(nT) \text{ XOR } d_s((n-14)T) \text{ XOR } d_s((n-17)T)$$

where T is the data period. The signal flowchart of the modem scrambler is shown in Figure 3-9.

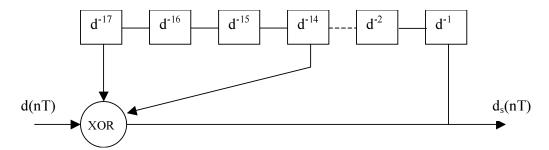


Figure 3-9. V.22bis scrambler

### 3.1.4.2 Descrambler

The descrambler is intended for recovering the originally transmitted quadbit. The output of the descrambler is described by:

$$d(nT) = d_s(nT) \text{ XOR } d_s((n-14)T) \text{ XOR } d_s((n-17)T)$$

Where T is the data period. The signal flowchart of the modem scrambler is shown in Figure 3-10.

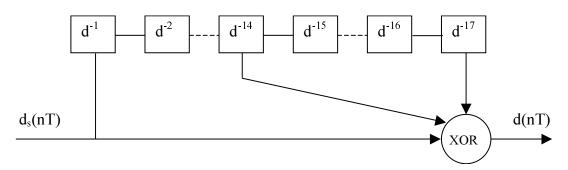


Figure 3-10. V.22bis descrambler

#### 3.1.4.3 Handshake

The handshake sequence for achieving synchronization between calling and answering modems is shown in Figure 3-11. If both calling and answering modems are V.22bis modems, the handshake will normally condition both modems to operate at 2400 bit/s. If however one or both of the modems has been set to operate at 1200 bit/s, then the handshake will condition both modems to operate at 1200 bit/s. If either the calling or answering modem is a V.22 modem the handshake will condition both the V.22bis and V.22 modem to operate at 1200 bit/s.

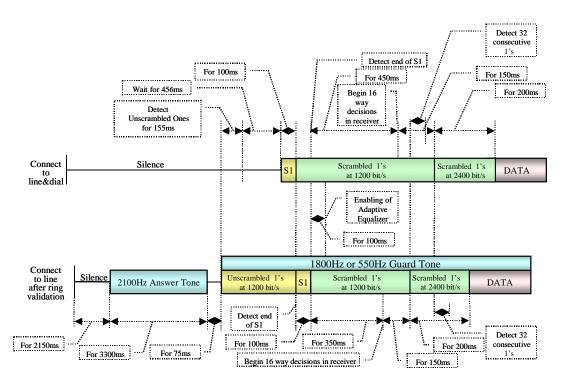


Figure 3-11. V.22bis handshake sequence

Handshake sequence for calling modem (2400 bit/s operation)

- a) On connection to the line the calling modem shall be conditioned to receive signals in the high channel at 1200 bit/s and transmit signals in the low channel at 1200 bit/s. The modem shall initially remain silent.
- b) After  $155 \pm 10$ ms of unscrambled binary 1's have been detected, the modem shall remain silent for a further  $456 \pm 10$ ms then transmit an unscrambled repetitive double dibit pattern of 00 and 11 at 1200 bit/s for  $100 \pm 3$  ms. Following this signal the modem shall transmit scrambled binary 1's at 1200 bit/s.
- c) If the modem detects scrambled binary 1's in the high channel at 1200 bit/s for 270 ± 40ms, the handshake shall continue according to the 1200 bit/s Handshake sequence for a calling modem. However, if unscrambled repetitive double dibit 00 and 11 at 1200 bit/s are detected in the high channel, then the end of receipt of this signal should be detected.
- d)  $600 \pm 10$ ms after detection of the end of repetitive double dibit 00 and 11 at 1200 bit/s in the high channel, the modem shall begin transmitting scrambled binary 1's at 2400 bit/s, and  $450 \pm 10$ ms after detecting this double dibit signal the receiver may begin making 16-way decisions.
- e) Following transmission of scrambled binary 1's at 2400 bit/s for  $200 \pm 10$ ms, the modem shall be ready to transmit data at 2400 bit/s.
- f) When 32 consecutive scrambled 1's at 2400 bit/s have been detected in the high

channel the modem shall be ready to receive data at 2400 bit/s.

Handshake sequence for answering modem (2400 bit/s operation)

- a) On connection to the line the answering modem shall be conditioned to transmit signals in the high channel at 1200 bit/s and receive signals in the low channel at 1200 bit/s. Following transmission of the answer sequence in accordance with Recommendation V.25, the modem shall transmit unscrambled binary 1 at 1200 bit/s.
- b) If the modem detects scrambled binary 1's in the low channel at 1200 bit/s for 270 ± 40ms, the handshake shall continue according to the 1200 bit/s Handshake sequence for an answering modem. However, if unscrambled repetitive double dibit 00 and 11 at 1200 bit/s is detected in the low channel, at the end of receipt of this signal the modem shall be detected and then transmit an unscrambled repetitive double dibit pattern of 00 and 11 at 1200 bit/s for 100 ± 3ms. Following these signals the modem shall transmit scrambled binary 1's at 1200 bit/s.
- c)  $600 \pm 10$ ms after the detection of the end of double dibit 00 and 11 at 1200 bit/s the modem shall begin transmitting scrambled binary 1's at 2400 bit/s.  $450 \pm 10$ ms after detection the receiver may begin making 16-way decisions.
- d) Following transmission of scrambled binary 1's at 2400 bit/s for  $200 \pm 10$ ms, the modem shall be ready to transmit data at 2400 bit/s.
- e) When 32 consecutive scrambled binary 1's at 2400 bit/s have been detected in the low channel the modem shall be ready to receive data at 2400 bit/s.

Handshake sequence for calling modem (1200 bit/s operation)

The following handshake is identical to the Recommended V.22 handshake.

- a) On connection to the line the calling modem shall be conditioned to receive signals in the high channel at 1200 bit/s and transmit signals in the low channel at 1200 bit/s. The modem shall initially remain silent.
- b) After  $155 \pm 10$ ms of unscrambled binary 1's have been detected, the modem shall remain silent for a further  $456 \pm 10$ ms then transmit scrambled binary 1's at 1200 bit/s.
- C) On detection of scrambled binary 1's in the high channel at 1200 bit/s for  $270 \pm 40$ ms the modem shall be ready to receive data at 1200 bit/s.
- d)  $765 \pm 10$ ms after that the modem shall be ready to transmit data at 1200 bit/s.

Handshake sequence for answering modem (1200 bit/s operation)

The following handshake is identical to the Recommended V.22 handshake.

- a) On connection to the line the answering modem shall be conditioned to transmit signals in the high channel at 1200 bit/s and receive signals in the low channel at 1200 bit/s. Following transmission of the answer sequence in accordance with V.25 the modem shall transmit unscrambled binary 1's at 1200 bit/s.
- b) On detection of scrambled binary 1's in the low channel at 1200 bit/s for  $270 \pm 40$ ms

- the modem shall transmit scrambled binary 1 at 1200 bit/s.
- c) After scrambled binary 1's have been transmitted at 1200 bit/s for  $765 \pm 10$ ms the modem shall be ready to transmit and receive data at 1200 bit/s.

# 3.1.4.4 Retrain sequence (2400 bit/s operation)

A retrain sequence may be initiated during data transmission between two V.22bis modems if either modem incorporates a means of detecting loss of equalization.

Transmission of a retrain sequence shall be initiated either by detection a loss of equalization or by detection of unscrambled repetitive double dibit 00 and 11 at 1200 bit/s from the remote modem (S1 sequence).

The following sequence of events shall take place during the retrain:

- a) Following detection of a loss of equalization or the end of detection of unscrambled repetitive double dibit 00 and 11 at 1200 bit/s from the remote modem. The modem shall transmit an unscrambled repetitive double dibit pattern of 00 and 11 at 1200 bit/s for  $100 \pm 3$ ms. Following this signal the modem shall transmit scrambled binary 1's at 1200 bit/s.
- b)  $600 \pm 10$ ms after the end of detection of unscrambled repetitive double dibit 00 and 11 at 1200 bit/s from the remote modem, the modem shall begin transmitting scrambled binary 1's at 2400 bit/s.  $450 \pm 10$ ms after the end of this double dibit detection the receiver may begin making 16-way decisions.
- c) Following transmission of scrambled binary 1's at 2400 bit/s for  $200 \pm 10$ ms, the modem shall be ready to transmit data at 2400 bit/s.
- d) When 32 consecutive scrambled binary 1's at 2400 bit/s have been detected from the remote modem, the modem shall be ready to receive data at 2400 bit/s.

A retrain between two modems is shown in Figure 3-12.

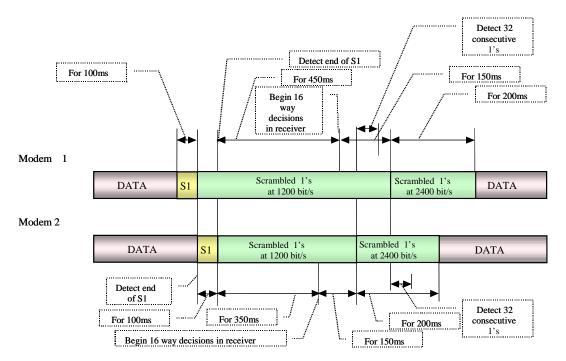


Figure 3-12. V.22bis retrain sequence

If a modem has transmitted a retrain signal and has not received unscrambled repetitive double dibit 00 and 11 at 1200 bit/s immediately prior, or during, or within a time interval equal to the maximum expected two-way propagation delay, the modem shall return to the beginning of the retrain signal as defined above and repeat the procedure until unscrambled repetitive double dibit 00 and 11 is received from the remote modem. A time interval of 1.2 seconds is recommended for the maximum expected two-way propagation delay.

If the modem fails to synchronize on the received retrain sequence, the modem shall transmit another retrain signal.

# 3.1.4.5 Implementation

The requirements of V.22bis are implemented in the v22bis.c file. This module is responsible for the initialization of the V.22bis handshake control structures  $(Rx(Tx)\_V22bis\_DH\_handshake\_t)$ , the initialization of the QAM control structures  $(Rx(Tx)\_control\_QAM\_t)$ , for details about these structures refer to Sections 3.1.7.4/3.1.7.5), scrambling, descrambling, and controlling the handshake process  $(Rx\_V22bis\_handshake\_routine)$ .

The description of the main functions and control structures of *v22bis.c* are given below.

#### 3.1.4.6 RX V.22bis handshake data handler structure

```
struct Rx_V22bis_DH_handshake_t
{
```

```
enum Rx V22bis hadshake state t state;
   bool retrain:
   uint32 det unscr ones ms;
   uint32 det scr ones 1200 ms;
   uint32 wait send ms;
  uint32 wait data 1200 ms;
   uint32 send s1 ms;
   uint32 det s1 ms;
  uint8 current s1;
  uint32 ones counter;
   uint32 wait adaptive equalizer ms;
   uint32 wait 16 way decision ms;
   uint32 wait scr ones 2400 ms;
  uint32 send scr ones 2400 ms;
   uint8 det scr ones 2400 bits;
   uint32 current counter;
   uint32 global counter;
};
```

- **state** Current state of the RX V22bis Handshake Data Handler. This field is initialized in function *Rx\_V22bis\_handshake\_init*.
- **retrain** Indicates the retrain mode. This field is initialized in function  $Rx\_V22bis\_handshake\_init$ .
- **det\_unscr\_ones\_ms** Duration of unscrambled binary 1's detection in ms. This field is initialized in function *Rx\_V22bis\_handshake\_init*.
- **det\_scr\_ones\_1200\_ms** Duration of the scrambled binary 1's on 1200 b/s detection in ms. This field is initialized in function *Rx\_V22bis\_handshake\_init*.
- wait\_send\_ms Specified time before starting to transmit scrambled binary 1's in ms. This field is initialized in function *Rx\_V22bis\_handshake\_init*.
- wait\_data\_1200\_ms Specified time before Carrier Detect Response in ms. This field is initialized in function *Rx V22bis handshake init*.

- **send\_s1\_ms** Duration of the S1 sequence. This field is initialized in function *Rx\_V22bis\_handshake\_init*.
- **det\_s1\_ms** Detection time of the S1 sequence. This field is initialized in function *Rx\_V22bis\_handshake\_init*.
- **current\_s1** Contains one or zero. Is used for S1 detection. This field is initialized in function *Rx\_V22bis\_handshake\_init* to 0.
- **ones\_counter** Counter used in the S1 detection state. This field is initialized in function  $Rx\_V22bis\_handshake\_init$  to 0.
- wait\_adaptive\_equalizer\_ms Time period after which the Adaptive equalizer is enabled. This field is initialized in function *Rx\_V22bis\_handshake\_init*.
- **training\_adaptive\_equalizer\_ms** Time period of the Adaptive equalizer training in ms. This field is initialized in function *Rx\_V22bis\_handshake\_init*.
- wait\_16\_way\_decision\_ms, wait\_scr\_ones\_2400\_ms, send\_scr\_ones\_2400\_ms, det\_scr\_ones\_2400\_bits Time intervals during the handshake process according to V.22bis recommendations. These fields are initialized in function Rx\_V22bis\_handshake\_init.
- **current\_counter** The counter that is used for counting the different data sequences (e.g. for detecting 32 consecutive 1's). This field is initialized in function *Rx\_V22bis\_handshake\_init* to 0.
- **global\_counter** Current counter of the bits received throughout the handshaking phase. This field is initialized in function *Rx\_V22bis\_handshake\_init* to 0.
- mode Specifies 1200 bit/s (V22BIS\_MODE\_1200) or 2400 bit/s (V22BIS\_MODE\_2400) mode. This field is initialized in function Rx\_V22bis\_handshake\_init.

#### 3.1.4.7 TX V.22bis handshake data handler structure

```
struct Tx_V22bis_DH_handshake_t
{
   enum Tx_V22bis_hadshake_state_t state;
   uint8 current_bit;
};
```

- **state** Current state of the Tx V22bis Handshake Data Handler. This field is initialized in function *Tx\_V22bis\_handshake\_init*.
- **current\_bit** Contains the current quadbit (dibit) to be transmitted. This field is initialized in function *Tx\_V22bis\_handshake\_init* to 1.

# V22bis\_scramble\_bit

## Call(s):

uint8 V22bis\_scramble\_bit(struct channel\_t \* channel, uint8 bit)

#### **Arguments:**

Table 3-23. V22bis\_scramble\_bit arguments

channel	in	Pointer to the channel control data structure
bit	in	Bit to scramble

**Description:** 

Scrambles input bits according to the V.22bis recommendation.

$$d_s(nT) = d(nT) \text{ XOR } d_s((n-14)T) \text{ XOR } d_s((n-17)T).$$

Scrambler register is stored in

channel->Tx\_control\_ptr->data\_pump\_ptr->scrambler\_register.

**Returns:** 

Scrambled bit.

```
uint8 temp_current_nbits, current_nbits;
struct channel_t * channel;
...
temp_current_nbits = V22bis_scramble_bit(channel, current_nbits);
...
```

# V22bis\_descramble\_bit

# Call(s):

uint8 V22bis descramble bit(struct channel t \* channel, uint8 bit)

#### **Arguments:**

Table 3-24. V22bis\_descramble\_bit arguments

channel	in	Pointer to the channel control data structure
bit	in	Bit to descramble

**Description:** 

Descrambles input bits according to the V.22bis recommendation.

$$d(nT) = d_s(nT) \text{ XOR } d_s((n-14)T) \text{ XOR } d_s((n-17)T).$$

Descrambler register is stored in

channel->Rx\_control\_ptr->data\_pump\_ptr->descrambler\_register.

**Returns:** 

Descrambled bit.

```
uint8 temp_current_nbits, current_nbits;
struct channel_t * channel;
...
temp_current_nbits = V22bis_descramble_bit(channel, current_nbits);
...
```

# Tx\_V22bis\_init

#### Call(s):

void Tx\_V22\_init (struct channel\_t \* channel,struct Tx\_control\_QAM\_t
\*Tx\_control\_QAM, bool calling,enum v22bis\_mode\_t mode)

### **Arguments:**

#### Table 3-25. Tx V22bis init arguments

channel	in	Pointer to the channel control data structure	
Tx_control_QAM	in	Pointer to the QAM transmitter control data structure	
calling	in	Indicates either modem in call or answer mode	
mode	in	Specifies 1200 bit/s (V22BIS_MODE_1200) or 2400 bit/s (V22BIS_MODE_2400) mode	

#### **Description:**

This function initializes the Transmitter control Structures pointed to by  $Tx\_control\_QAM$  and channel according to V.22bis and calls  $QAM\_modulator\_init()$ .

It fills the appropriate fields of the  $Tx\_control\_t$  structure, pointed to by  $channel->Tx\_control\_ptr$  (fields:  $n\_bits$ ,  $n\_bits\_mask$ ,  $baud\_rate$ ,  $number\_samples$ ,  $data\_pump\_call\_func$ , state,  $process\_count$ ) and the  $Tx\_control\_QAM$  structure (fields:  $f\_carrier$ ,  $f\_guard$ ,  $amplitude\_carrier$ ,  $amplitude\_guard$ ,  $constellation\_phase\_quadrant$ ,  $phase\_quadrant\_change$ , scrambler).

This function is called from *state\_machine()*.

**Returns:** None

```
struct channel_t * channel;
struct Tx_control_QAM_t *Tx_control_QAM;
...
Tx_V22bis_init (channel, Tx_control_QAM, TRUE, V22bis_MODE_2400);
...
```

# Rx\_V22bis\_init

#### Call(s):

void Rx\_V22bis\_init (struct channel\_t \* channel,struct Rx\_control\_QAM\_t
\*Rx control QAM, bool calling,enum v22bis mode t mode)

### **Arguments:**

#### Table 3-26. Rx\_V22bis\_init arguments

channel	in	Pointer to the channel control data structure
Rx_control_QAM	in	Pointer to the QAM receiver control data structure
calling	in	Indicates either modem in call or answer mode
mode	in	Specifies 1200 bit/s (V22BIS_MODE_1200) or 2400 bit/s (V22BIS_MODE_2400) mode

#### **Description:**

This function initializes the Receiver control Structure according to V.22*bis*.

It fills the appropriate fields of the  $Rx\_control\_t$  structure, pointed to by  $channel->Rx\_control\_ptr$  (fields:  $n\_bits$ ,  $n\_bits\_mask$ ,  $baud\_rate$ ,  $number\_samples$ ,  $data\_pump\_call\_func$ , state,  $process\_count$ ) and the  $Rx\_control\_QAM$  structure (fields:  $f\_carrier$ ,  $constellation\_phase\_quadrant$ ,  $constellation\_threshold$ ,  $noise\_threshold$ ,  $phase\_quadrant\_change$ , descrambler).

This function is called from *state\_machine()*.

**Returns:** None

```
struct channel_t * channel;
struct Rx_control_QAM_t *Rx_control_QAM;
...

Rx_V22bis_init (channel, Rx_control_QAM, TRUE, V22bis_MODE_1200);
...
```

# Rx\_V22bis\_handshake\_init

#### Call(s):

void Rx\_V22bis\_handshake\_init (struct channel\_t \* channel, struct
Rx V22bis DH handshake t \*Rx V22bis handshake, bool calling, bool retrain)

## **Arguments:**

Table 3-27. Rx\_V22bis\_handshake\_init arguments

channel	in	Pointer to the channel control data structure
Rx_V22bis_handshake	in	Pointer to RX handshake data handler structure
calling	in	Indicates either modem in call or answer mode
retrain	in	Indicates retrain mode
mode	in	Specifies 1200 bit/s (V22BIS_MODE_1200) or 2400 bit/s (V22BIS_MODE_2400) mode

#### **Description:**

This function initializes the receiver V.22bis handshake data handler control structure. It fills the appropriate fields of  $Rx\_control\_t$  structure, pointed to by  $channel->Rx\_control\_ptr$  (fields:  $data\_handler\_call\_func$ ,  $data\_handler\_state$ ,  $number\_n\_bits$ ) and the  $Rx\_V22bis\_handshake$  structure (fields: retrain, state,  $send\_sl\_ms$ ,  $det\_sl\_ms$ ,  $send\_scr\_ones\_2400\_ms$ ,  $det\_scr\_ones\_2400\_ms$ ,  $wait\_adaptive\_equalizer\_ms$ ,  $det\_scr\_ones\_1200\_ms$ ,  $det\_unscr\_ones\_ms$ ,  $wait\_send\_ms$ ,  $wait\_data\_1200\_ms$ ,  $wait\_16\_way\_decision\_ms$ ,  $wait\_scr\_ones\_2400\_ms$ ,  $current\_sl$ ,  $ones\_counter$ ,  $current\_counter$ ,  $global\_counter$ ). It sets up the data handler for the receiver to perform the  $Rx\_V22bis\_handshake\_routine$ .

This function is called from *state\_machine()*.

**Returns:** None

```
struct channel_t * channel;
struct Rx_V22bis_DH_handshake_t Rx_V22_DH_hshake;
...
Rx_V22bis_handshake_init(channel,&Rx_V22bis_DH_hshake,TRUE,FALSE);
...
```

# Rx\_V22bis\_handshake\_routine

### Call(s):

void Rx\_V22bis\_handshake\_routine(struct channel\_t \* channel)

#### **Arguments:**

Table 3-28. Rx\_V22bis\_handshake\_routine arguments

d			
	channel	in	Pointer to the channel control data structure

#### **Description:**

This is the RX V.22bis Handshake Data Handler routine. This function reads data (symbols) from the Rx\_data[] buffer. It controls the handshake process. The inner state diagram for this routine during handshake is shown in Fig 3.1.4.14. If handshake is completed successfully *Rx\_control->data\_handler\_state* is set to *RX\_DH\_STATE\_COMPLETED*, else to *RX\_DH\_STATE\_FAILED*.

This function is called by the  $Rx\_data\_handler()$  function via the  $channel->Rx\_control\_ptr-> data\_handler\_call\_func()$ .

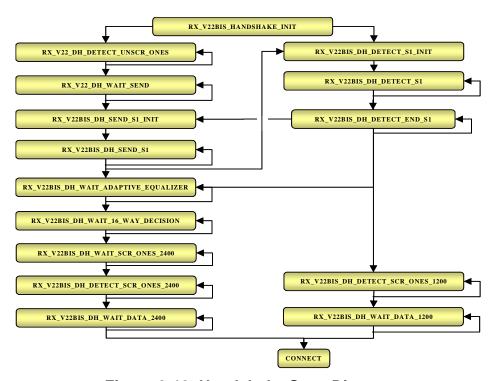


Figure 3-13. Handshake State Diagram

**Returns:** None

**Code example:** 

struct channel\_t \* channel;
...

Rx\_V22bis\_handshake\_routine (channel);

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# Tx\_V22bis\_handshake\_init

#### Call(s):

void Tx\_V22bis\_handshake\_init (struct channel\_t \* channel, struct
Tx V22bis DH handshake t \*Tx V22bis handshake)

## **Arguments:**

#### Table 3-29. Tx\_V22bis\_handshake\_init arguments

channel	in	Pointer to the channel control data structure
Tx_V22bis_handshake	in	Pointer to the TX V.22bis handshake data handler structure

#### **Description:**

This function initializes the transmitter V.22bis handshake data

handler control structure.

It fills the appropriate fields of *Tx\_control\_t* structure, pointed to by

*channel->Tx\_control\_ptr* (fields: *data\_handler\_ptr*,

data\_handler\_call\_func, data\_handler\_state, number\_n\_bits) and the *Tx\_V22bis\_handshake* structure (fields: *state*, *current\_bit*). It

sets up the data handler for the receiver to perform the

 $Tx_V22bis_handshake_routine.$ 

This function is called from *state\_machine()*.

**Returns:** None

```
struct channel_t * channel;
...
Rx_V22bis_handshake_init (channel);
```

# Tx\_V22bis\_handshake\_routine

# Call(s):

void Tx\_V22bis\_handshake\_routine(struct channel\_t \* channel)

#### **Arguments:**

Table 3-30. Tx\_V22bis\_handshake\_routine arguments

channel	in	Pointer to the channel control data structure
---------	----	---

**Description:** This is the TX V.22*bis* Handshake Data Handler routine. Writes

sequences of 1's or S1 into the TX\_data[] buffer depending on the

state of the TX handshake data handler.

This function is called by the  $Tx\_data\_handler()$  function via the

channel->Tx\_control\_ptr-> data\_handler\_call\_func ().

**Returns:** None

```
struct channel_t * channel;
...

Tx_V22bis_handshake_routine (channel);
...
```

# Tx\_V22bis\_change\_mode

## Call(s):

void Tx\_V22bis\_change\_mode(struct channel\_t \* channel,enum v22bis\_mode\_t mode)

#### **Arguments:**

#### Table 3-31. Tx\_V22bis\_change\_mode arguments

channel	in	Pointer to the channel control data structure
mode	in	Specifies the mode to switch to (either 2400 b/s or 1200 b/s).

# **Description:**

Changes the mode of the Tx V22bis Data Pump to 1200 bit/s or to 2400 bit/s by assigning new values to  $n\_bits$  (the number of bits in symbol),  $n\_bits\_mask$  (the mask for retrieving  $n\_bits$  from the byte) and  $constellation\_phase\_quadrant$  (the constellation points).

**Returns:** None

```
struct channel_t * channel;
...

Tx_V22bis_change_mode(channel, V22BIS_MODE_1200);
...
```

# Rx\_V22bis\_change\_mode

### Call(s):

void Rx V22bis change mode(struct channel t \* channel,enum v22bis mode t mode)

# **Arguments:**

# Table 3-32. Rx\_V22bis\_change\_mode arguments

channel	in	Pointer to the channel control data structure
mode	in	Specifies the mode to switch to (either 2400 b/s or 1200 b/s).

**Description:** Changes the mode of the Rx V22bis Data Pump to 1200 bit/s or to

2400 bit/s by assigning new values to  $n\_bits$  (the number of bits in symbol),  $n\_bits\_mask$  (the mask for retrieving  $n\_bits$  from the byte)

and *constellation\_phase\_quadrant* (the constellation points).

**Returns:** None

```
struct channel_t * channel;
...

Rx_V22bis_change_mode(channel, V22BIS_MODE_1200);
```

# 3.1.5 FSK

The FSK (Frequency Shift Keying) module is used by the V.21 and V.23 modems. This module consists of two blocks: a FSK transmitter and receiver. In the following subsections, the operation of the FSK modem transmitter and receiver are described.

Binary FSK (usually referred to simply as FSK) is a modulation scheme used to send digital information between digital equipment. The data is transmitted by shifting the frequency of a continuous carrier in a binary manner to one or the other of two discrete frequencies. One frequency is designated as the "mark" frequency and the other as the "space" frequency. The mark and space correspond to binary 1 and 0 respectively.

#### 3.1.5.1 FSK Transmitter

The transmitter takes data from the Tx\_data[] buffer, performs FSK modulation taking into account the bits for transmitting, and puts the resultant samples into the Tx\_sample[] buffer (see Figure 3-14).

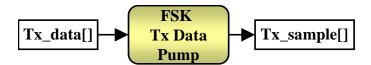


Figure 3-14. FSK Tx Data Pump

The FSK modulation is represented in the following manner:

$$S(n) = A * cos((\omega_c \pm \Delta \omega)*n + \phi)$$

#### Where

- S(n) = Transmitted signal
- A = Amplitude of transmitted signal
- $\omega_c$  = Carrier (mean) frequency
- $\Delta \omega$  = Frequency deviation
- n = Number of a sample
- $\phi$  = Phase shift

### For V.21:

- $\omega_c = 1080$  Hz (channel No.1); 1750 Hz (channel No.2).
- $\Delta \omega = 100 Hz$

#### For V.23:

•  $\omega_c$  = 1500 Hz (channel 600 baud); 1700 Hz (channel 1200 baud); 420 Hz (channel 75 bauds)

•  $\Delta\omega$  = 200 Hz (channel 600 baud); 600 Hz (channel 1200 baud); 30 Hz (channel 75 baud)

The higher characteristic frequency corresponds to the Space frequency (binary 0), the second one corresponds to the Mark frequency (binary 1).

The structure of the FSK Transmitter (modulator) is shown in Figure 3-15.

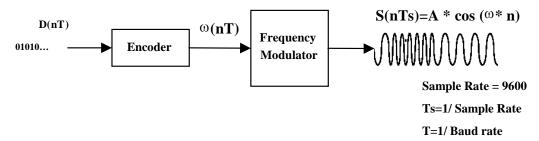


Figure 3-15. FSK Transmitter structure

The encoder takes binaries for transmiting from the Tx\_data[] buffer. Depending on the binary the encoder changes the current phase step (shift) this means that it changes the current frequency (Mark or Space) of the Frequency modulator. The Frequency modulator generates the current frequency tone. Some of the generated harmonics fall into the frequency region reserved for the receiver. To eliminate these harmonics, the modulator output must be digitally filtered. For this purpose a 38-tap bandpass filter (BPF) is used. The generated result samples are placed into the Tx\_sample[] buffer.

The FSK Tx Data Pump is implemented in the *FSK\_modulator()* routine.

#### 3.1.5.2 FSK Receiver

The receiver takes samples from the Rx\_sample[] buffer, demodulates them and puts the resultant bits into the Rx\_data[] circular buffer (see Figure 3-16).

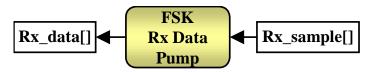


Figure 3-16. FSK Rx Data Pump

The structure of the FSK Receiver (demodulator) is shown in Figure 3-17.

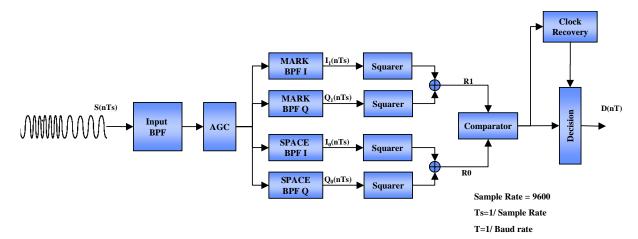


Figure 3-17. FSK Demodulator structure

The received signal from the Rx\_sample[] buffer is filtered by the Input BPF in order to reject unwanted out-of-band noise components. Then the filtered signal goes to the AGC (Automatic Gain Control) block. This adaptively maintains the output signal at a constant level that is necessary for proper operation of the algorithms of the modem receiver.

The algorithm of the AGC that is used in the current implementation is described below.

The AGC monitors the signal and calculates the gain correction factor. The input signal is multiplied by this gain correction factor so that the signal maximum remains within a certain range. The gain correction factor is calculated once every three bauds by a two-step process. First, the three maximum values of the signal, each one corresponding to one baud, are monitored and added to each other. The previous value of the average signal level is then added to this sum and divided by four to obtain the new average signal level. At the second step, the gain correction factor is calculated. The gain correction factor is the result of dividing the maximum-allowed signal level by the average signal level.

Also the AGC block determines if a carrier is present in the received signal by comparing the average signal level with the noise threshold.

After AGC the signal goes to the actual FSK demodulator. This type of FSK demodulator uses two types of filters, one is centered at the Mark frequency and the other at the Space frequency.

The Input BPF is centered about a mean frequency, Mark and Space BPFs are centered about the desired mark and space frequency. A simplified spectrum for the demodulator is shown in Fig 3.1.5.2.3.

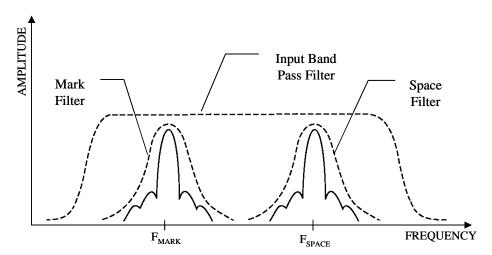


Figure 3-18. Simplified demodulation spectrum of a binary FSK signal

Also the Mark and Space filters are designed to make the signal complex. The complex signal has two components, a real (I channel) and an imaginary part (Q channel). The outputs of these filters are exactly 90 degrees out of phase with respect to each other.

The outputs of the Mark and Space filters are squared, then I and Q energies are summerised:

$$R1 = (I_{1})^{2} + (Q_{1})^{2}$$

$$R0 = (I_{0})^{2} + (Q_{0})^{2}$$

The powers of amplitudes of the two filter outputs (R1 and R0) are compared to determine whether the signal instantaneously falls mostly in the vicinity of the Mark(R1>R0) or is closer to the space (R0>R1). The resultant demodulated bits are then placed into the Rx\_data[] buffer.

The FSK Rx Data Pump is implemented in the *FSK\_demodulator()* routine.

The FSK module routines can be found in the "fsk.c" file.

The FSK Tx and Rx control data structures are defined in "fsk.h".

#### The Tx FSK control data structure:

```
uint16 phase;
uint16 phase_incr[2];
uint16 baud_frac;
uint16 baud_incr;

uint8 current_bit;
int16 amplitude;
};
```

The Tx\_control\_FSK\_t structure parameter descriptions:

- **f\_space** Space frequency, in Hz. It is used to represent binary 0. It is initialized in  $Tx_V21_{init}()$  or  $Tx_V23_{init}()$ .
- **f\_mark** Mark frequency, in Hz. It is used to represent binary 1. It is initialized in  $Tx_V21_{init}()$  or  $Tx_V23_{init}()$ .
- **phase** Current phase of the frequency generator, in Q15 format. It is initialized in *FSK\_modulator\_init()* to 0.
- **phase\_incr** Frequency phase increment per sample, in Q15 format. phase\_incr[0] is used for Space frequency and phase\_incr[1] is used for Mark frequency. It is initialized in *FSK\_modulator\_init()*.
- **baud\_frac** Fractional part of the baud, in Q15 format. It is initialized in *FSK modulator init()* to 0.
- **baud\_incr** Increment to <baud\_frac> per sample, in Q15 format. It is initialized in *FSK\_modulator\_init()*.
- **current\_bit** Current bit of data for modulation. It is initialized in *FSK\_modulator\_init()* to 0.
- **amplitude** Amplitude of the output signal, in Q14. It is initialized in  $Tx_V21_init$  () or  $Tx_V23_init$ ().
- **filter\_size** The number of coefficients used by the BPF. This field is initialized in *FSK\_modulator\_init()*.
- **filter\_coef** Coefficients of the BPF. This field is initialized in *QAM\_modulator\_init()*.
- **filter\_buf** Cyclic buffer used by the filter algorithm of the BPF. This field is initialized in *FSK\_modulator\_init()* to 0.
- **filter\_buf\_ptr** Pointer to the current element of the filter\_buf[]. This field is initialized in *FSK\_modulator\_init()*.

#### The Rx FSK control data structure:

```
/***************************
* FSK Receiver control data sructure
************************
typedef struct Rx control FSK t
   uint32 f space;
   uint32 f mark;
   int16 * lpf coef;
   uint32 filter_size;
   int16 filter space i[FSK FILTER COEF NUMBER];
   int16 filter_space_q[FSK_FILTER_COEF_NUMBER];
   int16 filter mark i[FSK FILTER COEF NUMBER];
   int16 filter mark q[FSK FILTER COEF NUMBER];
   int16 filter buf[FSK FILTER BUF SIZE];
   uint32 filter buf ptr;
   uint16 baud incr;
   uint16 baud frac;
   uint8 last bit;
   uint32 filter bpf size;
   int16 filter bpf coef[FSK BPF COEF NUMBER];
   int16 filter bpf buf[FSK BPF BUF SIZE];
   uint32 filter bpf buf ptr;
   uint32 agc;
   uint32 agc average;
   int16 agc cnt;
   int16 signal max;
   int16 signal present;
};
```

The Rx\_control\_FSK\_t structure Member Descriptions:

• **f\_space** - Space frequency, in Hz. It is used to represent binary 0. It is initialized in  $Rx_V21_init()$  or  $Rx_V23_init()$ .

- **f\_mark** Mark frequency, in Hz. It is used for to represent binary 1. It is initialized in  $Rx_{V21\_init}()$  or  $Rx_{V23\_init}()$ .
- **lpf\_coef** Pointer to coefficients of the FIR LPF(low-pass filter). It is used for calculation of coefficients for Space and Mark BPF filters. It is initialized in *Rx\_V21\_init()* or *Rx\_V23\_init()*.
- **filter\_size** The number of coefficients used by the Space and Mark BPF filters. It is initialized in *Rx V21 init()* or *Rx V23 init()*.
- **filter\_space\_i** Array of I channel coefficients of the Space BPF filter. It is initialized in *FSK\_demodulator\_init()*.
- **filter\_space\_q** Array of Q channel coefficients of the Space BPF filter. It is initialized in *FSK\_demodulator\_init()*.
- **filter\_mark\_i** Array of I channel coefficients of the Mark BPF filter. It is initialized in *FSK demodulator init()*.
- **filter\_mark\_q** Array of Q channel coefficients of the Mark BPF filter. It is initialized in *FSK\_demodulator\_init()*.
- **filter\_buf** Cyclic buffer used by the Mark and Space BPF filters. It contains the samples passed through the input BPF and AGC blocks. It is initialized in *FSK\_demodulator\_init()* to 0's.
- **filter\_buf\_ptr** Relative pointer to the current element in *filter\_buf[]*. It is initialized in *FSK\_demodulator\_init()*.
- **baud\_incr** Increment to <baud\_frac> per sample, in Q15 format. It is initialized in *FSK\_demodulator\_init()*.
- **baud\_frac** Fractional part of the baud, in Q15 format. It is initialized in *FSK\_demodulator\_init()* to 0.
- **last\_bit** Last demodulated bit of data (0 or 1). It is initialized in *FSK\_demodulator\_init()* to 0.
- **filter\_bpf\_size** The number of coefficients used by the Input BPF filter. It is initialized in *FSK\_demodulator\_init()*.
- **filter\_bpf\_coef** Array of coefficients of the Input BPF filter. It is initialized in *FSK\_demodulator\_init()*.
- **filter\_bpf\_buf** Cyclic buffer used by the Input BPF filter. It contains the samples received from the Rx\_sample[] buffer. It is initialized in *FSK\_demodulator\_init()* to 0's.
- **filter\_bpf\_buf\_ptr** Relative pointer to the current element in *filter\_bpf\_buf[]*. It is initialized in *FSK\_demodulator\_init()*.
- **agc** The AGC gain correction factor, in Q14 format. It is initialized in *FSK\_demodulator\_init()* to 1 in Q14 format (*COS\_BASE*).

#### **Data Pump**

- **agc\_average** The current average signal level in Q14 format. It is initialized in *FSK\_demodulator\_init()* to 0.
- **agc\_cnt** Counter of bauds left for calculation of the new gain correction factor. It is initialized in *FSK\_demodulator\_init()* to 3.
- **signal\_max** The maximum value of the input signal during a baud in Q14 format. It is initialized in *FSK\_demodulator\_init()* to 0.
- **signal\_present** Indicates the carrier signal presence. It is equal to TRUE if the signal is present and to FALSE otherwise. It is initialized in *FSK\_demodulator\_init()* to FALSE.

# FSK\_modulator\_init

## Call(s):

void FSK modulator init(struct channel t \* channel);

#### **Arguments:**

#### Table 3-33. FSK\_modulator\_init arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

# **Description:**

This function sets the fields of the Tx FSK control data structure (*phase\_incr, baud\_incr, phase, baud\_frac, current\_bit*), pointed to by the *channel->Tx\_control\_ptr->data\_pump\_pt*r to their default values..

This function is called by the Tx\_V21\_init() and Tx\_V23\_init() functions.

**Returns:** None.

```
struct channel_t channel;
...
FSK_modulator_init(&channel);
...
```

# FSK\_modulator

## Call(s):

void FSK modulator(struct channel t \* channel);

#### **Arguments:**

## Table 3-34. FSK\_modulator arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

# **Description:**

This function gets bits of data from the Tx\_data[] buffer, processes them performing binary FSK modulation (see 3.1.5.1.), and places the resultant samples into the Tx\_sample[] buffer. The number of

generated samples, per call, is defined by *channel->Tx\_control\_ptr->number\_samples*.

This function is called by the  $Tx\_data\_pump()$  function via the

channel->Tx\_control\_ptr-> data\_pump\_call\_func ().

#### **Returns:**

None.

```
struct channel_t channel;
...
FSK_modulator (&channel);
...
```

# FSK\_demodulator\_init

## Call(s):

void FSK demodulator init(struct channel t \* channel);

### **Arguments:**

#### Table 3-35. FSK\_demodulator\_init arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This function sets the fields of the Rx FSK control data structure (baud\_incr, baud\_frac, filter\_buf, filter\_buf\_ptr, last\_bit, filter\_space\_i, filter\_space\_q, filter\_mark\_i, filter\_mark\_q, filter\_bpf\_size, filter\_bpf\_buf, filter\_bpf\_buf\_ptr, filter\_bpf\_coef, agc, agc\_average, agc\_cnt, signal\_max, signal\_present), pointed to by channel->Rx\_control\_ptr->data\_pump\_ptr to their default values..

This function is called by the Rx\_V21\_init() and Rx\_V23\_init() functions.

**Returns:** 

None.

```
struct channel_t channel;
...
FSK_demodulator_init(&channel);
...
```

# FSK\_demodulator

# Call(s):

void FSK demodulator(struct channel t \* channel);

#### **Arguments:**

# Table 3-36. FSK\_demodulator arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This function gets samples from the Rx\_sample[] buffer, processes

them performing binary FSK demodulation (see 3.1.5.2.), and places

the resultant bits into the Rx\_data[] buffer.

This function is called by the  $Rx\_data\_pump()$  function via the

channel->Rx\_control\_ptr-> data\_pump\_call\_func ().

**Returns:** None.

```
struct channel_t channel;
...
FSK_demodulator (&channel);
...
```

# 3.1.6 **DPSK**

The Differential Phase Shift Keying (DPSK) modulation technique is used in the V.22 modem. This technique is implemented in the module *dpsk.c*. This module consists of two main blocks: the DPSK transmitter and DPSK receiver. In the following subsections, the operation of the modem transmitter and receiver are described.

The transmitter accepts data (bits) from the Data Terminal Equipment (DTE). It then performs the necessary processing in order to place this data into the proper form for transmission through the Public Switched Telephone Network. This process basically consists of the modulation of the baseband information (logical 1's and 0's sent by the DTE) into the passband of the communications channel for transmission. The receiver collects the information from the telephone network and converts it back into its original form, i.e. the bits sent by the DTE.

The *dpsk.c* module consists of two main blocks: the DPSK transmitter and the DPSK receiver. These blocks are called the data pump. They are executed every time the *Tx\_sample* buffer is almost empty (the modulator or TX data pump is executed) or when *Rx\_sample* is full enough (the demodulator or RX data pump is executed). They are called from the *modem.c* module.

#### 3.1.6.1 DPSK transmitter

In Differential Phase Shift Keying, the information is encoded as the phase change of the transmitter carrier. With  $\phi(n)$  denoting the phase that contains the information to be transmitted, the transmitted signal s(n) is represented mathematically by:

$$s(n) = A cos(\omega n + \phi(n))(1)$$

Where  $\omega$  is the carrier frequency. The parameter A determines the amplitude of the transmitted signal. It can also be written as:

$$s(n) = A (\cos(\omega n) \cos(\phi(n)) - \sin(\omega n) \sin(\phi(n)))(2)$$

The substitution of

$$I(n) = A \cos(\phi(n))$$

$$Q(n) = -A \sin(\phi(n))$$

into (2) results in (3) used to describe DPSK modulation systems:

$$s(n) = I(n) cos(\omega n) + Q(n) sin(\omega n)(3)$$

Each value of the  $\{I(n),Q(n)\}$  sequence corresponds to one signaling element (symbol) transmitted. The number of signaling elements transmitted per second is commonly referred to as the baud rate, which for V.22 is set by the protocol to 600. The set of possible values of the sequence  $\{I(n),Q(n)\}$  determines the signal constellation, which is given in two-dimensional representation. The signal constellation, commonly referred to as the

constellation diagram, is a geometric picture that emphasizes the fact that the two channels are 90 degrees (Quadrature) out-of-phase. The V.22, with 600 baud rate, accomplishes the transmission of 1200 bps by encoding two incoming bits (dibit) in a single baud. Since there are four possible values for every dibit, the constellation diagram for V.22 contains four points.

The overall block diagram for the DPSK transmitter is shown in Figure 3-19. The basic structural blocks are the scrambler, encoder, digital modulator and digital filter.

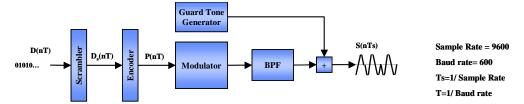


Figure 3-19. DPSK transmitter block diagram

#### 3.1.6.1.1 Scrambler

The purpose of the scrambler is to randomize the input binary data sequence, which means that it converts this sequence in a pseudo-random binary sequence with some defined period. This is required for the clock recovery module. For more detailed information about the scrambler refer to section 3.1.3.2.

#### 3.1.6.1.2 Encoder

The encoder implements encoding of the incoming sequence  $d_s(n)$  into the appropriate phase changes i.e. its function is to map every two incoming bits of the incoming sequence  $d_s(n)$  to a total phase. The total phase is then transmitted. The mapping rule can be seen in section 3.1.3.1.

In the current implementation of the encoder, it changes the carrier phase according to the incoming dibit and is executed once per baud (i.e. with sample rate = 9600 the phase changes once per 16 samples).

#### 3.1.6.1.3 Modulator

The modulator modulates the cosine wave (which contains the phase change). The modulating frequency for V.22 is 1200 Hz for a calling modem and 2400 Hz for an answering modem.

### 3.1.6.1.4 Bandpass filter

After modulation, the signal cannot be directly transmitted through the telephone line. The reason is that the instantaneous changes of  $I(nT_b)$  and  $Q(nT_b)$  generate higher-order harmonics. Some of these harmonics fall into the frequency region reserved for the receiver.

To eliminate these harmonics, the modulator output must be digitally filtered. For this purpose a 38-tap bandpass filter (BPF) is used. This filter is based on a lowpass (LPF) FIR filter and its coefficients are calculated according to the following rule:

$$h_b(nT_s) = 2 h_l(nT_s) \cos(n\omega T_s)$$

Where  $T_s$  is the sampling period and  $h_l(nT_s)$  are the coefficients of the LPF. The coefficients of the BPF are calculated in the initialization routine (*DPSK\_modulator\_init*) and depend on the carrier frequency.

## 3.1.6.1.5 Guard tone generator

This module generates the guard tone, which is then added to the main signal. For V.22 the frequency of the guard tone is 1800 Hz  $\pm$  20 Hz or 550  $\pm$  20 Hz, and is only transmitted when the modem is transmitting in the high channel (carrier frequency is 2400 Hz).

#### 3.1.6.2 DPSK receiver

This subsection describes the functional blocks required to implement a DPSK receiver. The receiver structure is more sophisticated than that of the transmitter. An overall diagram of the modem receiver is shown in Fig 3.1.6.3. The basic structural blocks of the modem receiver are the input bandpass filters, the automatic gain control (AGC), the demodulator, the decision block, the decoder, the descrambler, the carrier recovery, and the clock recovery.

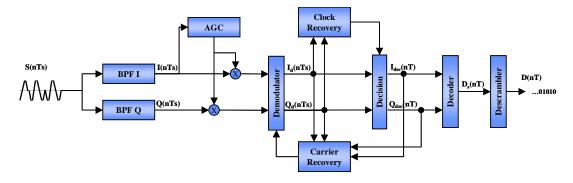


Figure 3-20. DPSK modem receiver block diagram

## 3.1.6.2.1 Bandpass filters

The input signal is bandpass filtered for the following reasons:

- Rejection of out-of-band noise, including the rejection of the transmit signal spectrum
- Introduction of a 90-degree relative phase shift required for I and Q channel separation

• Fixed equalization for line distortion

In the current implementation a 38-tap bandpass FIR filter is used.

## 3.1.6.2.2 Automatic Gain Control (AGC)

The incoming signal level may vary over a large range due to attenuations in the telephone line. However for clock and carrier recovery modules the signal level must be independent of the attenuation introduced by the communications channel and remain constant. Automatic Gain Control performs this task. It adjusts the envelope of the I and Q channels so that they are of the same magnitude. The algorithm of the AGC that is used in the current implementation is described below.

The AGC monitors the I channel of the receiver and calculates the gain correction factor. The gain correction factor is calculated once every three baud by a two-step process. First, the three maximum values of the signal, each one corresponding to one baud (16 samples), are monitored and added to each other. The previous value of the average signal level is then added to this sum and divided by four to obtain the new average signal level.

In the second step, the gain correction factor is calculated. The gain correction factor is the result of dividing the maximum-allowed signal level by the average signal level. The AGC block also determines if a carrier is present in the received signal by comparing the average signal level with the threshold.

#### 3.1.6.2.3 Demodulator

With  $I_p(nT_s)$  and  $Q_p(nT_s)$  as inputs to the demodulator, the outputs  $I'(nT_s)$  and  $Q'(nT_s)$  are given by:

$$\begin{split} I'(nT_s) &= I_p(nT_s) \cos(\phi(nT_{s)}) + Q_p(nT_s) \sin(\phi(nT_s)) \\ Q'(nT_s) &= -I_p(nT_s) \sin(\phi(nT_s)) + Q_p(nT_s) \cos(\phi(nT_s)) \end{split}$$

Where  $\phi$  is the local carrier phase.

#### 3.1.6.2.4 Decision block and Decoder

The decision block calculates the current phase from the values of the baseband I and Q. The phase change is then calculated by subtracting the previous value of the phase from the current value of the phase.

The decoder then finds the dibit that corresponds to the last phase change (according to table 3.1.3).

#### 3.1.6.2.5 Descrambler

The descrambler is intended to recover the originally transmitted dibit. For more detailed information about the descrambler refer to section 3.1.3.3.

## 3.1.6.2.6 Clock Recovery

The purpose of Clock Recovery is to detect the middle of a baud. The decision block makes decisions exactly in the middle of the baud. The energy of the incoming signal at this point is maximal, so the probability of a mistake during decision making is minimal.

In the current implementation the early-late method of Clock Recovery is used. Figure 3-21 shows the energy values for each sample over the baud. It can be seen in this figure that if the energy sample E7 is in the middle of the baud, it has the highest energy value and the rest of the samples are located symmetrically around it, i.e. E1=E13, E2=E12, E3=E11 and so on. In other words E1-E13 = 0, E3-E11 = 0.

If E7 is shifted left and is not on top of the energy 'hill', then E11 becomes greater than E3 or E3 – E11 < 0. If E7 is shifted right, E3 – E11 > 0. So the difference (E3-E11) may be used as an error value. If this value is negative, it means that the middle of the local band occurred earlier than the middle of the incoming band. Therefore, the local band clock must be delayed. If positive – the local band clock must be advanced.

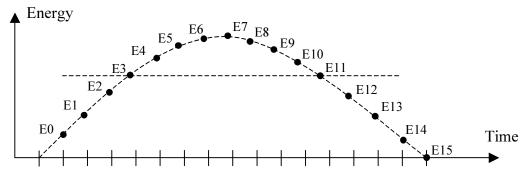


Figure 3-21. Signal energies over a baud

In the current implementation the energy of the third and eleventh samples are calculated. The difference between these energies is then found. This difference is then lowpass filtered by a first-order IIR filter. The output of the filter is used for local baud clock adjusting. The block diagram of the clock recovery algorithm is shown in Figure 3-22.

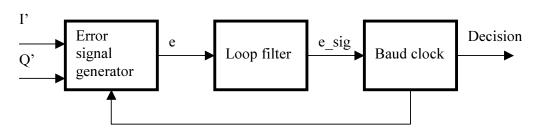


Figure 3-22. Clock Recovery block diagram

## 3.1.6.2.7 Carrier Recovery

Carrier Recovery is used for adjusting the phase of the local carrier to match the phase of the incoming carrier. It is important to generate a local carrier that has the same phase and frequency as the incoming carrier because it is used in the demodulator to demodulate the incoming signal and retrieve the baseband information.

To implement Carrier Recovery a phase-locked loop is used (Figure 3-23).

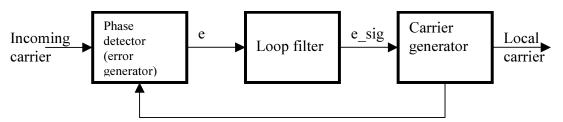


Figure 3-23. Carrier Recovery block diagram

The phase detector generates an error that is used to synchronize the local carrier to the incoming carrier. This error signal contains the information about the phase and frequency difference between the local and the incoming carriers.

The constellation points that were chosen in the current implementation are shown in Figure 3-24. The output of the phase detector is of the form:

$$E(nT_b) = sgn(Q'(nT_b)) I'(nT_b) - sgn(I'(nT_b)) Q'(nT_b)$$

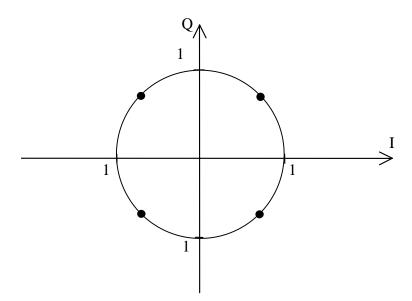


Figure 3-24. DPSK receiver decision points

The error  $E(nT_b)$  is a geometrical distance from the point used to make the decision and the line that intersects the points (0,0) and the optimum decision point.

The optimum decision points are the ideal points for a decision to be made (the black points in Fig.3.1.6.3.7.2). If the local carrier does not match the remote carrier, the actual received points are rotated from the ideal points by some angle.

Now, draw a line through the two points: the optimum or ideal point and the origin (0,0). The error E is the distance from this line to the point that is used to make the decision.

Therefore, E shows how far out our new point (the point which is used to make decision) is from its ideal position (the optimum point).

This error is then lowpass filtered by the *loop filter* (first-order IIR filter) and is used to adjust the current value of the local carrier phase (either delay or advance the phase of the local carrier). So, as a result the Clock Recovery block 'rotates' the constellation diagram to match the one shown in Fig 3.1.6.3.7.2. The filter coefficients and thresholds change depending on the state of the receiver (handshake mode or data mode).

The phase correction is made once per baud and is performed in the middle of the baud.

### 3.1.6.3 DPSK modulator control structure

```
typedef struct Tx_control_DPSK_t
{
   uint32 f_carrier;
   uint32 f_guard;
```

#### **Data Pump**

```
uint16 carrier phase;
   uint16 guard phase;
   uint16 carrier phase incr;
   uint16 guard phase incr;
   uint16 baud frac;
   uint16 baud incr;
   uint16 * omega ptr;
   uint8 current nbits;
   int16 amplitude carrier;
   int16 amplitude guard;
   scrambler call func t scrambler;
   bool scrambler enabled;
   uint32 scrambler register;
   uint32 filter size;
   int16 filter coef[DPSK FIXED EQ COEF NUMBER];
   int16 filter buf[DPSK FILTER BUF SIZE];
   uint32 filter_buf ptr;
};
```

- **f\_carrier** Carrier frequency in Hz. This field is initialized in function *Tx\_V22\_init*.
- **f\_guard** Guard tone frequency in Hz. This field is initialized in function *Tx\_V22\_init*.
- **carrier\_phase** Current phase of the carrier frequency generator in Q15 format. This field is initialized in function *DPSK\_modulator\_init* to 0.
- **guard\_phase** Current phase of the guard tone generator in Q15 format. This field is initialized in function *DPSK\_modulator\_init* to 0.
- **carrier\_phase\_incr** Carrier frequency phase increment per sample in Q15 format. This field is initialized in function *DPSK\_modulator\_init*.

- **guard\_phase\_incr** Guard tone frequency phase increment per sample in Q15 format. This field is initialized in function *DPSK\_modulator\_init*.
- **baud\_frac** Fractional part of the baud in Q14 format. (i.e. 0.5 = the middle of the baud). This field is initialized in function *DPSK\_modulator\_init* to 0.
- **baud\_incr** Increment to <baud\_frac> per sample in Q15 format. This field is initialized in function *DPSK\_modulator\_init*.
- **omega\_ptr** Pointer to Nbits-to-Phase Change Correspondence table. This field is initialized in function *Tx\_V22\_init*.
- **current\_nbits** N-bits of data for modulation. This field is initialized in function *DPSK\_modulator\_init* to 0.
- **amplitude\_carrier** Amplitude of the output signal. This field is initialized in function  $Tx_V22_{init}$ .
- **amplitude\_guard** Amplitude of the Guard tone signal. This field is initialized in function *Tx\_V22\_init*.
- **scrambler** Pointer to the scrambler function. This field is initialized in function *Tx\_V22\_init*.
- **scrambler\_enabled** Indicates if the scrambler is enabled. This field is initialized in function *DPSK\_modulator\_init* to *TRUE*.
- **scrambler\_register** Shift Register used by the scrambler. This field is initialized in function *DPSK\_modulator\_init* to 0.
- **filter\_size** The number of coefficients used by the BPF. This field is initialized in function *DPSK\_modulator\_init*.
- **filter\_coef** Coefficients of the BPF. This field is initialized in function *DPSK\_modulator\_init*.
- **filter\_buf** Cyclic buffer used by the filter algorithm of the input BPFs. This field is initialized in function *DPSK\_modulator\_init* to 0.
- **filter\_buf\_ptr** Pointer to the current element of the filter\_buf[]. This field is initialized in function *DPSK\_modulator\_init*.

### 3.1.6.4 DPSK demodulator control structure

```
typedef struct Rx_control_DPSK_t
{
   uint32 f_carrier
   uint16 carrier_phase;
   uint16 carrier_phase_incr;

uint16 baud_frac;
```

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#### **Data Pump**

```
uint16 baud incr;
uint16 * omega_ptr;
uint16 total phase;
uint32 filter size;
int16 filter_mark_i[DPSK_FIXED_EQ_COEF_NUMBER];
int16 filter_mark_q[DPSK_FIXED_EQ_COEF_NUMBER];
int16 filter_buf[DPSK_FILTER_BUF_SIZE];
uint32 filter_buf_ptr;
uint16 arbitrary phase;
bool decision enabled;
scrambler call func t descrambler;
bool descrambler enabled;
uint32 descrambler register;
int16 noise threshold;
int16 agc;
uint32 agcave;
int16 avecnt;
int16 smax;
int16 carrier_locked;
int16 PLL1;
int16 CPLL1;
int32 cerror_sig;
```

```
int32 clerror_sig;
int32 clerror2_sig;
uint32 baud_nrg3;
uint32 baud_nrg7;
int16 signal_present;
int16 clock_corrected;
};
```

- **f\_carrier** Carrier frequency in Hz. This field is initialized in function *Rx\_V22\_init*.
- **carrier\_phase** Current phase of the carrier frequency generator in Q15 format. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **carrier\_phase\_incr** Carrier frequency phase increment per sample in Q15 format. This field is initialized in function *DPSK\_demodulator\_init*.
- **baud\_frac** Fractional part of the baud in Q15 format. (i.e. 0.5 = the middle of the baud). This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **baud\_incr** Increment to <baud\_frac> per sample in Q15 format. This field is initialized in function *DPSK\_demodulator\_init*.
- **omega\_ptr** Pointer to Nbits-to-Phase Change Correspondence table. This field is initialized in function *Rx V22 init*.
- **total\_phase** The phase transmitted during the previous baud interval in Q15 format. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **filter\_size** The number of coefficients used by the input BPF. This field is initialized in function *DPSK\_demodulator\_init*.
- **filter\_mark\_I** Coefficients of the BPF for I channel in Q14 format. This field is initialized in function *DPSK\_demodulator\_init*.
- **filter\_mark\_q** Coefficients of the BPF for Q channel in Q14 format. This field is initialized in function *DPSK\_demodulator\_init*.
- **filter\_buf** Cyclic buffer used by the filter algorithm of the input BPFs. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **filter\_buf\_ptr** Pointer to the current element of the filter\_buf[]. This field is initialized in function *DPSK\_demodulator\_init*.
- **decision\_enabled** Indicates if the decision block is enabled. This field is initialized in function *DPSK\_demodulator\_init* to *TRUE*.
- **descrambler** Pointer to the descrambler function. This field is initialized in function  $Rx\_V22\_init$ .
- **descrambler\_enabled** Indicates if the descrambler is enabled. This field is initialized in function *DPSK\_demodulator\_init* to *TRUE*.

- **descrambler\_register** Shift Register used by the descrambler. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **noise\_threshold** Threshold for signal/noise detection in Q14 format. This field is initialized in function *Rx\_V22\_init*.
- **agc** The AGC gain factor in Q11 format. This field is initialized in function *DPSK\_demodulator\_init* to 1.
- **agcave** The current average signal level in Q14 format. This field is initialized in function *DPSK\_demodulator\_init* to 0.0625.
- **avecnt** The counter of the baud in the AGC block. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **smax** The maximum I value over the baud in Q14 format. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **carrier\_locked** Indicates that the carrier is locked and the receiver is in data mode. This field is initialized in function *DPSK\_demodulator\_init* to *FALSE*.
- **PLL1** Filter coefficient for the Carrier Recovery loop filter. This field is initialized in function *DPSK\_demodulator\_init*.
- **CPLL1** Filter coefficient for the Clock Recovery loop filter. This field is initialized in function *DPSK\_demodulator\_init*.
- **cerror\_sig** Filtered error signal for Carrier Recovery and the state of the loop filter. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **clerror\_sig** Filtered error signal for Clock Recovery and the state of the loop filter. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **clerror2\_sig** Indicates the false lock of baud clock for Clock Recovery. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **baud\_nrg3** Energy value of the 3rd sample. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **baud\_nrg7** Energy value of the 7th sample. This field is initialized in function *DPSK\_demodulator\_init* to 0.
- **signal\_present** Indicates the carrier frequency signal presence. This field is initialized in function *DPSK\_demodulator\_init* to *FALSE*.
- **clock\_corrected** Indicates that the Clock was corrected during the current baud. This field is initialized in function *DPSK\_demodulator\_init* to *FALSE*.

# **DPSK\_modulator\_init**

## Call(s):

void DPSK modulator init(struct channel t \* channel)

#### **Arguments:**

## Table 3-37. DPSK\_modulator\_init arguments

channel	in	Pointer to the channel control data structure
---------	----	---

**Description:** Initialization of the DPSK transmitter control structure

(Tx\_control\_DPSK\_t). The pointer to this structure is contained in

 $channel->Tx\_control\_ptr->data\_pump\_ptr.$ 

Fills up the appropriate fields of this structure with default values (carrier phase, baud frac, scrambler register and so on (refer to

Section 3.1.6.4 for details)).

This function is called by the Tx\_V22\_init() functions of the v22.c

module.

**Returns:** None

```
struct channel_t * channel;
...

DPSK_modulator_init (channel);
...
```

# **DPSK\_modulator**

## Call(s):

void DPSK modulator(struct channel t \* channel)

#### **Arguments:**

## Table 3-38. DPSK\_modulator arguments

channel	in	Pointer to the channel control data structure
---------	----	---

### **Description:**

This function performs all the operations of the DPSK transmitter. It gets bits of data from the Tx\_data[] buffer, processes them, and places the resultant samples into the Tx\_sample[] buffer. The number of generated samples, per call, is defined by *channel->Tx\_control\_ptr->number\_samples*. This function contains the scrambler, the encoder, digital modulator and digital filter. For more detailed information about these blocks refer to Section 3.1.6.2.

This function is called by the  $Tx\_data\_pump()$  function of the modem.c module via the  $channel->Tx\_control\_ptr-> data\_pump\_call\_func().$ 

**Returns:** 

None

```
struct channel_t * channel;
...

DPSK_modulator(channel);
...
```

# DPSK\_demodulator\_init

#### Call(s):

void DPSK demodulator init(struct channel t \* channel)

## **Arguments:**

### Table 3-39. DPSK\_demodulator\_init arguments

channel in	Pointer to the channel control data structure	
------------	---	--

**Description:** Initialization of the DPSK receiver control structure

(Rx\_control\_DPSK\_t). The pointer to this structure is contained in

 $channel->Rx\_control\_ptr->data\_pump\_ptr.$ 

Fills up the appropriate fields of this structure with default values

(carrier\_phase, carrier\_phase\_incr, total\_phase, baud\_frac,

descrambler\_register, filter\_buf and so on (refer to Section 3.1.6.5

for more details)).

This function is called by the Rx\_V22\_init() functions of the *v22.c* 

module.

**Returns:** None

# **DPSK\_demodulator**

#### Call(s):

void DPSK demodulator(struct channel t \* channel)

#### **Arguments:**

## Table 3-40. DPSK\_demodulator arguments

channel	in	Pointer to the channel control data structure
---------	----	---

### **Description:**

This function performs all the operations of the DPSK receiver. It gets samples from the Rx\_sample[] buffer, processes them, and places the resultant bits into the Rx\_data[] buffer. It contains the input bandpass filters, the automatic gain control (AGC), the demodulator, the decision block, the decoder, the descrambler, the carrier recovery, and the clock recovery. For more detailed information about these blocks refer to Section 3.1.6.3.

This function is called by the  $Rx\_data\_pump()$  function via  $channel->Rx\_control\_ptr->data\_pump\_call\_func()$ .

**Returns:** None

## 3.1.7 QAM

The Quadrature Amplitude modulation (QAM) technique is required to be used in the V.22bis modem. This technique is implemented in the qam.c module. This module consists of two main blocks: the QAM transmitter and the QAM receiver. In the following subsections, the operation of the modem transmitter and receiver are described.

The transmitter accepts data (bits) from the Data Terminal Equipment (DTE). It then performs the necessary processing to place this data into the proper form for transmission through the Public Switched Telephone Network. This process basically consists of the modulation of the baseband information (logical 1's and 0's sent by the DTE) into the passband of the communications channel for transmission. The receiver collects the information from the telephone network and converts it back into its original form, i.e. the bits sent by the DTE.

The *qam.c* module consists of two main blocks: the QAM transmitter and the QAM receiver. These blocks are called the data pump. They are executed every time the *Tx\_sample* buffer is almost empty (the modulator or TX data pump is executed) or when the *Rx\_sample* is full enough (the demodulator or RX data pump is executed). They are called from the *modem.c* module.

## 3.1.7.1 QAM transmitter

In QAM, the information is encoded as the phase and amplitude change of the transmitter carrier. With  $\phi(n)$  denoting the phase that contains the information to be transmitted, the transmitted signal s(n) is represented mathematically by

$$s(n) = A(n) \cos(\omega n + \phi(n)) \tag{4}$$

Where  $\omega$  is the carrier frequency. The parameter A(n) determines the amplitude of the transmitted signal. It can also be written as:

$$s(n) = A(n) (\cos(\omega n) \cos(\phi(n)) - \sin(\omega n) \sin(\phi(n)))$$
 (5)

The substitution of

$$I(n) = A(n) \cos(\phi(n))$$

$$Q(n) = -A(n) \sin(\phi(n))$$

into (5) results in (6) used to describe QAM modulation systems:

$$s(n) = I(n)\cos(\omega n) + Q(n)\sin(\omega n) \tag{6}$$

Each value of the  $\{I(n),Q(n)\}$  sequence corresponds to one signaling element (symbol) transmitted. The number of signaling elements transmitted per second is commonly referred to as the baud rate, which for V.22bis is set by the protocol to 600. The set of possible values of the sequence  $\{I(n),Q(n)\}$  determines the signal constellation, which is

given in a two-dimensional representation. The signal constellation, commonly referred to as the constellation diagram, is a geometric picture that emphasizes the fact that the two channels are 90 degrees (Quadrature) out-of-phase. The V.22bis, with 600-baud rate, accomplishes the transmission of 2400 bps by encoding four incoming bits (quadbit) in a single baud. Since there are sixteen possible values for every quadbit, the constellation diagram for V.22bis contains sixteen points.

The overall block diagram for the QAM transmitter is shown in Fig 3.1.7.2. The basic structural blocks are the scrambler, encoder, digital modulator and digital filter.

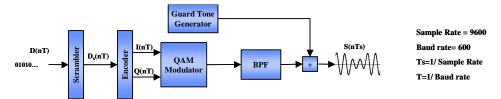


Figure 3-25. QAM transmitter block diagram

#### 3.1.7.1.1 Scrambler

The purpose of the scrambler is to randomize the input binary data sequence, which means that it converts this sequence into a pseudo-random binary sequence with some defined period. This is required for the clock recovery module. For more detailed information about the scrambler refer to Section 3.1.4.2.

#### 3.1.7.1.2 Encoder

The encoder implements encoding of the incoming sequence  $d_s(n)$  into the values of the sequence  $\{I(n),Q(n)\}$ , i.e. its function is to map every two incoming bits of the incoming sequence  $d_s(n)$  to a total phase. The total phase is then represented by the values of the sequence  $\{I(n),Q(n)\}$ , and the latter is transmitted. The mapping rule can be seen in Section 3.1.4.1.

In the current implementation of the encoder, it finds the appropriate I and Q values for the incoming quadbit (or dibit) and is executed once per baud.

#### 3.1.7.1.3 Modulator

The modulator moduLates the cosine wave. The output of the modulator is given by:

$$s(n) = I(n) \cos(\phi(n)) + Q(n) \sin(\phi(n))$$

Where  $\phi(n)$  is the current carrier phase and s(n) is the sample ready for transmission.

The modulating frequency for V.22bis is 1200 Hz for a calling modem and 2400 Hz for an answering modem.

#### 3.1.7.1.4 Transmit filter

The transmit lowpass filters are implemented using 48-tap FIR structures, whose frequency responses exhibit a raised-cosine shape. The raised-cosine response is used since it minimizes the intersymbol interference.

## 3.1.7.1.5 Guard tone generator

This module generates the guard tone, which is then added to the main signal. For V.22bis the frequency of the guard tone is  $1800 \text{ Hz} \pm 20 \text{ Hz}$  or  $550 \pm 20 \text{ Hz}$ , and is only transmitted when the modem is transmitting in the high channel (carrier frequency is 2400 Hz).

#### 3.1.7.2 QAM receiver

This subsection describes the functional blocks required to implement a QAM receiver. The receiver structure is more sophisticated than that of the transmitter. An overall diagram of the modem receiver is shown in Fig 3.1.7.3. The basic structural blocks of the modem receiver are the input bandpass filters (the fixed equalizer), the automatic gain control (AGC), the demodulator, the adaptive equalizer, the decision block, the decoder, the descrambler, the carrier recovery, and the clock recovery.

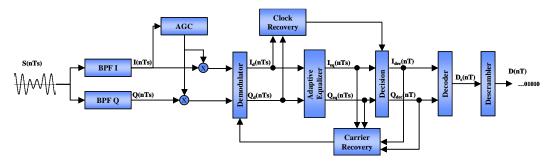


Figure 3-26. QAM modem receiver block diagram

## 3.1.7.2.1 Bandpass filters

The input signal is bandpass filtered for the following reasons:

- Rejection of out-of-band noise, including the rejection of the transmit signal spectrum
- Introduction of 90-degree relative phase shift required for I and Q channel separation
- Fixed equalization for line distortion

In the current implementation a 38-tap bandpass FIR filter is used.

## 3.1.7.2.2 Automatic Gain Control (AGC)

The incoming signal level may vary over a wide range due to attenuation in the telephone line. However for clock and carrier recovery modules the signal level must be independent of the attenuation introduced by the communications channel and remain constant. Automatic Gain Control performs this task. It adjusts the envelope of the I and Q channels so that they are of the same magnitude. The algorithm of AGC that is used in the current implementation is described below.

The AGC monitors the I channel of the receiver and calculates the gain correction factor. The gain correction factor is calculated once every three baud in a two-step process. First, the three maximum values of the signal, each one corresponding to one baud (16 samples), are monitored and added to each other. The previous value of the average signal level is then added to this sum and divided by four to obtain the new average signal level. The average signal level is then filtered by a first-order IIR filter.

In the second step, the gain correction factor is calculated. The gain correction factor is the result of dividing the maximum-allowed signal level (or the middle value of the signal level window for 2400 bit/s) by the filtered average signal level. For a transmission speed of 2400 bit/s the AGC adjusts the signal to keep it in a certain window, because its amplitude may vary.

Also the AGC block determines if a carrier is present in the received signal by comparing the average signal level with the threshold.

#### 3.1.7.2.3 Demodulator

With  $I_p(nT_s)$  and  $Q_p(nT_s)$  as inputs to the demodulator, the outputs  $I'(nT_s)$  and  $Q'(nT_s)$  are given by:

$$\begin{split} I'(nT_s) &= I_p(nT_s) \cos(\phi(nT_{s)}) + Q_p(nT_s) \sin(\phi(nT_s)) \\ Q'(nT_s) &= -I_p(nT_s) \sin(\phi(nT_s)) + Q_p(nT_s) \cos(\phi(nT_s)) \end{split}$$

Where  $\phi$  is the local carrier phase.

## 3.1.7.2.4 Decision block and Decoder

The decision block identifies the current quadrant from the values of the baseband I and Q and compares it with the previous quadrant to obtain the phase change. Using this phase change, the decoder retrieves the first two bits of the original quadbit (or whole dibit) according to Table 3.1.4.1. The last two bits of quadbit are then retrieved using the constellation diagram (see Fig 3.1.4.1).

#### 3.1.7.2.5 Descrambler

The descrambler is intended to recover the originally transmitted dibit. For more detailed information about the descrambler refer to Section 3.1.4.3.

## 3.1.7.2.6 Adaptive Equalizer

The Adaptive Equalizer is an adaptive filter that compensates for intersymbol interference (ISI) and telephone line distortion. Adaptive equalizers are based on either a statistical approach, such as the least-mean square (LMS), or a deterministic approach, such as the recursive least-squares (RLS) algorithm. The major advantage of the LMS algorithm that is used in the current implementation is its computational simplicity.

An adaptive filter consists of two parts: an FIR filter and an adaptation algorithm that adjusts the coefficients of the filter to improve its performance. In the current implementation an LMS adaptation technique is used, so the equalizer's coefficients are updated using the following equation:

$$\mathbf{C}_{k+1} = \mathbf{C}_k + \Delta \, \boldsymbol{\varepsilon}_k \, \mathbf{V}_k$$

Where  $C_k$  is the vector of equalizer coefficients on the k iteration,  $\Delta$  defines the convergence speed of the equalizer,  $\varepsilon_k = I_k' - I_k^{\wedge}$  (the difference between the optimal and the actual value of I or Q) is the error signal on the k iteration,  $V_k$  is the vector of demodulated I or Q values (taken from the equalizer's buffer). There are two separate filters, one for the I-channel, and the other for the Q-channel. Therefore the coefficients of these filters are updated separately as well.

## 3.1.7.2.7 Clock Recovery

The purpose of Clock Recovery is to detect the middle of the baud. The decision block makes decisions exactly in the middle of the baud. The energy of the incoming signal at this point is maximal, so the probability of a mistake ocurring during the decision process is minimal.

In the current implementation the early-late method of Clock Recovery is used. Fig 3.1.7.3.7.1 shows the energy values for each sample over the baud. It can be seen in this Figure that if the energy sample E7 is in the middle of the baud, it has the highest energy value and the rest of the samples are located symmetrically around it, i.e. E1=E13, E2=E12, E3=E11 and so on. In other words E1-E13 = 0, E3-E11 = 0.

If E7 is shifted left and is not on top of the energy 'hill', then E11 becomes greater than E3 or E3 – E11 < 0. If E7 is shifted right, E3 – E11 > 0. So the difference (E3-E11) may be used as an error value. If this value is negative, it means that the middle of the local band

occurred earlier than the middle of the incoming baud. Therefore, the local baud clock must be delayed. If positive – the local baud clock must be advanced.

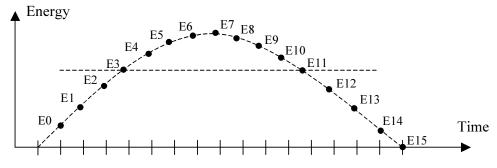


Figure 3-27. Signal energies over a baud

For smooth clock adjustment an Interpolation filter is used (Int I and Int Q in Fig.3.1.7.3). This has 16 filter coefficients from 64 filter banks. Each bank phase shifts the incoming signal by a certain value.

In the current implementation the energy of the third and the eleventh samples is calculated. The difference between these energies is then found. This difference is then lowpass filtered by a first-order IIR filter. The output of the filter is used as the input to the accumulator. If the value accumulated over a certain amount of time exceeds the threshold, the local baud clock is adjusted by choosing a different filterbank in the interpolation filter. The block diagram of the clock recovery algorithm is shown in Figure 3-28.

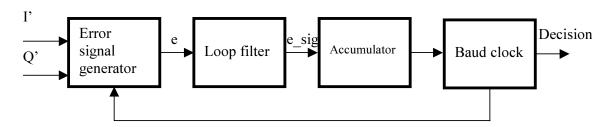


Figure 3-28. QAM Clock Recovery block diagram

# 3.1.7.2.8 Carrier Recovery

Carrier Recovery is used for adjusting the phase of the local carrier to match with the phase of the incoming carrier. It is very important to generate a local carrier that has the same phase and frequency as the incoming carrier because it is used in the demodulator to demodulate the incoming signal and retrieve the baseband information.

For implementation of Carrier Recovery a phase-locked loop is used (see Figure 3-29).

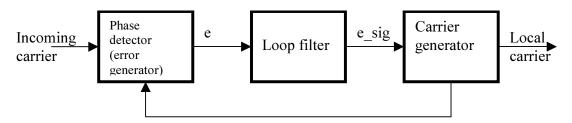


Figure 3-29. QAM Carrier Recovery block diagram

The phase detector generates an error that is used to synchronize the local carrier to the incoming carrier. This error signal contains the information about the phase and frequency difference between the local and the incoming carriers.

The constellation points are shown in Figure 3-30. The output of the phase detector is of the form

$$E(nT_b) = \frac{Q^{\wedge}(nT_b) \; I'(nT_b) - I^{\wedge}(nT_b) \; Q'(nT_b)}{Q^{\wedge}(nT_b) \; Q'(nT_b) + I^{\wedge}(nT_b) \; I'(nT_b)}$$

Where  $(I^{n}(nT_{b}))$  and  $Q^{n}(nT_{b})$  are the optimum decision points,  $(I^{n}(nT_{b}))$  and  $Q^{n}(nT_{b})$  are the points used to make a decision.

The error  $E(nT_b)$  is a geometrical distance from the point used to make the decision and the optimum decision point.

This error is then lowpass filtered by the *loop filter* (first-order IIR filter) and used to adjust the current value of the local carrier phase (either delay or advance the phase of the local carrier). So as a result the Clock Recovery block 'rotates' the constellation diagram to match the one shown in Figure 3-30. The filter coefficients and thresholds change depending on the state of the receiver (handshake mode or data mode).

The phase correction is made once per baud and is performed in the middle of the baud.

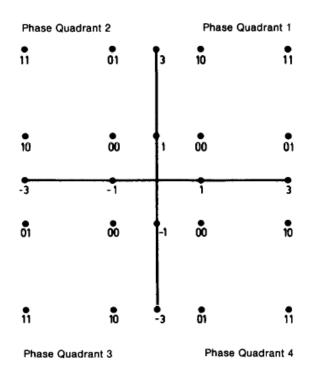


Figure 3-30. QAM receiver decision points

## 3.1.7.3 QAM modulator control structure

```
typedef struct Tx_control_QAM_t
{
   uint32 f_carrier;
   uint32 f_guard;

   uint16 carrier_phase;
   uint16 guard_phase;

   uint16 guard_phase_incr;
   uint16 guard_phase_incr;

   uint16 baud_frac;
   uint16 baud_frac;
   int16 amplitude_carrier;
   int16 amplitude guard;
```

```
scrambler_call_func_t scrambler;
bool scrambler_enabled;
uint32 scrambler_register;

int16 * constellation_phase_quadrant;
uint8 * phase_quadrant_change;
int16 current_I;
int16 current_Q;
uint8 current_quadrant;

uint32 filter_size;
int16 filter_coef[TX_QAM_FIXED_EQ_COEF_NUMBER];

int16 filter_buf[TX_QAM_FILTER_BUF_SIZE];
uint32 filter_buf_ptr;
};
```

- **f\_carrier** Carrier frequency in Hz. This field is initialized in function *Tx\_V22bis\_init*.
- **f\_guard** Guard tone frequency in Hz. This field is initialized in function *Tx\_V22bis\_init*.
- **carrier\_phase** Current phase of the carrier frequency generator in Q15 format. This field is initialized in function *QAM\_modulator\_init* to 0.
- **guard\_phase** Current phase of the guard tone generator in Q15 format. This field is initialized in function *QAM\_modulator\_init* to 0.
- **carrier\_phase\_incr** Carrier frequency phase increment per sample in Q15 format. This field is initialized in function *QAM\_modulator\_init*.
- **guard\_phase\_incr** Guard tone frequency phase increment per sample in Q15 format. This field is initialized in function *QAM\_modulator\_init*.
- **baud\_frac** Fractional part of the baud in Q14 format. (i.e. 0.5 = the middle of the baud). This field is initialized in function *OAM modulator init* to 0.
- **baud\_incr** Increment to <baud\_frac> per sample in Q15 format. This field is initialized in function *QAM\_modulator\_init*.
- **amplitude\_carrier** Amplitude of the output signal. This field is initialized in function *Tx V22bis init*.

- **amplitude\_guard** Amplitude of the Guard tone signal. This field is initialized in function  $Tx\_V22bis\_init$ .
- **scrambler** Pointer to the scrambler function. This field is initialized in function *Tx\_V22bis\_init*.
- **scrambler\_enabled** Indicates if the scrambler is enabled. This field is initialized in function *QAM\_modulator\_init* to *TRUE*.
- **scrambler\_register** Shift Register used by the scrambler. This field is initialized in function *QAM\_modulator\_init* to 0.
- **constellation\_phase\_quadrant** Pointer to the signal constellation of the first quadrant. This field is initialized in function *Tx\_V22bis\_init*.
- **phase\_quadrant\_change** Quadrant change according to the first two bits of quadbit. This field is initialized in function *Tx\_V22bis\_init*.
- **current\_I** Current In-phase signal for modulation. This field is initialized in function *QAM\_modulator\_init* to 0.
- **current\_Q** Current Quadrature-phase signal for modulation. This field is initialized in function *QAM\_modulator\_init* to 0.
- **current\_quadrant** Current Phase Quadrant. This field is initialized in function *QAM\_modulator\_init* to 0.
- **filter\_size** The number of coefficients used by the BPF. This field is initialized in function *QAM\_modulator\_init*.
- **filter\_coef** Coefficients of the BPF. This field is initialized in function *QAM\_modulator\_init*.
- **filter\_buf\_I** Cyclic buffer used by the filter algorithm of the I channel. This field is initialized in function *QAM\_modulator\_init* to 0.
- **filter\_buf\_Q** Cyclic buffer used by the filter algorithm of the Q channel. This field is initialized in function *QAM\_modulator\_init* to 0.
- **filter\_buf\_ptr** Pointer to the current element of the filter\_buf[]. This field is initialized in function *QAM\_modulator\_init*.

## 3.1.7.4 QAM demodulator control structure

```
typedef struct Rx_control_QAM_t
{
   uint32 f_carrier
   uint16 carrier_phase;
   uint16 carrier_phase_incr;

uint16 baud_frac;
```

#### **Data Pump**

```
uint16 baud incr;
uint32 filter size;
int16 filter mark i[RX QAM FIXED EQ COEF NUMBER];
int16 filter_mark_q[RX_QAM_FIXED_EQ_COEF_NUMBER];
int16 filter buf[RX QAM FILTER BUF SIZE];
uint32 filter_buf_ptr;
bool decision enabled;
scrambler call func t descrambler;
bool descrambler enabled;
uint32 descrambler register;
int16 * constellation phase quadrant;
uint8 * phase_quadrant_change;
uint8 last quadrant;
int16 constellation treshold;
int16 noise_threshold;
int16 agc;
uint32 agcave;
int16 avecnt;
int16 smax;
int16 CPLL1;
int16 CPLL2;
int16 BPLL1;
int16 BPLL2;
int16 BCNT;
int16 AGCPLL1;
```

#### **Data Pump**

```
int16 AGCPLL2;
  int32 cerror sig;
  int32 clerror sig;
   int32 clerror2 sig;
   uint32 baud nrg4;
  uint32 baud nrg8;
   int16 signal present;
   int16 clock corrected;
  int32 agc lp;
   int16 phase step;
  int16 BINTG;
  int16 AGC GC;
   int16 BAUD TH;
  int16 PHASE TH;
   int16 cl cnt;
  bool adaptive equalizer enabled;
   int32 filter adaptive ffe I coef[ADAPTIVE FILTER FFE COEF NUMBER];
   int16 filter adaptive ffe I buf[ADAPTIVE FILTER FFE BUF SIZE];
  int32
         filter adaptive ffe Q coef[ADAPTIVE FILTER FFE COEF NUMBER];
   int16 filter adaptive ffe Q buf[ADAPTIVE FILTER FFE BUF SIZE];
  uint32 filter adaptive ffe buf ptr;
  bool retrain enabled;
   uint8 retrain current s1;
  uint32 retrain s1 counter;
};
```

• **f\_carrier** - Carrier frequency in Hz. This field is initialized in function *Rx\_V22bis\_init*.

- **carrier\_phase** Current phase of the carrier frequency generator in Q15 format. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **carrier\_phase\_incr** Carrier frequency phase increment per sample in Q15 format. This field is initialized in function *QAM\_demodulator\_init*.
- **baud\_frac** Fractional part of the baud in Q15 format. (i.e. 0.5 = the middle of the baud). This field is initialized in function *QAM\_demodulator\_init* to 0.
- **baud\_incr** Increment to <baud\_frac> per sample in Q15 format. This field is initialized in function *QAM\_demodulator\_init*.
- **filter\_size** The number of coefficients used by the input BPF. This field is initialized in function *QAM\_demodulator\_init*.
- **filter\_mark\_I** Coefficients of BPF for the I channel in Q14 format. This field is initialized in function *QAM\_demodulator\_init*.
- **filter\_mark\_q** Coefficients of BPF for the Q channel in Q14 format. This field is initialized in function *QAM\_demodulator\_init*.
- **filter\_buf** Cyclic buffer used by the filter algorithm of the input BPFs. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **filter\_buf\_ptr** Pointer to the current element of the filter\_buf[]. This field is initialized in function *QAM\_demodulator\_init*.
- **decision\_enabled** Indicates if the decision block is enabled. This field is initialized in function *QAM\_demodulator\_init* to *TRUE*.
- **descrambler** Pointer to the descrambler function. This field is initialized in function *Rx\_V22bis\_init*.
- **descrambler\_enabled** Indicates if the descrambler is enabled. This field is initialized in function *QAM\_demodulator\_init* to *TRUE*.
- **descrambler\_register** Shift Register used by the descrambler. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **constellation\_phase\_quadrant** Pointer to the signal constellation of the first quadrant. This field is initialized in function *Rx\_V22bis\_init*.
- **phase\_quadrant\_change** Quadrant change according to the first two bits of the quadbit. This field is initialized in function *Rx\_V22bis\_init*.
- **last\_quadrant** The last phase Quadrant. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **constellation\_threshold** Threshold for constellation point detection. This field is initialized in function *Rx\_V22bis\_init*.
- **noise\_threshold** Threshold for signal/noise detection in Q14 format. This field is initialized in function *Rx\_V22bis\_init*.
- **agc** The AGC gain factor in Q11 format. This field is initialized in function *QAM\_demodulator\_init* to 1.

- **agcave** The current average signal level obtained after AGC in Q14 format. This field is initialized in function *QAM\_demodulator\_init* to 0.0625.
- **avecnt** The counter of the baud in the AGC block. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **smax** The maximum I value obtained after AGC over the baud in Q14 format. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **pre\_agcave** The current average signal level in Q14 format. This field is initialized in function *QAM\_demodulator\_init* to 0.0625.
- **pre\_smax** The maximum I value over the baud in Q14 format. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **CPLL1, CPLL2** Filter coefficients for the Carrier Recovery loop filter. This field is initialized in function *QAM\_demodulator\_init*.
- **BPLL1**, **BPLL2** Filter coefficients for the Clock Recovery loop filter. This field is initialized in function *QAM\_demodulator\_init*.
- **BCNT** The correction of the Baud Clock is performed once per **BCNT** baud. This field is initialized in function *QAM\_demodulator\_init*.
- **AGCPLL1, AGCPLL2** Filter coefficients for the filter of the AGC block. This field is initialized in function *QAM\_demodulator\_init*.
- **cerror\_sig** Filtered error signal for Carrier Recovery and the state of the loop filter. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **clerror\_sig** Filtered error signal for Clock Recovery and the state of the loop filter. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **clerror2\_sig** Indicates false lock of the baud clock for Clock Recovery. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **baud\_nrg4** Energy value of the 4th sample. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **baud\_nrg8** Energy value of the 8th sample. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **signal\_present** Indicates the carrier frequency signal presence. This field is initialized in function *QAM\_demodulator\_init* to *FALSE*.
- **clock\_corrected** Indicates that the Clock was corrected during the current baud. This field is initialized in function *QAM\_demodulator\_init* to *FALSE*.
- **agc\_lp** The state of the filter of the AGC block. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **last\_cerror** Contains the previous phase error in the Carrier Recovery block. This field is initialized in function *QAM\_demodulator\_init*.
- **phase\_unwrap\_enabled** Indicates whether phase unwrapping will be performed or not. This field is initialized in function *QAM\_demodulator\_init*.

- **phase\_step** The step of phase correction in the Carrier Recovery block. This field is initialized in function *QAM\_demodulator\_init*.
- **BINTG** Accumulates an error in Clock Recovery block. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **AGC\_GC** AGC gain correction coefficient. This field is initialized in function *QAM\_demodulator\_init*.
- **BAUD\_TH** The threshold for the clock error in Clock Recovery block. This field is initialized in function *QAM\_demodulator\_init*.
- **PHASE\_TH** The threshold for the phase error in Carrier Recovery block. This field is initialized in function *QAM\_demodulator\_init*.
- **cl\_cnt** The counter of baud in the Clock Recovery block. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **training\_mode** is TRUE if the modem is in training mode and FALSE if the modem is in data mode. This field is used in the Clock Recovery block and is Initialized in function *QAM\_demodulator\_init* to FALSE.
- **adaptive\_equalizer\_enabled** Indicates if the adaptive equalizer is enabled. This field is initialized in function *QAM\_demodulator\_init*.
- **adaptive\_equalizer\_training** Indicates if the adaptive equalizer is in training mode. This field is initialized in function *QAM\_demodulator\_init*.
- **filter\_adaptive\_ffe\_I\_coef, filter\_adaptive\_ffe\_Q\_coef** The coefficients of the adaptive filter for the I and Q channels. This field is initialized in function *QAM\_demodulator\_init*.
- filter\_adaptive\_ffe\_I\_buf, filter\_adaptive\_ffe\_Q\_buf Buffer of the adaptive filter for the I and Q channels. This field is initialized in function *OAM demodulator init*.
- **filter\_adaptive\_ffe\_buf\_ptr** Pointer to the current element of filter\_buf[]. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **mu** Defines the speed of adaptation of the adaptive equalizer, in Q14 format. This field is initialized in function *QAM\_demodulator\_init*.
- **int\_filter\_size** The number of coefficients used by the interpolation filter. This field is initialized in function *QAM\_demodulator\_init*.
- **int\_filter\_coef** Coefficients of the interpolation filter in Q14 format. This field is initialized in function *QAM\_demodulator\_init*.
- **int\_filter\_bufI** Cyclic buffer used by the interpolation filter algorithm for I channel. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **int\_filter\_bufQ** Cyclic buffer used by the interpolation filter algorithm for Q channel. This field is initialized in function *QAM\_demodulator\_init* to 0.

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- **int\_filter\_buf\_ptr** Pointer to the current element of the int\_filter\_bufI[]. This field is initialized in function *QAM\_demodulator\_init*.
- **int\_filter\_number** Specifies the number of active filterbanks of the interpolation filter. This field is initialized in function *QAM\_demodulator\_init* to 32.
- **retrain\_enabled** Specifies if auto retrain is enabled or disabled. This field is initialized in function *Rx\_V22bis\_init*.
- **retrain\_current\_S1** Contains one or zero. Used for S1 detection. This field is initialized in function *Rx\_V22bis\_handshake\_init* to 0.
- **retrain\_s1\_counter** Counter that is used in the S1 detection state during the retrain. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **retrain\_noise** Accumulator of signal error, used for retrain initiation in Q14 format. This field is initialized in function *QAM\_demodulator\_init* to 0.
- **retrain\_noise\_cnt** Baud counter when *retrain\_noise* is greater than *QAM\_RETRAIN\_NOISE\_THRESHOLD*. This field is initialized in function *QAM\_demodulator\_init* to 0.

# **QAM\_modulator\_init**

## Call(s):

void QAM modulator init(struct channel t \* channel)

#### **Arguments:**

## Table 3-41. QAM\_modulator\_init arguments

channel	in	Pointer to the channel control data structure
---------	----	---

**Description:** Initialization of the QAM transmitter control structure

 $(Tx\_control\_QAM\_t)$ . The pointer to this structure is contained in

*channel->Tx\_control\_ptr->data\_pump\_ptr.* 

Fills up the appropriate fields of this structure with default values (carrier phase, baud frac, scrambler register and so on (refer to

Section 3.1.7.4 for more details)).

This function is called by the Tx\_V22bis\_init() functions of the

v22bis.c module.

**Returns:** None

```
struct channel_t * channel;
...
QAM_modulator_init (channel);
...
```

# **QAM\_modulator**

## Call(s):

void QAM modulator(struct channel t \* channel)

#### **Arguments:**

## Table 3-42. QAM\_modulator arguments

channel	in	Pointer to the channel control data structure
---------	----	---

### **Description:**

This function performs all operations of the QAM transmitter. It gets bits of data from the Tx\_data[] buffer, processes them, and places the resultant samples into the Tx\_sample[] buffer. The number of generated samples, per call, is defined by

channel->Tx\_control\_ptr->number\_samples. This function contains the scrambler, the encoder, digital modulator and digital filter. For more detailed information about these blocks refer to Section 3.1.7.2.

This function is called by the  $Tx\_data\_pump()$  function of modem.c module via the  $channel->Tx\_control\_ptr-> data\_pump\_call\_func$  ().

**Returns:** None

# **QAM\_demodulator\_init**

#### Call(s):

void QAM demodulator init(struct channel t \* channel)

#### **Arguments:**

#### Table 3-43. QAM\_demodulator\_init arguments

channel	in	Pointer to the channel control data structure
---------	----	---

**Description:** Initialization of the QAM receiver control structure

 $(Rx\_control\_QAM\_t)$ . The pointer to this structure is contained in

*channel->Rx\_control\_ptr->data\_pump\_ptr.* 

Fills up the appropriate fields of this structure with default values

(carrier\_phase, carrier\_phase\_incr, total\_phase, baud\_frac,

descrambler\_register, filter\_buf and so on (refer to Section 3.1.7.5

for more details)).

This function is called by the Rx\_V22bis\_init() functions of the

*v22bis.c* module.

**Returns:** None

# **QAM\_demodulator**

## Call(s):

void QAM demodulator(struct channel t \* channel)

#### **Arguments:**

## Table 3-44. QAM\_demodulator arguments

channel	in	Pointer to the channel control data structure
---------	----	---

### **Description:**

This function performs all operations of the QAM receiver. It gets samples from the Rx\_sample[] buffer, processes them performing QAM demodulation (see 3.1.7.), and places the resultant bits into the Rx\_data[] buffer. It contains the input bandpass filters, the automatic gain control (AGC), the demodulator, the adaptive equalizer, the decision block, the decoder, the descrambler, the carrier recovery, and the clock recovery. For more detailed information about these blocks refer to Section 3.1.7.3.

This function is called from the function  $Rx\_data\_pump$  of modem.c module via the  $channel->Rx\_control\_ptr->data\_pump\_call\_func$  ().

**Returns:** None

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## 3.2 V.42 Error Correction

The V.42 error correcting protocol is used to deliver data correctly between two DCEs. Data is sent in frames. Each frame has a Frame Check Sequence (FCS) field. This field is used to determine whether a frame was damaged during transmission or not. If a frame was damaged, it will be automatically resent by the remote modem. Additionally, this protocol implements flow control between two DCEs.

The relationship of the V.42 module with the other modules is shown in Fig.3.2.1. The receiver takes data from the Rx\_data[] buffer, processes it according to the V.42 (V.14) protocol, and puts "useful" data into the Rx\_uart\_data[] buffer (from this buffer data goes to the terminal). The transmitter takes data from the Tx\_uart\_data[] buffer (data from the terminal comes to this buffer), wraps it according to the V.42 (V.14) protocol, and puts the result data into the Tx\_data[] buffer.

If the V.42bis module is enabled and used, it performs data compression (in the transmitter) and data decompression (in the receiver).

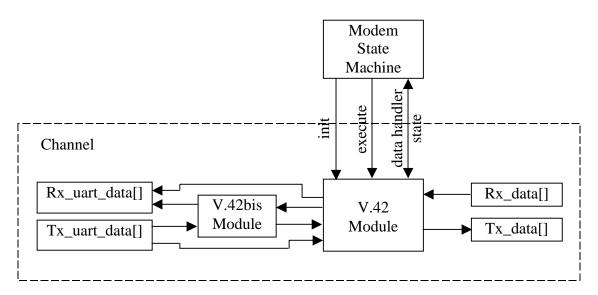


Figure 3-31. External relationships of the V.42 module

The V.42 module consists of three main functions:  $v42\_init()$ ,  $v42\_rx\_data()$ , and  $v42\_tx\_data()$ . The modem state machine calls the  $v42\_init()$  function when physical handshaking is completed. It initializes the data structure and obtains the required parameters. Each channel needs to have its own data structure, in order to handle incoming and outgoing data independently of the other channels. The v42\_init() function assigns the V.42 data structure to the Rx and Tx data control structures:

# Freescale Semiconductor, Inc.

#### **V.42 Error Correction**

```
void
v42_init (struct channel_t * channel, V42_DATA_STRUCT * data_ptr, ...)
{
         ...
channel->Rx_control_ptr->data_handler_ptr = data_ptr;
         channel->Tx_control_ptr->data_handler_ptr = data_ptr;
         ...
}
```

Since the V.42 receiver influences the V.42 transmitter, all receiver, transmitter and common (shared) fields are grouped into one data structure, having type V42\_DATA\_STRUCT, and the same instance of that structure is assigned to both the receiver and the transmitter.

Additionally, the data handler function has to be assigned to both the Rx and Tx data control structures:

Since incoming and outgoing data is handled in a different way by nature, different data handlers are required for each direction of data transfer. But, for the same channel, they will dial using the same V.42 data structure.

The  $v42\_rx\_data()$  function is called only if the amount of data in the Rx\_data[] buffer exceeds a preset threshold. The same takes place with the  $v42\_tx\_data()$  function. It is called only if the amount of data in the Tx\_data[] buffer is less than a preset threshold. These thresholds are also set in the v42 init() function:

```
void
v42_init (struct channel_t * channel, ...)
```

#### **V.42 Error Correction**

```
{
     ...
channel->Tx_control_ptr->number_n_bits = 8;
     channel->Rx_control_ptr->number_n_bits = 8;
     ...
}
```

# 3.2.1 General structure of the V.42 module

The structure of the V.42 module is shown in Fig.3.2.2. A received block of data is checked for the FCS first, (except if it is not a flag or if it is data according to the V.14 protocol), then it is passed to the receiver if the check sum is correct. The receiver handles that frame in compliance with the state of the V.42 module. Additionally, it can change the state of the V.42 module. If it is an information frame, the user data will be passed to the terminal (Rx\_uart\_data[] buffer). The data can be decompressed in the V42bis decoder if the V.42bis protocol was negotiated with the remote modem.

The transmitter takes data from the terminal (Tx\_uart\_data[] buffer), encodes it with the use of V.42bis compression, forms an I frame, calculates the FCS and passes it to the Tx\_data[] buffer. Depending on the state of the V.42 module, the transmitter can also issue supervisor and unnumbered frames.

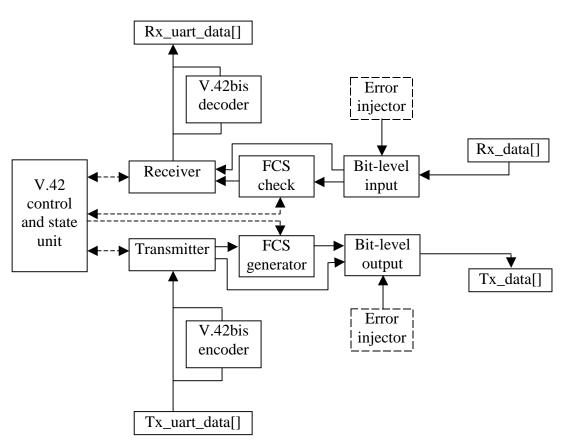


Figure 3-32. General structure of the V.42 module

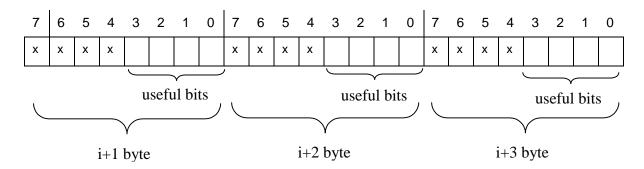
For demo and test purposes, an error injector can be used. Once per each period of time it inverts one bit of received and/or transmitted data. This simulates a noisy line. The time period between injections is a pseudorandom value. The time periods are derived from the pseudorandom generator based on the feedback shift register ("feedback shift register" is the name of a register type that (register) can be shifted and has a feedback link). To enable the error injector in each direction, constants \$\text{V42\_TX\_ERROR\_INJECTOR}\$ and \$\text{V42\_RX\_ERROR\_INJECTOR}\$ have to be defined in the v.42.h file.

A more detailed description of the receiver and transmitter is given in the following sections.

# 3.2.2 Accessing the data buffers

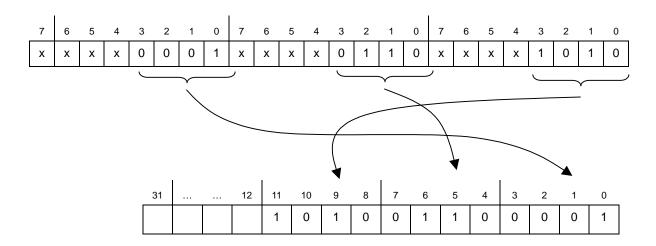
Each byte of the Rx\_data[] and Tx\_data[] buffers contains only nbits of useful data. This data occupies the least significant bits of the byte. The rest of the bits are useless.

Address: 0x00 0xFF



Byte i+1 is a byte that was received earlier than byte i+2, byte i+2 was received earlier than byte i+3 and so on. Inside of each group of bits, bit number 0 is the least significant bit (lsb), and bit number 3 (if nbits is 4) is the most significant bit (msb).

It is not convenient to process this data as it is. Some transposition is required. The following approach is implemented. Useful bits are stored consecutively in a 32-bit buffer rx buffer (in the receiver):



Now, bytes or, rather, nbits are placed consecutively one after another without gaps between them. The lsb of each byte resides in the next available bit in the  $rx\_buffer$  with the lowest number and the msb resides in the bit of the  $rx\_buffer$  having a higher number. In the above example, bit number 0 is the first received bit, while bit number 11 is the last received bit. Reading from this 32-bit buffer is performed from the lsb in order to follow the original data order. When data is read, the contents of the buffer can be stored easily in a register, right shifted by the number of read bytes, and then stored back into a buffer.

In the receiver, two input functions are implemented:  $v42\_getbits()$  and  $v42\_viewbits()$  to access the 32-bit buffer. At the beginning of the  $v42\_rx\_data()$  function, this 32-bit buffer is filled. In the body of this function the 32-bit buffer is read via the  $v42\_getbits()$  and  $v42\_viewbits()$  functions. It is important to understand, that the number of bits to read

(*n* parameter of these functions) cannot be greater than the size of the buffer (in this case 32 bits) minus the value of nbits. If the *v42\_getbits()* function is called with a *transparent* parameter set to TRUE, discarding the transparent bits also needs to be taken into account. So, the value passed to *n* parameter must be even less than 32 minus nbits.

The transmitter works in the same way. At the beginning of transmission, nbit fixed blocks of data are read from the 32 bit buffer, this data is then stored in the  $Tx_{data}[]$  buffer. The  $v42\_putbits()$  function is called in the body of the transmitter to place data into the 32-bit  $tx\_buffer$  buffer. It is important to understand that the number of bits to be written (n parameter) should not be grater than the size of the buffer (32 bits) minus nbits and minus the number of any inserted transparent bits (if the  $v42\_putbits()$  function is called with the transparent parameter set to TRUE).

# 3.2.3 Timing control

Four timers are implemented in the V.42 module: T400, T401, T403, and TI. A description of the first three timers is given in Section 9 "System parameters" of the ITU-T V.42 Recommendation. TI is an interval timer. It is described in Appendix II "Data forwarding conditions" (paragraph c) of the ITU-T V.42 Recommendation.

These timers are clocked by the *v42\_getbits()* and *v42\_putbits()* functions. When a correspondent function is called, it adds the number of bits requested to read/send to the counter of a timer. T400, T401, and T403 timers are clocked by both functions. This is done in order to take into account different baud rates of the transmitter and receiver, which takes place, for example, in the V.23 protocol. The thresholds of these three timers are oriented on a fixed amount of sent/received bits, regardless of the data pump protocol used. This means that, for example, the T400 timer overruns when its counter contains a value greater than *T400\_THRESHOLD*. This value corresponds to different values of absolute time, when V.21 (300 bps) and V.22 (1200 bps) are used.

The Interval Timer (TI) calculates absolute time. It determines a period, during which characters are accumulated to be sent in a single information frame. So, the threshold for this timer is set up in *v42 init()* function and depends on the data pump protocol used.

# 3.2.4 Detection phase

The V.42 receiver is a state machine shown in Fig.3.2.3. It works, mainly, in two states: "get data" according to the V.42 protocol, and "get data" according to the V.14 protocol.

The initial state is set up in the v42 init() function depending on the user's settings.

If V.42 is desired and the detection phase is enabled, the receiver determines the role of the modem via an *originator* variable. If it acts as the originator, the receiver detects an (answerer detection pattern) ADP sequence. If this sequence is recognized within a T400

#### **V.42 Error Correction**

time period, the receiver changes its state and waits for a frame opening flag. Otherwise, it starts working according to the V.14 protocol or it breaks the connection – depending on the user's settings.

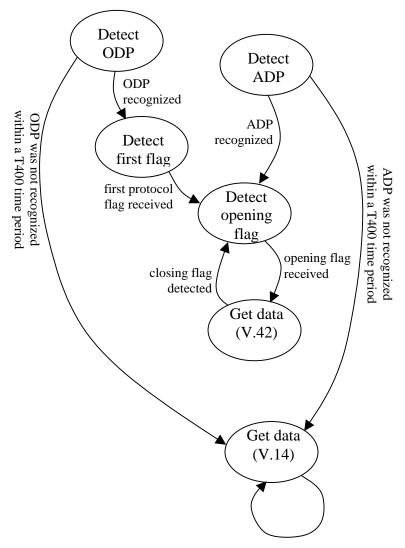


Figure 3-33. State machine of the V.42 receiver

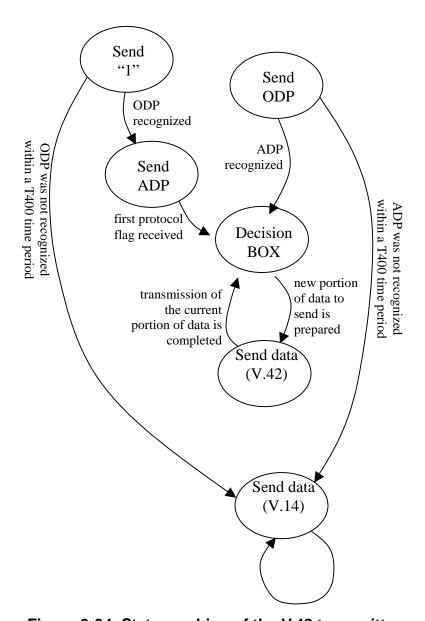


Figure 3-34. State machine of the V.42 transmitter

If the modem acts as an answerer, the receiver detects an (originator detection pattern) an ODP sequence. When an ODP sequence is received, the receiver changes its state and waits for the first protocol flag. This is done to change the transmitter to the correct state. After receiving the first flag, the receiver changes its state and waits for a frame opening flag. If ODP was not recognized within a T400 time period, it starts working according to the V.14 protocol or breaks the connection – depending on the user's settings.

If the modem is the originator, the transmitter sends an ODP sequence until an ADP sequence is recognized in the receiver. The state machine of the transmitter is shown in Fig.3.2.4. When an ADP sequence is received, the receiver changes the state of the transmitter to <code>DECISION\_BOX</code>. In this state, the transmitter makes a decision on what data to

send: flag, unnumbered frame, supervisor frame or information. Before sending the first protocol frame (command XID), it issues 16 flags.

If the modem is in answering mode, the transmitter sends "1" until the receiver receives an ODP sequence. When ODP is received, the receiver changes the state of the transmitter and finally starts sending an ADP sequence. When the remote modem gets the ADP sequence, it starts sending the first protocol frame flags. When a first flag is received, the receiver moves the transmitter to a <code>DECISION\_BOX</code> state.

# 3.2.5 V.42 Receiver

The structure of the V.42 receiver is shown in Figure 3-35. The following subsections give detailed information about the implementation of each block.

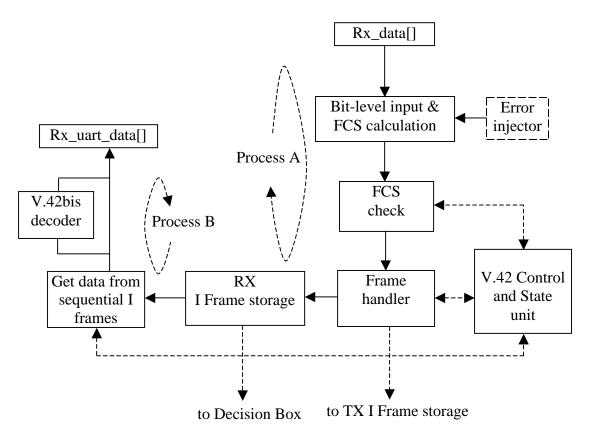


Figure 3-35. Structure of the V.42 receiver

# 3.2.5.1 Bit-level input

The receiver works in two states when it gets data according to the V.42 protocol:

 $v42\_{getting\_flag}$  and  $v42\_{getting\_data}$ .

In the  $V42\_GETTING\_FLAG$  state, it monitors the  $rx\_buffer$  in order to recognize a frame opening flag. ("01111110" is a frame opening and a frame closing flag. Each frame has to

#### **V.42 Error Correction**

be preceded with a frame opening flag and a frame-closing flag must follow the frame. The closing flag for one frame can serve as an opening flag for the next frame.)

Until any flag is recognized, the receiver extracts bit by bit from the rx\_buffer using the v42\_getbits(dp, &value, 1, FALSE) function. If a flag is recognized but it is not a frame opening flag (other flag follows this one), the receiver gets it using the v42\_getbits(dp, &value, 8, FALSE) function. The last parameter determines whether a read is performed discarding transparent bits (TRUE) or not (FALSE). When a frame opening flag is received, index and size variables are initialized to their default values. The receiver changes its state to V42\_GETTING\_DATA.

When the receiver gets a frame, it places the contents of the *value* variable into the rx\_frame buffer. Filling of the rx\_frame buffer with a frame, starts from the end of buffer, that is the first byte of the frame will be place into the last byte of the buffer, the second byte of the frame goes into the next to last position and so on. Before a byte is placed into the *rx\_frame* buffer, some analysis of the received data is required. Reading the data from the *rx\_buffer* is performed by discarding any "0" bits that directly follow five consecutive "1" bits. If noise occurred on the line, some bits (bytes) can change their value. There is a probability that some false transparent bits can occur (or some true transparent bits can be inverted to "1") in the bit stream. As a result, a frame has an incorrect length at the receiving modem, after the transparent bits are discarded from the bit stream. So, on the receiving side, the frame-closing flag can be misaligned from a byte boundary. In the same way, seven consecutive "1" bits or greater may occur in the bit stream. Such situations have to be eliminated and data already received should be ignored.

17 bits have to be examined at once. If there is a flag in the lsb, or the flag is placed as shown in the diagram below and rx\_count1 (current amount of last consecutive "1" bits) is equal to five, it is assumed that the frame-closing flag has been received.

Bit number 0 contains a transparent bit (not a data bit), because the rx\_count1 variable is equal to five. According to the specification, the transmitter must insert a "0" bit after 5 consecutive "1" bits, regardless of the following bit, and the receiver must discard any "0" bits that follow 5 consecutive "1" bits. So, before the previous reading, the 32-bit buffer could contain the xxxxxxxxx011111100111110yy value. During the previous reading, the value 111110yy was taken from the buffer, the rx\_count1 variable became equal to five, and the buffer was right shifted by 8 bits. The transparent bit remained in the buffer. It will be discarded when the next byte is taken from the 32-bit buffer, or when the closing flag occurs.

16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Х	х	х	х	х	х	х	х	0	1	1	1	1	1	1	0	0

So, receiving of this frame is now complete and the frame check sequence (already calculated) can be examined. When a frame has been processed, the receiver returns to the V42\_GETTING\_FLAG state.

#### V.42 Error Correction

Three different situations can abort the reception of data. The first situation occurs when one part of the flag lies in the right most byte, and another – in the next to right most byte:

						10											
Х	х	х	Х	х	0	1	1	1	1	1	1	0	Х	Х	Х	Х	

The second situation takes place when getting a byte from the  $rx\_buffer$  (discarding transparent bits) causes damage to the flag and as a result, the frame-closing flag will not be recognized. Assume the  $rx\_buffer$  is filled with the following bits and the  $rx\_count1$  variable is equal to five:

16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0	1	1	1	1	1	1	0	1	1	0	1	1	1	1	1	0

During execution of the  $v42\_getbits(dp, \&value, 8, TRUE)$  function, a "0" situated in bit number 0 will be discarded as a transparent bit along with a "0", situated in bit number 6 of the  $rx\_buffer$ . As a result, bit number 8 moves to position 6, while bit number 9 (the first bit of the flag) moves to position 7. After that, the value 0xFF is returned as a parameter into &value parameter, and the  $rx\_buffer$  is right shifted by 10 bits:

						10											
Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	0	1	1	1	1	1	1	

The bit stream is also examined for the occurrence of seven or more consecutive "1's". It is merged with checking on flag misalignment.

The third situation, when reception of data has to be aborted, is when a frame being received is unbounded (Section 8.5.5 of ITU-T V.42 Recommendation). In this implementation, the rx index variable equals "0".

For all of these situations the receiver moves back into the V42\_GETTING\_FLAG state, where it discards all bits preceding the frame-closing flag (and others flags until another frame-opening flag is detected).

In all other cases one byte can be read from the  $rx\_buffer$  without the threat of damaging the flag, except one. If the flag is placed in the second byte (bits 8 to 15 inclusive), reading the first byte has to be performed without discarding any transparent zeroes. After such a reading, the frame-closing flag will be appropriately aligned in the buffer. Any error will be disclosed after checking the FCS.

# 3.2.5.2 FCS check

This implementation supports 16- and 32-bit FCS. Calculation of the frame check sequence is performed during the reception of a frame. This means that this checksum is updated

upon receiving each byte from the bit stream. Such an approach reduces the peak load of the processor and improves overall system performance.

Depending on which sequence is used (16- or 32-bit), the appropriate checksum is calculated. In some cases, calculation of both checksums is required (Section 8.10.2 of ITU-T V.42 Recommendation). When a whole frame is received, the calculated check sum(s) is examined. The receiver only handles a frame if it has a valid FCS. Otherwise, the frame is discarded and the receiver changes its state to V42\_GETTING\_FLAG

## 3.2.5.3 Frame handler

The frame handler receives control, if the FCS check passed successfully. Depending on the control field of the frame, it identifies the corresponding parser to handle the frame in the appropriate way. If the received frame is not one of the frames listed in section 8.2.4.1 of the ITU-T V.42 Recommendation, then the FRMR frame will be issued.

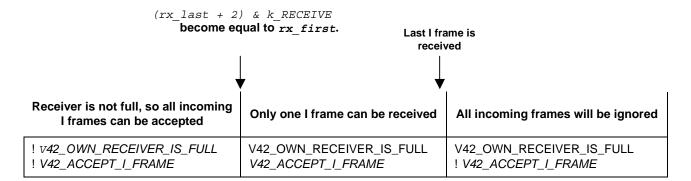
# 3.2.5.3.1 Information frame handling

Upon receiving an I frame, two conditions have to be analyzed:

- 1. Own-receiver is busy
- 2. Frame Reject condition

Own-receiver is busy condition.

The receiver accepts I frames only if it is not in a "busy" condition. Setting this flag is performed in two stages. In the first stage, the  $v42\_OWN\_RECEIVER\_IS\_FULL$  flag is set up along with  $v42\_ACCEPT\_I\_FRAME$  if there is space in the  $rx\_iframe$  storage for one additional I frame  $(((rx\_last + 2) & k\_RECEIVE) == rx\_first)$ . So, while the RNR frame is being sent to the remote modem, that modem has enough time to start sending the next I frame following the last frame already sent. Since the  $v42\_ACCEPT\_I\_FRAME$  flag is set, the receiver accepts that frame and clears that flag. This is the second stage. This approach avoids retransmission of the last frame when the receiver becomes empty.



There is a case, when the V42\_ACCEPT\_I\_FRAME flag will not be set up. If SREJ is used, then the next to last frame ("last" means a last frame that can be stored into rx\_iframe storage, but not the last frame in the series) can be damaged. After that the last frame is received. Later on, the next to last frame is received again after retransmission. In this case the rx\_last index points to the position next to the one pointed to by rx\_first. Only the V42\_OWN\_RECEIVER\_IS\_FULL flag has to be set up in this case.

## Frame-reject condition.

Two fields in  $rx_iframe$  storage are dedicated to frame-reject control. If  $notify_rej$  is set, the transmitter issues a reject frame (REJ or SREJ) for the corresponding frame.  $notify_rej$  is set only if the  $sent_rej$  field is cleared. This is done to avoid multiple rejection of the same I frame. The acceptance of I frames depends on the usage of the selective reject function. If selective reject is not used, or the  $v42\_ACCEPT\_I\_FRAME$  flag is set, only that frame is stored in  $rx_iframe$  storage, which corresponds to a position pointed to by  $rx_last$ . If a different frame is received, it is not accepted and the  $notify_rej$  and  $sent_rej$  fields are processed appropriately.

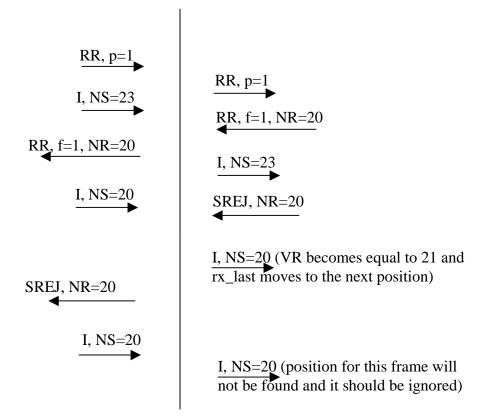
If selective reject is used, the receiver checks if there is a corresponding position in the  $rx\_iframe$  buffer for the received I frame. With the selective reject function enabled, the following situations are possible.

# rx\_iframe

		V	notify_rej	sent_rej
	22	0	1	
	23	1	0	
	24			
rx_last →	20	0	1	
	21	0	1	

#### Remote modem

### Soft modem



The remote modem issues a receiver ready (RR) request with the p bit set, and just after that starts to send an information frame (e.g. 23). The Soft modem receives the RR command and sends a reply RR with bit f set. According to the ITU-T V.42 recommendation, the remote modem has to resend all unacknowledged I frames, starting from I20. When the Soft modem receives I23, it sets the notify\_rej field for frames 20, 21, and 22. As a result the transmitter issues the SREJ frame, with the value 20 in its NR field. When the first I frame with NS=20 is received (which was sent by remote modem in reaction to the received RR reply), variable VR is incremented by 1 and the rx\_last pointer moves to the next position. When the remote modem receives SREJ, it assumes that frame I20 was not received and resends it again. The soft modem has to ignore this frame.

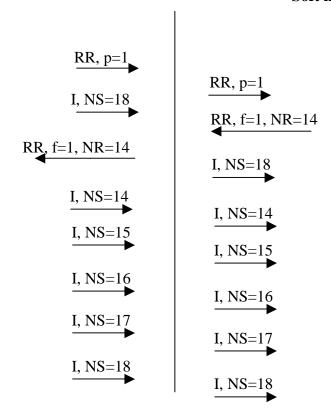
#### **V.42 Error Correction**

Another situation is shown below.

		V	notify_rej	sent_rej
		0		
		0		
	18	1		
rx_last →	14	0		
	15	0		

#### Remote modem

## Soft modem



The remote modem issues a RR request with the p bit set, and just after that starts to send an information frame (e.g. 18). The Soft modem receives the RR command and sends a reply RR with bit f set. According to the ITU-T V.42 recommendation, the remote modem has to resend all unacknowledged I frames, starting from I14 to I18 inclusive. Assume, I18 was properly received along with I14, I15, I16 and I17. When frame I17 is received, the VR variable becomes equal to 19 instead of 18. The <code>rx\_last</code> pointer is increased by 2 instead 1, because I18 is already in <code>rx\_iframe</code> storage. So, when the I18 frame is received again, the Soft modem has to ignore it.

As a result, only when the appropriate position is found, is data from the recently received frame placed there. *notify\_rej* and *sent\_rej* fields of all previous not received frames are processed appropriately.

Along with receiving frames, the receiver puts data from  $rx_iframe$  storage into the Rx\_uart\_data[] buffer. This is shown as independent Process B on Fig.3.2.5. Also, the receiver handles the situation there, when the  $rx_iframe$  buffer becomes empty.

The soft modem can issue an "early" REJ frame, if selective reject is not used. This is done to improve protocol performance. If a frame was not received properly (it was not properly bounded by two flags, or it was received with a CRC error), but its size is greater than *EARLY\_REJ\_CONDITION*, then there is a relatively high probability that it was an information frame. *notify\_rej* and *sent\_rej* fields are processed appropriately to issue a REJ frame without waiting for the next I frame with an incorrect NS.

# 3.2.5.3.2 Supervisor frame handling

Frames RR, RNR, REJ, and SREJ are handled in a specified way described in the ITU-T V.42 Recommendation. The two cases to be focused on here are the reception of a reply frame with the f bit set and the reception of an SREJ frame.

As mentioned earlier, the transmitter has to resend all I frames that are still being unacknowledged after receiving a reply frame with the f bit set. Assume an I frame has a sequence number (NS) equal to k that is currently being transmitted to the remote modem. If the receive sequence number (NR) of the received reply frame is less than k (by module 127), sending of this I frame should not be stopped if the selective reject function is used. Moreover, the flag for retransmission of this frame should not be set. This increases protocol performance. An example is described below.

Assume the transmitter is sending an I frame with NS equal to 18, when the receiver accepts an RR frame with the f bit set, and the NR field equal to 17. After the transmitter resends frame 17, it will start to resend frame 19, because frame 18 will have been received by the remote modem before frame 17.

tx_iframe			
	ns	retry	V
	16	0	0
tx_first ─►	17	1	1
	18	0	1
	19	1	1
	20	1	1
tx_last ─►	21	0	0

The same has to be done if a valid SREJ frame is received.

If selective reject is not used, transmission of the current I frame has to be stopped, if its NS is not equal to the NR of the received reply. Otherwise, transmission of this I frame is continued and the retry flag should not be set up. This increases protocol performance as well.

## 3.2.5.4 The V.42 Transmitter

The structure of the V.42 transmitter is shown in Fig.3.2.6. The following subsections give more detailed information about implementation of some of these blocks.

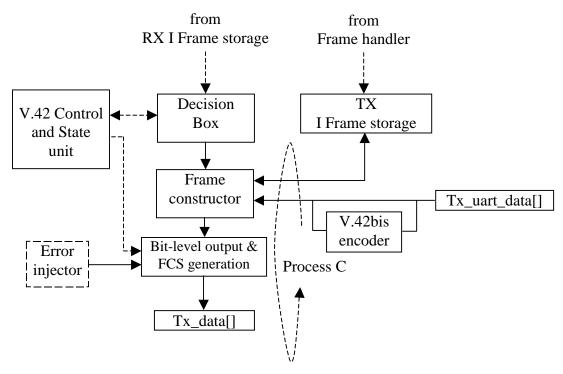


Figure 3-36. Structure of the V.42 transmitter

# 3.2.5.5 Frame Capsulation and Sending Data

The transmitter operates in four states when it sends data according to the V.42 protocol: V42\_DECISION\_BOX, V42\_DATA\_SENDING, V42\_FCS\_SENDING, and V42\_FLAG\_SENDING When the decision box receives control it has to make a decision on what frame to send. If there is no frame to send, it sends a flag and changes its state to V42\_FLAG\_SENDING This block can be divided logically into three sub blocks: the first block makes a decision when the V.42 protocol is in the protocol establishment phase (state < V42\_CONNECTED), the second block makes a decision when the V.42 protocol is in connected state (state == V42\_CONNECTED), and a third block makes a decision when the V.42 disconnection procedure starts (state == V42\_DISCONNECTING). The decision box also determines whether a frame will be sent as a command or a reply, and if the p/f bit is set or cleared.

If it is decided to send a frame, the transmitter constructs it. After this is done, the transmitter changes its state to  $V42\_DATA\_SENDING$  In this state, the transmitter puts data, which it has prepared in the previous state, into the  $tx\_buffer$ . Upon placing a byte of data into that buffer, the transmitter updates the frame control sum with this value. When all the data is sent (address, control and, possibly, information fields), the transmitter places the calculated frame control sum into the  $tx\_frame$  buffer directly after the sent data, and changes its state to V42 FCS SENDING

When the FCS field has been sent, the transmitter sends a frame closing flag by changing its state to V42\_FLAG\_SENDING After this is done, the transmitter returns back into the V42 DECISION BOX state.

# 3.2.5.6 Sending Information frames

Outgoing I frames are stored in tx\_iframe storage. Two flags are associated with each row in that storage: v (valid), and retry. If v is set, it means that the frame has not been acknowledged yet, and it should not be removed from tx\_iframe. The retry flag indicates whether a frame should be retransmitted or not. If the number of unacknowledged frames is equal to the window size, a new frame will not be put into this storage.

# 3.2.5.7 Sending Supervisor frames

Transmission of m-SREJ frames has to be described. These frames implement an acknowledgement function. So, on the receiving side, the NR field of these frames is used as acknowledgement of all previous frames. Assume the Soft Modem is in the following state in rx\_iframe storage:

rx_iframe				
		٧	notify_rej	sent_rej
tx_last ─►	16	0	0	1
	17	0	0	1
	18	1	0	0
	19	0	0	0
	20	1	1	1

This means that frames starting from NS 16 are not acknowledged. So, all frames, which have v bit cleared have to be added to the list of m-SREJ frame regardless of their sent\_rej field (this is done in the decision box by setting the notify\_rej field for all invalid frames, without taking into account the sent\_rej field). Otherwise, the NR field will be equal to 18 (incorrect) having not received acknowledgement for frames 16 and 17 from the remote modem. This situation causes a protocol error.

# 3.2.6 Support of the V.14 protocol

The V.42 module supports data transfer according to the V.14 protocol. This protocol is used when the error-correcting protocol cannot be established. The V.14 protocol cannot guarantee that data will be delivered to the remote modem correctly. Also, it cannot guarantee that received data was not damaged because of noise on the line.

The V.42 module falls back to the V.14 protocol, when the V.42 error-correcting protocol cannot be established between two modems, or when use of the V.42 protocol is denied by the user. The transition to the V.14 protocol is shown in the Figures 3.2.3 and 3.2.4.

When the receiver receives data by the V.14 protocol, it operates in the V14\_DATA\_IN state. The receiver moves to the V14\_EMPTY\_BUFFER sub-state, if the attempt to establish the V.42 protocol fails. During the V.42 detection phase the modem could have received data, in which it did not recognize the ADP/ODP patterns. In this sub-state it outputs all that data to the Rx\_uart\_data[] buffer. After that, it changes its state to V14\_1\_GETTING In the V14\_1\_GETTING state, the receiver gets data bit by bit from the rx\_buffer until it identifies a "0" bit. This bit in the V.14 protocol acts as a start bit of a byte. Eight data bits plus one "1" stop bit follows this start bit. So, the receiver changes its state to V14\_SYMBOL\_GETTING in order to get nine bits (eight data bits plus one stop bit) from the rx\_buffer and places the data bits into the Rx\_uart\_data[] buffer. When it is finished, the receiver returns back to the V14\_1\_GETTING state.

When the transmitter sends data according to the V.14 protocol, it operates in the V14\_DATA\_OUT state. If there is no data to send (there is no data in the Tx\_uart\_data[] buffer), it sends a series of "1" bits. Otherwise, it takes one byte from the Tx\_uart\_data[] buffer, puts a start bit "0" at the beginning and a stop bit "1" at the end of the sequence, and places the resultant ten bits into the tx\_buffer.

# 3.2.7 The V.42 module function description

This chapter describes the functions implemented in the V.42 module.

Function arguments for each routine are described as *in* or *inout*. An *in* argument means that the parameter value is an input only to the function. An *inout* argument means that a parameter is an input to the function, but the same parameter is also an output from the function. *Inout* parameters are typically input pointer variables in which the caller passes the address of a pre-allocated data structure to a function. The function stores its result within that data structure. The actual value of the *inout* pointer parameter is not changed.

# ret\_connection\_code

## Call(s):

# **Arguments:**

# Table 3-45. ret\_connection\_code arguments

channel	in	Pointer to the Channel control data structure.
v42_established	in	TRUE, if the V.42 protocol is established, FALSE otherwise.
params	in	Contains a mask of negotiated optional functions of the V.42 protocol for established connection.

# **Description:**

This function sets the <code>connection\_code</code> field of the <code>channel->Rx\_control\_ptr</code> data structure according to the established connection. As a result, the appropriate answer codes appear in the terminal window, with the description of the protocols used and the compression.

**Returns:** None.

## **Code example:**

ret\_connection\_code(channel, TRUE, dp->optional\_functions);

# v42\_getbits

## Call(s):

## **Arguments:**

## Table 3-46. v42\_getbits arguments

dp	in	Pointer to the V42 data structure.
value	inout	Pointer to a section of memory, where read bits are placed.
n	in	Number of bits to read.
transparency	in	If TRUE, each "0" bit that follows 5 contiguous "1" bits will be discarded as transparent.

# **Description:**

This function gets n bits from the rx\_buffer. These bits are placed into the lsb of the variable, pointed to by value. The remaining bits of that variable are cleared. If transparency == TRUE, the function discards transparent bits from the bit stream.

#### **Returns:**

V42\_READ\_SUCCESS, if the rx\_buffer contains the required number of bits, and V42\_RX\_BUFFER\_EMPTY otherwise.

```
uint32 value;
V42_DATA_STRUCT * dp = channel->Rx_control_ptr->data_handler_ptr;
...
v42_getbits(dp, &value, 8, TRUE);
...
```

# v42\_init

## Call(s):

# **Arguments:**

# Table 3-47. v42\_init arguments

channel	in	Pointer to the channel control data structure.
data_ptr	in	Pointer to initialise the V.42 data structure.
originator	in	TRUE, if the modem acts as the originator of the connection, FALSE, if the modem acts as an answerer.
ec_mode	in	Error-correction mode: 0x00: no error correction (V.14); 0x01: ARQ (V.42) / no error correction (V.14) - autodetect; 0x02: ARQ (V.42) only - hang up if V.42 cannot be established.
dphase_enabled	in	Detection phase enabled (TRUE) / disabled (FALSE) - valid only if ec_mode == 0x02.
srej	in	0x00: use of SREJ is denied by user; 0x01: use of single SREJ only is permitted by user; 0x02: use of multiple SREJ only is permitted by user; 0x03: use of both single and multiple SREJ permitted by user.
extended_fcs	in	TRUE, if the use of a 32-bit FCS is permitted by the user, FALSE otherwise.
test_frames	in	TRUE, if the use of TEST frames is permitted by the user, FALSE otherwise.
compression	in	Request for data compression: 0x00: data compression disabled; 0x01: data compression is permitted for transmitting; 0x02: data compression is permitted for receiving; 0x03: data compression is permitted for both directions of transfer.

**Description:** 

This function sets the fields of the current V.42 data structure, pointed to by the *data\_ptr*, to their default values. Additionally, it assigns the V.42 transmit and V.42 receive data handler routines to the channel control data structure.

**Returns:** None.

```
static V42_DATA_STRUCT v42data;
...

v42_init(channel,
v42data,
((S[13] & S13_ORIGINATE_MODE) == S13_ORIGINATE_MODE),
(S[22] & S22_M1),
((S[22] & S22_ENABLE_DETECTION) == S22_ENABLE_DETECTION),
((S[22] & S22_SREJ) >> BIT_SREJ),
((S[22] & S22_FCS_32) == S22_FCS_32),
((S[22] & S22_TEST_FRAMES) == S22_TEST_FRAMES),
(S[23] & S23_COMPRESSION));
```

# v42\_putbits

## Call(s):

## **Arguments:**

# Table 3-48. v42\_putbits arguments

dp	in	Pointer to the V42 data structure.
value	in	Data to write into the tx_buffer.
n	in	Number of bits to write.
transparency	in	If TRUE, a "0" bit will be inserted after 5 contiguous "1" bits as a transparent bit.

**Description:** 

This function writes n bits from value to the  $tx_buffer$ . If transparency == TRUE, the function inserts a "0" bit after each sequence of 5 contiguous "1" bits.

**Returns:** 

V42\_WRITE\_SUCCESS, if the tx\_buffer contains enough space to place the required amount of bits (including insertion of transparent bits), and V42 TX BUFFER FULL otherwise.

# **Code example:**

# 3.2.8.5.

# v42\_rx\_data

## Call(s):

void v42 rx data(struct channel t \* channel)

## **Arguments:**

## Table 3-49. v42\_rx\_data arguments

channel	in	Pointer to the channel control data structure.
---------	----	--

**Description:** This function gets data from the Rx\_data[] buffer, processes it

according to the V.42 protocol, and places the user's data into the

Rx\_uart\_data[] buffer.

**Returns:** None.

```
channel->Rx_control_ptr->data_handler_call_func = v42_rx_data;
...
    struct Rx_control_t *Rx_control = channel->Rx_control_ptr;
Rx_control->data_handler_call_func (channel);
```

# v42\_tx\_data

# Call(s):

void v42 tx data(struct channel t \* channel)

# **Arguments:**

## Table 3-50. v42\_tx\_data arguments

channel	in	Pointer to the channel control data structure.
---------	----	--

**Description:** This fund

This function gets data from the Tx\_uart\_data[] buffer, processes it according to the V.42 protocol, and places the resultant data into the

Tx\_data[] buffer.

**Returns:** None.

```
channel->Tx_control_ptr->data_handler_call_func = v42_tx_data;
...
    struct Tx_control_t *Tx_control = channel->Tx_control_ptr;
Tx_control->data_handler_call_func (channel);
```

# v42\_viewbits

# Call(s):

# **Arguments:**

# Table 3-51. v42\_viewbits arguments

dp	in	Pointer to the V42 data structure.
value	inout	Pointer to the memory space where read bits are placed.
n	in	Number of bits to read.

**Description:** 

This function reads *n* bits from the rx\_buffer, but they remain in the rx\_buffer. Read bits are placed into the lsb of the variable, pointed to by *value*. The rest of the bits of that variable are cleared.

**Returns:** 

V42\_READ\_SUCCESS, if the rx\_buffer contains the required number of bits, and V42\_RX\_BUFFER\_EMPTY otherwise.

```
"
uint32 status, value;
V42_DATA_STRUCT * dp = channel->Rx_control_ptr->data_handler_ptr;
"
status = v42_viewbits(dp, &value, 17);

if (status == V42_READ_SUCCESS)
{
"
"
}
```

# 3.3 V.42bis Data Compression Protocol

# 3.3.1 Protocol Background Information

The V.42bis algorithm defines a loss-less, single-pass, real-time compression scheme.

It uses a variant of the Lempel-Ziv-Welch (LZW) algorithm. The encoder translates 8 bit character sequences (strings) into a number of short transmission tokens - codewords. The algorithm uses adaptive dictionaries to store the most encountered strings along with codewords, which are used to represent those strings on either side of a V.42bis connection. The transmitter dictionary assembles itself from the input data file, while the receiver dictionary is assembled from the transmission tokens. The size of the dictionary can vary; the CCITT standard defines the minimal dictionary size as 512 nodes (strings). The actual dictionary size is established between modems at the V.42 negotiation time.

Both the soft modem and the remote modem exchange information on the size of dictionary they can support. The lower value is then selected. Both modems also negotiate a maximum string length, which can be stored in the dictionary, in the range from 6 to 250 characters.

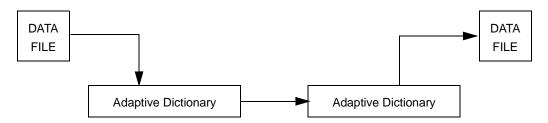


Figure 3-37. V.42bis Connection

The dictionary can be represented as a tree network, where each root node represents a single alphabet character, and vice versa, each character corresponds to the tree in the dictionary. All the trees represent a set of known strings that break out with a root node character. Each tree node corresponds to a set of strings of the dictionary, and a leaf node corresponds to a single known string. To begin with, each tree consists of only a root node with a unique codeword.

Data compression is achieved by growing a duplicate dictionary in both the encoder and decoder. The dictionaries are grown in lockstep using the transmitted address tokens. With a few exceptions (For the encoder – when it outputs an address token (codeword) that is already in the dictionary, for the decoder – when it inputs an address token (codeword) that is already in the dictionary), each address token creates a new dictionary node. The dictionary contains a pre-loaded section, and a learning section where new nodes are added during transmission. New nodes are created in both the encoder and decoder dictionaries so that both remain identical at all times.

Before encoding a data file, the dictionaries in both the encoder and decoder are pre-loaded from addresses 0 to 258. The first three dictionary addresses represent special control codes

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**V.42bis Data Compression Protocol** 

(ETM, FLUSH, STEPUP - for compressed mode, ECM, EID, RESET – for transparent mode), which enable the encoder to communicate with the decoder. Following data transmissions grow a new dictionary. An original 9 bit transmission code supports a dictionary with up to 512 nodes. When the dictionary exceeds 512 nodes, a 10 bit transmission code is utilised. At 1K nodes this transmission code is increased to 11 bits and so on. This is achieved by constantly clearing old dictionary nodes and recycling them according to a "least recently used" scheme. The next 256 nodes are then pre-loaded with an ASCII character code. A starting node is thus provided for each of the 256 ASCII characters, i.e., a root for 256 separate trees. Since the dictionary entries are shifted up by the three control codes, each ASCII code is stored in the address ASCII + 3. The actual string learning dictionary starts at address 259 and usually goes to 2K nodes.

During the encoding process, a tree dictionary is grown from the input file data. Each string of input characters, in effect, climbs the tree nodes to find matching strings already stored in the dictionary. The string matching process comes to an end when a codeword cannot be found in the dictionary. The codeword then identifies the "longest matching string" in the dictionary as learned during previous transmissions. This string is transmitted as the output codeword. This codeword, together with the last unmatched character, is stored in the next empty dictionary entry to add a new branch to the tree dictionary. The unmatched character is used as the first character of the next data string, continuing the string matching and transmission process. With few exceptions, each transmission creates a new node in both the encoder and decoder dictionary.

The decoder uses the input codeword code to retrieve a data string from its own duplicate dictionary. This is done by following the codeword trail backwards through the tree dictionary until a codeword smaller than 259 is detected. While jumping backwards from node to node through the dictionary, an ASCII character is recovered in each step. The character string is recovered in reverse order. The recovered string is the output data from the decoder. The decoder saves the input codeword in a buffer and waits for the next input codeword. The next input codeword is similarly converted back into an output data string. The first character of that next string is combined with the previous codeword (which was saved in the buffer) to create a new node in the receiver dictionary by storing the codeword - character pair in a next empty dictionary location. The next empty dictionary nodes are computed in the same way in both the encoder and decoder so that both the encoder dictionary node and the decoder dictionary node are stored in an identical dictionary codeword. Note that the encoder stores its new node one transmission earlier than the receiver. This can cause a problem if the encoder uses the just-created node to encode the next string. The receiver would then retrieve the wrong string because it does not yet have the correct node. The solution is for the transmitter to avoid using a newly-created node address to encode the string immediately following.

The ITU standard defines 2 modes of compression function: transparent and compressed. In the compressed mode, as mentioned above, the transmission is based on codewords. In the transparent mode, there is no codeword transmission and the characters entering the

encoder are not translated, being sent as a normal character. The transparent mode is very useful when a very mixed character stream is input to the encoder. There is a high probability, that each of the incoming characters will not be "matched" (such a situation often exists immediately after dictionary initialization). With every incoming "unmatched" character, the encoder outputs a byte, however the minimal codeword size is equal to 9, so in such a case the compression efficiency will be negative to a considerable percentage.

The V.42bis module contains a Data Compression Test that determines when data compression will be effective. Basically the test performs a comparison of bits before and after compression and calculates the efficiency of the mode used. The compression efficiency request is used to switch from the transparent to the compressed mode and vice versa, depending upon the characteristics of the transmitted data. According to the ITU standard, switching to compression mode is necessary when previous data (which was already transmitted and thus cannot be compressed) could have been compressed, increasing overall efficiency. Conversely switching into the transparent mode is required when the already compressed and transmitted data is being compressed with negative efficiency. It has to be pointed out that this switching from transparent to compressed modes is crucial to the efficient operation of the algorithm. The standard itself does not define a compression efficiency estimation procedure for when the switch is possible or required. Further details on this algorithm are included in section 3.3.2 of this document.

The Flush request is used to flush all outstanding data from the encoder. The standard does not note when the Flush request should be issued, but the assumption is made that it should be issued at least after a predefined timeout period (in the case of a user, working in the terminal mode).

The re-initialization request can be issued by the V.42 function as a reaction to an external event, or related to an inefficiency of compression with the current dictionary following an initialization. For the transmitter this means that the string matching procedure must be terminated. It is essential that the receiver's decoder understands this command.

# 3.4 Implementation Overview

The module implements the V.42bis data compression protocol used in conjunction with the error-correction protocol V.42, according to the CCITT Recommendation V.42bis. To enable V.42bis operation in a software modem, both V.42 and V.42bis must be compiled using the V42BIS definition directive. In this case, the V.42 error-correction module uses V.42bis compression is an integral part.

This V.42bis implementation provides buffered byte-oriented compression and decompression services, so it does not have to interface with the DTE buffers and can also be used as a standalone compressor on any platform, or used along with different v.42 implementations.

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Implementation Overview

The following declaration in the v42bis.h file defines default negotiation values suggested by the CCITT Recommendation V.42bis (Appendix II.1).

These values can be redefined with alternative values according to the CCITT Recommendation V.42bis. The maximum bits per codeword can be up to the value of  $2^{16}$ , because the implementation uses a 16-bit data type to represent the dictionary codewords.

Because the nodes in the dictionary are organized as a TRIE structure: (**Trie** is a kind of digital search tree. [Fredkin1960] introduced the *trie* terminology, which is abbreviated from "Re*trie*val".), the system provides fast dictionary look-up for a matched string and the current character, and effective dictionary modification (appending a new string and removing a node in the dictionary recovery mode).

The V.42bis algorithm constantly monitors the current compression ratios. If the ratio becomes negative, then the compression estimation function is switched off and communication changes to standard eight bit character codes (transparent mode). If the ratio becomes positive, communication changes to codeword transmission. The compression estimation function is used to measure the difference between the amount of bits transferred to an encoder output (ignoring control codewords), and the amount of bits of all bytes accepted by an encoder input. The transmitter can be in the transparent mode and transfer bytes to an output "as is", however the encoder can internally call the Data Compression Test. If this function becomes positive, it means that transmission using the transparent mode (the bits transferred is more than the bits accepted) is more effective, otherwise it is more useful to work in the compression mode. If the compression estimation function falls below a negative threshold then it a decision is made to switch over. Similarly, if the compression estimation function exceeds a positive threshold while still in the transparent mode - the transmitter is switched to the compression mode. The threshold for switching into the transparent mode is selected to be higher, to avoid continual switching away from compression mode for minor negative excursions.

# 3.4.1 Communication with V.42

The V.42 Error Control module invokes the V.42bis primitives during normal operations. The V.42 XID frames handle V42bis negotiation parameters during call initialization. There are two bits in the S23 register that either permit or disallow compression. Compression can be enabled from the caller to the answerer, from the answerer to the caller, or in both directions.

To transmit data with V.42bis compression enabled, the V.42 transmitter under normal conditions invokes V.42bis to encode data from the terminal. The encoded data is used for all further V.42 data processing. The special cyclic buffer – *encode buffer*, is used to

#### Implementation Overview

communicate between the V.42 transmitter and the V.42bis encoder. Two different functional blocks share this buffer:

- The V.42bis Encoding block
- The V.42 Frame Encapsulation block

The V.42bis encoding block compresses incoming data (from Tx\_uart\_data) and puts it into the *encode buffer*. The V.42 Frame Encapsulation block reads the data from that buffer and encapsulates it in the information field of an I-Frame. When the decision to encapsulate an I-Frame is made by the V.42 transmitter, it is possible that the *encode buffer* could be empty. In this case the V.42bis Flush request is used to produce non-zero information field data inside the encoder. (See Figure 3-38).

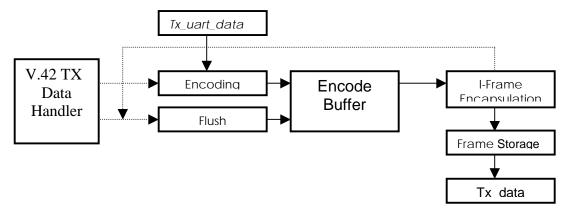


Figure 3-38. Compression flow

To receive data with V.42bis decompression enabled, the V.42 receiver under normal conditions invokes V.42bis to decode data from the remote modem. The special buffer –decode buffer, is used to communicate between the V.42 receiver and V.42bis decoder. The size of the decode buffer is estimated to fit the maximum possible compression ratio for the maximum information field length of the I-Frame. Thus, every processed frame becomes fully decompressed. All data from the decode buffer is then directly transferred into the DTE. The data that can not be transferred into the DTE immediately, because of flow control, remains in the decode buffer and the will be transferred next time, when I-Frame decapsulation restarts. (See Fig 3.3.3.2)

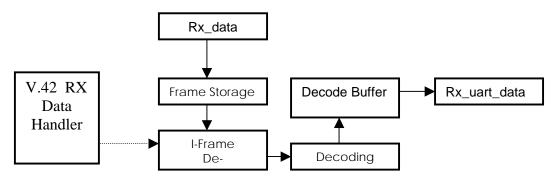


Figure 3-39. Decompression flow

This logic is enabled in the V.42 module when the V42BIS define statement is compiled. If compression in one direction is disabled, the V.42 module uses its own DTE data delivery scheme to/from the DCE. All the data structures of V.42bis are defined inside the V.42 module and have no relations with others, so the integral operation of V.42 and V.42bis protocols is completely transparent to other parts of the system.

# 3.4.2 Programming Interface

The programming interface with the v.42bis module consists of the following set of primitives:

Function	Description	Reference
v42bis_Init	Initialize v.42bis data compression	3.3.4.1
v42bis_Encode	Encode an input character	3.3.4.2
v42bis_Decode	Decode an buffer of characters	3.3.4.3
v42bis_Flush	Flush outstanding data from the encoder	3.3.4.4

Table 3-52. V42bis Functions

The V.42bis module maintains its internal encoder and decoder states in the v42bis\_t data structure. The structure itself can be located in the V.42bis service user memory block. The pointer to this structure must be specified to any interface function of this module.

Most of the functions return a value, specifying the amount of data that was processed. A special return value  $c_{\_ERROR}$  is used to indicate a V.42bis algorithm error, which may occur in the development phase only and is useful for developer. V.42bis shouldn't return this value on a tested & debugged product.

Note the routines do not verify input parameters for consistency. Thus the code speed is increased and code size is minimized. This is economical for embedded systems. The parameter validity is assumed to be the responsibility of a V.42bis service user. Invalid parameters may cause incorrect V.42bis function behavior or a system crash. The valid parameter conditions can be found in the 'Description' subsection for each routine.

# v42bis\_Init

## Call(s):

void v42bis Init (v42bis t \*dp, uint16 p1, uint8 p2)

## **Arguments:**

# Table 3-53. v42bis\_Init arguments

dp	in	pointer to the V.42bis control data structure
p1	in	Maximum number of codewords
p2	in	Maximum V.42bis string length

## **Description:**

Initializes the data compression function according to the CCITT Recommendation V.42bis section 7.2. The function initializes the V.42bis data structure according to the negotiated parameters. The routine shall be invoked immediately after XID negotiation by the V.42 protocol

The correct v.42bis function behavior is expected to adhere to the following conditions:

- 512 <= p1 <= 65535
- 6 <= p2 <= 250

according to the CCITT Recommendation V.42bis.

**Returns:** None.

```
v42bis_t v42bis;
uint16 code_words_num;
uint8 max_str_length;
...
v42bis_Init(&v42bis, code_words_num,
max_str_length);
```

# v42bis\_Encode

# Call(s):

```
int v42bis Encode(v42bis t *dp, uint8 ch, uint8 *obuf)
```

## **Arguments:**

## Table 3-54. v42bis\_Encode arguments

dp	in	Pointer to the v42bis data structure
ch	in	Character to encode
obuf	in	Output buffer for compressed data.

# **Description:**

Performs input character encoding according to the CCITT Recommendation V.42bis section 7.1. The function encodes a single character ch into a sequence of X bits, where 0 <= X <= 48 (with a 2048 codeword dictionary), and puts bytes of that sequence into the output buffer obuf. Bits that do not align to bytes will be taken into account during further  $v42bis\_Flush$  or  $v42bis\_Encode$  routine calls. Note that the user must specify an output buffer large enough otherwise the function can put data into an invalid memory location without reporting any error. The size of this buffer should be at least 6 bytes out of 2048 codewords. All bytes in the buffer should be initialised to zero before calling this function.

#### **Returns:**

The number of bytes put into the output buffer (can be zero), or  $c\_{\it ERROR}$  if an internal error has occurred (dictionary was overwritten with invalid data)

```
void Tx_data(...,v42bis_t *v42bis, uint8 *input_data, uint8 len)
{
    uint8    comp_buf[256*6];
    uint8    nBytes;
    uint8    *out = comp_buf;
    ...
    memset(comp_buf,0,sizeof(comp_buf));
    /*Perform compression from input_data buffer char by char*/
    for (i=0; i<len; i++)
        nBytes = v42bis_Encode(v42bis, input_data[i], out);
        out += nBytes;
    ...
}</pre>
```

# v42bis\_Decode

# Call(s):

# **Arguments:**

# Table 3-55. v42bis\_Encode arguments

dp	in	pointer to the v42bis data structure
ibuf	in	pointer to the input buffer
ilen	in	number of bytes in the input buffer
obuf	in	pointer to the buffer to output decompressed data.
olen	out	returns the actual byte count that was decoded and put into the output buffer $obuf$

# **Description:**

This function performs input buffer decoding according to the CCITT Recommendation V.42bis section 8. It decodes ilen bytes from the input buffer ibuf and puts decompressed olen bytes into the output buffer obuf. The size of the output buffer obuf must not be less than the established maximum string length (value of parameter p2 of  $v42bis\_Init$ ) for each character in the input buffer, i.e

olen ≤ ilen \* (maximum v.42bis string length)

#### **Returns:**

- 0 if all bytes from input the buffer were processed
- C\_ERROR if an invalid codeword in the input buffer was reached
- otherwise not all bytes from the input buffer were processed

# v42bis Flush

# Call(s):

int v42bis Flush (v42bis t \*dp, uint8 \*obuf,

# **Arguments:**

# Table 3-56. v42bis\_Flush arguments

dp	in	pointer to the v42bis data structure
obuf	in	pointer to the output buffer to flush remain data

## **Description:**

This function performs a FLUSH request according to the CCITT Recommendation V.42bis section 7.9. It flushes the outstanding bits from the encoder and returns the result as a byte aligned output, ready for transmission. Note that the user must specify an output buffer large enough otherwise the function can put data into an invalid memory location without reporting any error. The size of this buffer should be at least 4 bytes on 2048 codewords dictionary. All bytes in the buffer should be set to zero.

#### **Returns:**

The number of bytes that were put into the output buffer (can be zero).

# **Code example:**

```
void Tx data(..., v42bis t* v42bis, uint8 *input data, uint8 len)
{
                        flush buf[6];
                uint8
                uint8
                          nBytes;
uint16 encoded bytes = 0;
uint8
        needed to send something = 1;
/* >> Encoding goes here << */
encoded bytes += v42bis Encode(...);
if ((encoded bytes == 0) && (needed to send something))
int nBytes;
                memset(flush buf, 0, sizeof(flush buf));
                nBytes = v42bis Flush(v42bis, flush buf);
}
}
```

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### 3.4.3 Compression Ratio

This section describes a V.42bis compression ratio comparison for several files provided by the ITU-T V.56ter Recommendation. The results are obtained with the following V.42bis negotiation values:

- Maximum Number Of Codewords = 2048
- Maximum String Length = 32

These default values are used by most modems and placed into the scope of the recommendation. However, it is useful to know, that changing the above parameters has a great effect on the compression ratio, which depends very much on the characteristics of the data.

	Compression		Compressed			
V.56ter	capability	Original Size	V.56ter	spec.	V.42bis module	
			Size	Ratio	Size	Ratio
1x30.TST	Excellent	982 040	156 546	84.05%	154 897	84.22%
2x10.TST	Good	327 680	135 950	58.51%	135 844	58.54%
3x06.TST	Poor	196 608	142 125	27.71%	141 827	27.86%
4x04.TST	Catastrophic	131 072	131 268	-0.15%	131 072	0%

**Table 3-57. Compression Ratio Comparison** 

## 3.5 The V.14 Software Module

The V.14 Software module implements asynchronous-to-synchronous conversion according to the ITU-T V.14 recommendation.

The V.14 Software module contains two parts: theTx V.14 Data Handler and the Rx V.14 Data Handler.

## 3.5.1 V.14 Data Handler

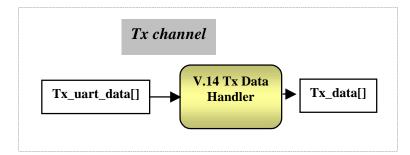


Figure 3-40. Tx V.14 Data Handler in the Tx channel

The V.14 Software Module

The Tx V.14 Data Handler takes data from the Tx\_uart\_data[] buffer encodes it and puts it into the Tx\_data[] buffer.

Each buffer element in the Tx\_uart\_data[] buffer contains 8 data bits(character).

If the Tx\_uart\_data[] buffer contains data, the Tx V.14 Data Handler takes the character from the Tx\_uart\_data[] using the Tx\_uart\_data\_read() function (see chapter 2.6.2 Circular Buffer Inline functions).

The Tx V.14 Data Handler then encodes the data according to the V.14 Recommendation (each character (8 data bits) is preceded by a Start bit (Start bit = "0") and is followed by a Stop bit (Stop bit = "1")). The general data flow according to the V.14 Recommendation is shown in Fig.3.4.3.

The Tx V.14 Data Handler then puts the encoded data into the Tx\_data[] buffer using the v14\_put\_bits() function. In the Tx\_data[] buffer each element contains tx\_nbits used bits.

tx\_nbits contains the number of bits per baud. Depending on the modulation protocol used, the value of tx\_nbits can be changed. Tx\_control->n\_bits contains the number of tx\_nbits. The v14\_put\_bits() function divides the encoded data by tx\_nbits data bits and puts the result into the Tx\_data[] buffer. The general operation of the Tx V.14 Data Handler is shown in Figure 3-41.

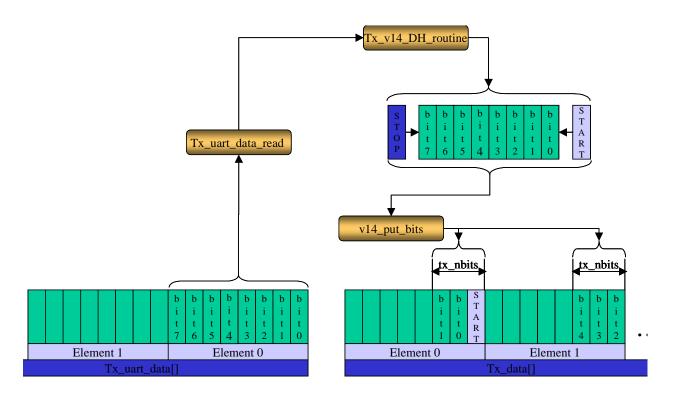


Figure 3-41. Tx V.14 Data Handler

The V.14 Software Module

If the Tx\_uart\_data buffer doesn't contain any data, the Tx V.14 Data Handler puts all "1"'s into the Tx\_data[] buffer using the v14\_put\_bits() function, according to the ITU V.14 Recommendation.

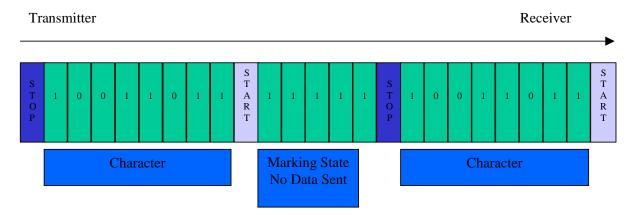


Figure 3-42. General Data Flow According to the V.14 Recommendation

The Tx V.14 Data Handler contains the following functions: Tx\_V14\_DH\_init(), Tx\_V14\_DH\_routine() and v14\_put\_bits().

# Tx\_V14\_DH\_init

#### Call(s):

void Tx\_V14\_DH\_init(struct channel\_t \* channel);

#### **Arguments:**

#### Table 3-58. Tx\_V14\_DH\_init arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** 

Initialisation of the Tx V14 Data Handler. The Tx V14 Data Handler takes control of the data structure. The user must call this function before calling the Tx\_V14\_DH\_routine() and v14\_put\_bits() functions. It is only necessary to call it once before Data Handler operation.

**Returns:** None.

```
...
struct channel_t channel;
...
Tx_V14_DH_init(channel);
...
Tx_control->data_handler_call_func=Tx_V14_DH_routine;
...
```

## Tx\_V14\_DH\_routine

#### Call(s):

void Tx V14 DH routine(struct channel t \* channel);

#### **Arguments:**

#### Table 3-59. Tx\_V14\_DH\_routine arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

#### **Description:**

This function gets the data from the Tx\_uart\_data[] buffer using the Tx\_uart\_data\_read() function, encodes it and puts the encoded data into the Tx\_data[] buffer using the v14\_put\_bits() function.

**Returns:** None.

#### **Code example:**

```
...
struct channel_t channel;
...
Tx_control->data_handler_call_func (channel);
...
```

switch(Tx control->data handler state)

#### **Code explanation:**

```
{
    default:
        case TX_DH_STATE_NORMAL:

Tx_control->data_handler_state contains the current state of the tx data handler.
If it is set to TX_DH_STATE_NORMAL, the tx data handler can transmit the data.

if (Tx_control->uart_data_tail != Tx_control->uart_data_head)
    {
        value = 0x200 | ((Tx_uart_data_read(Tx_control)) << 1);
    }
}</pre>
```

If the Tx\_uart\_data[] buffer contains data, take the data character(8 bits) from the Tx\_uart\_data[] buffer, add start and stop bits, and store the result in the variable value, then call the v14\_put\_bits() function to put 10 data bits from value into the Tx\_data[] buffer.

```
else
  v14 put bits(channel, 0xFF, 8);
```

v14 put bits(channel, value, 10);

If the Tx\_uart\_data[] buffer doesn't contain data, put 8 "1"s into the Tx\_data[] buffer.

## v14\_put\_bits

#### Call(s):

```
uint32 v14 put bits(struct channel t * channel, uint32 value, uint8 n);
```

#### **Arguments:**

#### Table 3-60. v14\_putbits arguments

channel	in	Pointer to the Channel control data structure
value	in	Contains the data bits being put into the Tx_data[] buffer
n	in	Contains the number of data bits being put into the Tx_data[] buffer

**Description:** This function puts data into the Tx\_data[] buffer.

**Returns:** None.

#### **Code example:**

```
...
struct channel_t channel;
...
value = 0x200 | ((Tx_uart_data_read(Tx_control)) << 1);
...
v14_put_bits(channel, value, 10);
...</pre>
```

#### **Code explanation:**

```
tx nbits = Tx control->n bits;
```

Gets the value of n\_bits (number of bits per baud – specified during data pump protocol initialization) from the Tx\_control structure.

```
number n bits = n/tx nbits+((n%tx nbits)!=0);
```

Calculates, how many elements in the Tx\_data[] buffer are required for n current data bits according to the current value of tx\_nbits. As you can see on the Fig.1-2, each element in the Tx\_data[] buffer contains tx\_nbits data bits only (in Fig.1-2 tx\_nbits=3). The other bits in the Tx\_data[] buffer element are not used. In this case, the n=10 bits must be split into portions, each 3 bits long so that each portion can be placed in a separate Tx\_data[] buffer element.

Example #1: n=10, tx\_nbits=3. There are 3 portions of 3 bits and one portion of 1 bit.

Portions of 3 bits must be put in elements #0-2 of the Tx\_data[] buffer, the last portion of one bit must be filled up with "1"s ( see if (n<tx\_nbits && n!=0) in the v14\_put\_bits() function) and must be put into element #3 of the Tx\_data[] buffer. Filling up with "1"s is required to eliminate misunderstanding 0's as the beginning of the character.

#### The V.14 Software Module

Example #2: n=10, tx\_nbits=2. There are 5 portions of 2 bits. It isn't required to fill any portions with "1"s.

The data pump takes only tx\_nbits bits from each Tx\_data[] buffer element.

Example:

```
data = value;
data <<= 32 - tx_nbits;
data >>= 32 - tx_nbits;
```

The variable data gets data bits from the variable value, and stores only  $tx\_nbits$  data bits to be put into the Tx\_data[] buffer.

```
Tx_data_write(Tx_control,data);
```

Inline function. Further description can be found in chapter 2.6.2 Circular Buffer Inline functions.

```
n -= tx_nbits;
```

After sending tx nbits data bits, n must be decremented by tx nbits.

```
value = value >> tx nbits;
```

tx nbits data bits are subtracted from the variable value.

If the remaining data bits are greater than 0 and less then  $tx_nbits$ , the element must be filled up with "1"s, to contain  $tx_nbits$  data bits.

#### 3.5.2 Rx V.14 Data Handler

The Rx V.14 Data Handler takes encoded data from the Rx\_data[] buffer, decodes it and puts the decoded data into the Rx\_uart\_data[] buffer.

The Rx\_data[] buffer contains rx\_nbits data bits in each element. The Rx\_uart\_data[] buffer contains a data character (8 data bits) in each element.

The Rx V.14 Data Handler gets the data from the Rx\_data[] buffer and stores it in the internal rx\_buffer[] using the v14\_get\_bits() function, decodes the data using the Rx\_V14\_DH\_routine() function and puts the encoded data into the Rx\_uart\_data[] buffer using the Rx\_uart\_data\_write() function (see chapter 2.6.2 Circular Buffer Inline functions).

The Rx V.14 Data Handler contains the following functions: R\_V14\_DH\_init(), Rx\_V14\_DH\_routine() and v14\_get\_bits().

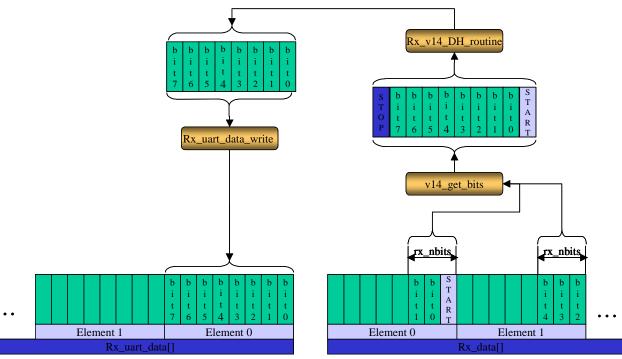
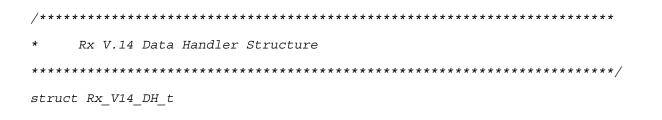


Figure 3-43. The Rx V.14 Data handler

The Rx V14 DH t structure is defined in "v14.h"



The V.14 Software Module

```
{
    uint32 rx_buffer;
    uint32 rx_num_bits;

enum V14_state_t rx_substate;
    uint16 rx_null_count;
};
```

The Rx\_V14\_DH\_t structure parameter descriptions are as follows:

- **rx\_buffer** Internal buffer for the Rx Data Handler into which the Rx Data Handler accumulates bits from the Rx\_data[] buffer and from which the Rx Data Handler takes data and puts it into the Rx\_uart\_data[] buffer. It is initialised in Rx\_V14\_DH\_init () to 0.
- **rx\_num\_bits** number of data bits in the rx\_buffer[]. It is initialised in Rx\_V14\_DH\_init() to 0.
- **rx\_substate** State identifier. It contains the current state of the Rx Data Handler. It is initialised in Rx\_V14\_DH\_init() to V14\_DH\_1\_GETTING.
- **rx\_null\_count** Specifies the number of received zeros to determine when there is no carrier. It is initialised in Rx\_V14\_DH\_init() to 0.

# Rx\_V14\_DH\_init

#### Call(s):

void Rx V14 DH init(struct channel t \* channel, struct Rx V14 DH t \*dp);

#### **Arguments:**

#### Table 3-61. Rx\_V14\_DH\_init arguments

channel	in	Pointer to the Channel control data structure
dp	in	Pointer to the Rx V14 Data Handler data structure

## **Description:**

Initialisation of the Rx V14 Data Handler. The Rx V14 Data Handler takes control of the data structure. The user must call this function before calling the Rx\_V14\_DH\_routine() and v14\_get\_bits() functions. It is only necessary to call it once before Data Handler operation.

**Returns:** None.

```
struct channel_t channel;
static struct Rx_V14_DH_t Rx_V14_DH;
...
Rx_V14_DH_init(channel ,&Rx_V14_DH);
...
Rx_control->data_handler_call_func=Rx_V14_DH_routine;
```

# Rx\_V14\_DH\_routine

#### Call(s):

```
void Rx V14 DH routine(struct channel t * channel);
```

#### **Arguments:**

#### Table 3-62. Rx\_V14\_DH\_routine arguments

channel	in	Pointer to the Channel control data structure
Charine	1111	1 officer to the charmer control data structure

#### **Description:**

This function gets the data from the Rx\_data[] buffer using the v14\_get\_bits() function, decodes it and puts the encoded data into the Rx\_uart\_data[] buffer using the Rx\_uart\_data\_write() function.

**Returns:** None.

#### **Code example:**

```
struct channel_t channel;

...

Rx_control->data_handler_call_func (channel);
...
```

#### **Code explanation:**

```
case V14_DH_1_GETTING:

    value = v14_get_bits(channel, dp, 1);

    if (value)
        dp->rx_null_count = 0;

    else
     {
        dp->rx_null_count ++;
        dp->rx_substate = V14_DH_SYMBOL_GETTING;
     }

    break;
```

dp->rx\_substate contains V14\_DH\_1\_GETTING. The Rx\_V14\_DH\_routine() function gets 1 data bit from the internal rx\_buffer using the v14\_get\_bits() function until a zero occurs in the data.

dp->rx\_substate then switches into V14\_DH\_SYMBOL\_GETTING dp->rx\_null\_count specifies the number of received zeros to determine the carrier loss.

```
case V14 DH SYMBOL GETTING:
                                       if
                                                   (dp->rx null count
((CARRIER_DETECT_TIME*Rx_control->baud_rate*Rx_control->n bits)/1000))
       Rx control->data handler state = RX DH STATE FAILED;
       return;
       }
    else
       value = v14 get bits(channel, dp, 9);
           if (value == 0)
              dp->rx null count += 9;
           else
              dp->rx null count = 0;
           if (value & 0x100)
             if (Rx control->data handler state != RX DH STATE ONLINE COMMAND)
                 Rx uart data write(Rx control, (uint8) value);
           dp->rx substate = V14 DH_1_GETTING;
    break;
```

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In substate V14\_DH\_SYMBOL\_GETTING the Rx\_V14\_DH\_routine() function determines the loss of the carrier. If there is no loss of carrier, it gets 9 data bits from the internal buffer dp->rx\_buffer[] using the v14\_get\_bits() function, and if the data packet isn't broken (Stop bit = 1) and the Data Handler state isn't in the Online command state, puts the data character (8 data bits) into the Rx\_uart\_data[] buffer using the Rx\_uart\_data\_write() function (see chapter 2.6.2 Circular Buffer Inline functions). If there is a loss of carrier the Data Handler goes to RX\_DH\_STATE\_FAILED.

# v14\_get\_bits

#### Call(s):

```
uint32 v14_get_bits(struct channel_t * channel, struct Rx_V14_DH_t * dp,
uint32 n);
```

#### **Arguments:**

#### Table 3-63. v14\_getbits arguments

channel	in	Pointer to the Channel control data structure
dp	in	Pointer to the Rx V14 Data Handler data structure
n	in	Contains the number of data bits required from the Rx_data[] buffer

**Description:** This function gets data from the Rx\_data[] buffer.

**Returns:** Data obtained from the Rx\_data[] buffer.

#### **Code example:**

```
struct channel_t channel;
struct Rx_V14_DH_t *dp;

uint32 value;
...
value = v14_get_bits(channel, dp, 1);
```

#### **Code explanation:**

```
while (dp->rx_num_bits < n)
{
         data = Rx_data_read(Rx_control);

        data <<= dp->rx_num_bits;
        dp->rx_buffer /= data;
        dp->rx_num_bits += rx_nbits;
     }
```

Get data bits from Rx\_data[] buffer, until the variable data contains n data bits, required by the v14\_get\_bits() function.

```
dp->rx_num_bits - number of data bits in the variable data.
```

*dp->rx\_buffer* – internal buffer for data bits.

```
ret_val = dp->rx_buffer;
```

```
ret_val <<= 32 - n;
ret_val >>= 32 - n;

dp->rx_buffer >>= n;
dp->rx_num_bits -= n;
return ret val;
```

When *dp->rx\_buffer* contains the required number of data bits, the data is stored in the variable *ret\_val* and returned to Rx\_V14\_DH\_routine() function.

## 3.6 V.8 Software Module

The V.8 Software Module determines automatically, prior to the initiation of modem handshake, the best available operational mode between two DCEs connected via the PSTN. It provides a timely indication to the Circuit Multiplication Equipment on the V-series modulation to be employed in a new session of data transmission. It also provides a means to enable a PSTN call to be passed on automatically to an appropriate DCE.

#### 3.6.1 V8 Data Handler

The Tx V8 Data Handler is used for transmission of CM, JM or CJ messages to the Remote Modem.

The Tx V8 Data Handler composes CM, JM or CJ sequences and puts them into the Tx\_data[] buffer.

The Tx V8 Data Handler contains the following functions: Tx\_V8\_DH\_init(), Tx\_V8\_DH\_routine() and v8\_put\_bits().

The  $Tx_v = DH_t$  structure, as defined in "v8.h".

The TX V8 DH t structure parameter descriptions:

- **v8\_state** State identifier. It contains the current state of the Tx V8 Data Handler. It is initialized in Tx V8 DH init () to V8 DCE CM.
- **v8\_sub\_state** State identifier. It contains the current sub state of the Tx V8 Data Handler. It is initialized in Tx\_V8\_DH\_init() to V8\_TEN\_ONES.
- **JM\_success** This variable allows the Tx\_V8\_DH\_routine() function to write data into the Tx\_data[] buffer and to send it to the other modem. It is initialized in Tx\_V8\_DH\_init() to 1 if the Soft Modem is in Calling Mode, and to 0 if the Modem is in Answering Mode.
- **CJ\_counter** Counter of the CJ messages sent. It is initialised in Tx\_V8\_DH\_init () to 0.

# Tx\_V8\_DH\_init

#### Call(s):

 $void \ {\tt Tx\_V8\_DH\_init} \ (struct \ channel\_t \ * \ channel, \ struct \ {\tt Tx\_V8\_DH\_t} \ *tx\_s, \ bool \ calling);$ 

#### **Arguments:**

#### Table 3-64. Tx\_V8\_DH\_init arguments

channel	in	Pointer to the Channel control data structure
tx_s	in	Pointer to the Tx V8 Data Handler data structure
calling	in	Boolean variable. Contains TRUE if the Soft Modem is in Calling Mode, and FALSE if the Soft Modem is in Answering Mode.

**Description:** Initialisation of the Tx V8 Data Handler. The Tx V8 Data Handler

takes control of the data structure. The user must call this function

before calling the Tx\_V8\_DH\_routine() and v8\_put\_bits()

functions. It is only necessary to call it once before Data Handler

operation.

**Returns:** None.

```
...
struct channel_t channel;
static struct Tx_V8_DH_t Tx_V8_DH;
...
bool calling;
...
calling = TRUE;
...
Tx_V8_DH_init(channel ,&Tx_V8_DH, calling);
...
Tx_control->data_handler_call_func=Tx_V8_DH_routine;
...
```

# Tx\_V8\_DH\_routine

#### Call(s):

void Tx V8 DH routine(struct channel t \* channel);

#### **Arguments:**

#### Table 3-65. Tx\_V8\_DH\_routine arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** 

This function composes CM, JM or CJ messages and puts them into the Tx\_data[] buffer using the v8\_put\_bits() function.

**Returns:** None.

#### **Code example:**

```
...
struct channel_t channel;
...
Tx_control->data_handler_call_func (channel);
...
```

#### **Code explanation:**

```
if (tx s->JM success)
```

Allows the Tx\_V8\_DH\_routine to send data. JM\_success allows data to be sent if the Soft Modem is in Calling Mode and in Answering Mode, if the JM message should be sent.

The state identifier *v8\_state* can be in one of the following states:

```
case V8 DCE CM:
```

The V8 Data Handler will send the CM message in Calling Mode or the JM message in Answering Mode.

State identifier v8\_sub\_state:

```
case V8_TEN_ONES:
```

The V8 Data Handler sends ten binary Ones and goes to the V8 SYNC state.

```
case V8 SYNC:
```

The V8 Data Handler sends the sync message and goes to the V8 CALL FUNCTION state.

```
case V8 CALL FUNCTION:
```

The V8 Data Handler sends the "call function" message and goes to the V8\_MODNO state.

```
case V8_MODN0:
case V8_MODN1:
case V8_MODN2:
case V8_ACCESS0:
```

The V8 Data Handler sends "modn0", "modn1", "modn2" and "access0" messages.

The Boolean variables in V8 MODNO, V8 MODN1, V8 MODN2 and V8 ACCESSO

are used to allow these messages to be sent (they are always sent in the CM message (Calling Mode)) and they are sent in the JM message, if these messages occurred in the received CM message (Answering Mode), according to the ITU-T V.8 Recommendation).

```
case V8_DCE_CJ:
```

In this case the V8 Data Handler will send the CJ message.

# v8\_put\_bits

#### Call(s):

uint16 v8\_put\_bits(struct channel\_t \* channel, uint16 current\_data);

#### **Arguments:**

#### Table 3-66. v8\_put\_bits arguments

channel	in	Pointer to the Channel control data structure
current_data	in	Contains data being put into the Tx_data[] buffer

**Description:** This function puts the data into the Tx\_data[] buffer.

**Returns:** None.

```
...
struct channel_t channel;
...
#define TEN_ONES 1023 //1111111111
...
v8_put_bits(channel, TEN_ONES);
```

#### 3.6.2 Rx V8 Data Handler

The Rx V8 Data Handler is used to receive CM, JM or CJ messages.

The Rx V8 Data Handler takes data from the Rx\_data[] buffer using the v8\_get\_bits() function, analyses received messages and transitions the Tx and Rx V8 Data Handlers into suitable states.

The Rx V8 Data Handler contains the following functions: Rx\_V8\_DH\_init(), Rx\_V8\_DH\_routine() and v8\_get\_bits().

The Rx V8 DH t structure, as defined in "v8.h".

```
/**************************
     Rx V.8 Data Handler Structure
*************************
struct Rx_V8_DH_t
  enum v8 calling state type v8 calling state;
  enum v8_state_type v8_state;
  enum v8_sub_state_type v8_sub_state;
  uint8 CM_OK_counter;
  uint8 ones_counter;
  uint8 CJ_counter;
  uint8 mod octet numb;
  uint8 null count;
  uint8 zeros CJ;
  bool modn0 oct;
  bool modn1 oct;
  bool modn2 oct;
  bool prot0_oct;
  bool access0 oct;
  bool V21 available;
  bool V23 available;
  bool V22 V22bis available;
};
```

The Rx V8 DH t structure parameter descriptions:

- **v8\_calling\_state** State identifier. It contains the current mode of the Soft Modem. It is initialized in Rx\_V8\_DH\_init () to CALLING\_TRUE if the Soft Modem is in Calling Mode, and to CALLING\_FALSE if the Soft Modem is in Answering Mode.
- **v8\_state** State identifier. It contains the current state of the Rx V8 Data Handler. It is initialized in Rx\_V8\_DH\_init () to V8\_DCE\_CM.
- **v8\_sub\_state** State identifier. It contains the current sub state of the Rx V8 Data Handler. It is initialised in Rx\_V8\_DH\_init() to V8\_TEN\_ONES.
- **CM\_OK\_counter** Counter of the received CM messages. It is initialised in Rx\_V8\_DH\_init() to 0.

- **ones\_counter** Counter of the 1's preceding each information sequence. It is initialised in Rx\_V8\_DH\_init() to 0.
- **CJ\_counter** Counter of the received CJ messages. It is initialised in Rx\_V8\_DH\_init() to 0.
- **mod\_octet\_numb** Contains 0 if the number of received modulation octets is "modn1", and 1 if the number of received modulation octets is "modn2". It is initialised in Rx\_V8\_DH\_init() to 0.
- **null\_count** Specifies the number of received 0's to determine the carrier loss. It is initialised in Rx\_V8\_DH\_init() to 0.
- **zeros\_CJ** Counter of the 0's in the Synchronization sequence (it contains 9 0's and one 1). It is initialized in Rx\_V8\_DH\_init() to 0.
- modn0\_oct Boolean variable. The JM message will include the same octets as in the CM message. modn0\_oct is TRUE if the received CM message contains the "modn0" octet. It is initialised in Rx\_V8\_DH\_init() to TRUE if the Soft Modem is in Calling Mode, and to FALSE if the Soft Modem is in Answering Mode.
- modn1\_oct Boolean variable. The JM message will include the same octets as in the CM message. modn1\_oct is TRUE if the received CM message contains the "modn1" octet. It is initialised in Rx\_V8\_DH\_init() to TRUE if the Soft Modem is in Calling Mode, and to FALSE if the Soft Modem is in Answering Mode.
- modn2\_oct Boolean variable. The JM message will include the same octets as in the CM message. modn2\_oct is TRUE if the received CM message contains the "modn2" octet. It is initialised in Rx\_V8\_DH\_init() to TRUE if the Soft Modem is in Calling Mode, and to FALSE if the Soft Modem is in Answering Mode.
- **prot0\_oct** Boolean variable. The JM message will include the same octets as in the CM message. prot0\_oct is TRUE if the received CM message contains the "prot0" octet. It is initialised in Rx\_V8\_DH\_init() to TRUE if the Soft Modem is in Calling Mode, and to FALSE if the Soft Modem is in Answering Mode.
- access0\_oct Boolean variable. The JM message will include the same octets as in the CM message. access0\_oct is TRUE if the received CM message contains the "access0" octet. It is initialised in Rx\_V8\_DH\_init() to TRUE if the Soft Modem is in Calling Mode, and to FALSE if the Soft Modem is in Answering Mode.
- V21\_available Boolean variable. V21\_available is TRUE if the Soft Modem is in Calling Mode, or if the Soft Modem is in Answering Mode and the Remote Modem supports V.21 Modulation protocol. It is initialised in Rx\_V8\_DH\_init() to TRUE if the Soft Modem is in Calling Mode, and to FALSE if the Soft Modem is in Answering Mode.
- V23\_available Boolean variable. V23\_available is TRUE if the Soft Modem is in Calling Mode, or if the Soft Modem is in Answering Mode and the Remote Modem supports V.23 Modulation protocol. It is initialised in Rx\_V8\_DH\_init() to TRUE if

- the Soft Modem is in Calling Mode, and to FALSE if the Soft Modem is in Answering Mode.
- V22\_V22bis\_available Boolean variable. V22\_V22bis\_available is TRUE if the Soft Modem is in Calling Mode, or if the Soft Modem is in Answering Mode and the Remote Modem supports V.22 or V.22bis Modulation protocols. It is initialised in Rx\_V8\_DH\_init() to TRUE if the Soft Modem is in Calling Mode, and to FALSE if the Soft Modem is in Answering Mode.

# Rx\_V8\_DH\_init

#### Call(s):

 $void \ \textit{Rx\_V8\_DH\_init} (\textit{struct channel\_t * channel}, \ \textit{struct} \ \textit{Rx\_V8\_DH\_t *rx\_s}, \ bool \ \textit{calling}); \\$ 

#### **Arguments:**

#### Table 3-67. Rx\_V8\_DH\_init arguments

channel	in	Pointer to the Channel control data structure
rx_s	in	Pointer to the Rx V8 Data Handler data structure
calling	in	Boolean variable. Contains TRUE, if the Soft Modem is in Calling Mode, and FALSE, if the Soft Modem is in Answering Mode.

**Description:** 

Initialisation of the Rx V8 Data Handler. The Rx V8 Data Handler takes control of the data structure. The user must call this function before calling the Rx\_V8\_DH\_routine() and v8\_get\_bits() functions. It is only necessary to call it once before Data Handler operation.

**Returns:** None.

```
...
struct channel_t channel;
static struct Rx_V8_DH_t Rx_V8_DH;
...
bool calling;
...
calling = TRUE;
...
Rx_V8_DH_init(channel ,&Rx_V8_DH, calling);
...
Rx_control->data_handler_call_func=x_V8_DH_routine;
...
```

## Rx\_V8\_DH\_routine

#### Call(s):

void Rx\_V8\_DH\_routine(struct channel\_t \* channel);

#### **Arguments:**

#### Table 3-68. Rx\_V8\_DH\_routine arguments

channel i	in	Pointer to the Channel control data structure
-----------	----	---

**Description:** This

This function gets data from the Rx\_data[] buffer using the v8\_get\_bits() function, analyzes CM, JM or CJ messages and transitions the Tx and Rx V8 Data Handlers into suitable states.

**Returns:** None.

#### **Code example:**

```
...
struct channel_t channel;
...
Rx_control->data_handler_call_func (channel);
...
```

#### **Code explanation:**

The state identifier v8 state can be in the following state:

```
case V8 DCE CM:
```

In this case the V8 Data Handler will receive the CM message in Answering Mode or JM message in Calling Mode.

The state identifier v8\_sub\_state can be in one of the following states:

```
case V8_TEN_ONES:
```

In this case the V8 Data Handler receives ten binary 1's and goes to the V8\_PRE\_SYNC state.

```
case V8 PRE SYNC:
```

In this case the V8 Data Handler determines the beginning of the sync message and goes to the *v8 SYNC* state.

```
case V8 SYNC:
```

In this case the V8 Data Handler interprets the sync message and goes to the V8\_CALL\_FUNCTION state. If the sync message wasn't received the V8 Data Handler will go back to the V8\_TEN\_ONES state.

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```
case V8_CALL_FUNCTION:
case V8 MODN1:
```

In these cases the V8 Data Handler interprets the "call function", "modn0", "modn1" and "modn2" messages, stores the TRUE value into the corresponding Boolean variables if the corresponding message was interpreted, and stores the TRUE value into the corresponding Boolean variables if the corresponding modulation protocol is supported by the remote modem.

After receiving two interpreted CM (Answering Mode) or JM (Calling Mode) messages the V8 Data Handler stores the available modulation protocols in hidden registers, and allows the CJ message to be sent (Calling Mode) or allows the JM message to be sent and goes to the V8 DCE CJ state (Answering Mode).

```
case V8_DCE_CJ:
```

In this case the V8 Data Handler will receive the CJ message, detect the loss of carrier (loss of connection) and direct the Soft Modem to establish a connection according the chosen modulation protocol or to break the connection if the modulation protocol wasn't chosen.

# v8\_get\_bits

#### Call(s):

uint16 v8\_get\_bits(struct channel\_t \* channel);

#### **Arguments:**

#### Table 3-69. v8\_get\_bits arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This function gets data from the Rx\_data[] buffer.

**Returns:** Data obtained from Rx\_data[] buffer.

```
...
struct channel_t channel;
...
uint16 current_data=0;
...
current_data=v8_get_bits(channel);
...
```

## 3.6.3 The V.8 start-up procedure in calling mode

The Soft Modem seeks to detect ANS or ANSam.

If ANSam is detected, the Soft Modem transmits no signal for a period of at least 1s prior to transmitting the CM message.

If ANS is detected, the Soft Modem will go into the physical handshake according to the preselected V-series modulation mode.

After no signal transmission the Soft Modem initiates transmission of the CM message using the Tx\_V8\_DH\_routine() function and conditions its receiver to detect the JM message using the Rx\_V8\_DH\_routine() function.

After a minimum of 2 identical JM sequences have been received the Soft Modem transmits the CJ message using the Tx\_V8\_DH\_routine() function. Following CJ, the Soft Modem transmits no signal for a period of 75 ms then proceeds in accordance with the selected V-series modulation mode.

## 3.6.4 The V.8 start-up procedure in answering mode

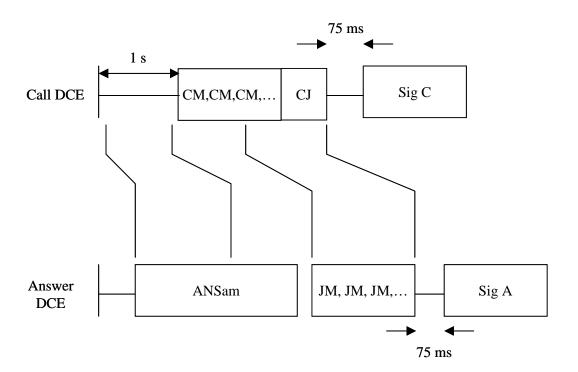
For a period of at least 0.2 s after connection to the line, the Soft Modem transmits no signal.

If the V.8 software module is enabled, the Soft Modem transmits ANSam and detects CM messages using the Rx\_V8\_DH\_routine() function.

Upon receiving a minimum of 2 identical CM sequences, the Soft Modem transmits the JM messages.

JM transmission continues until three CJ messages have been received using the Rx\_V8\_DH\_routine() function.

Upon receiving three CJ messages, the Soft Modem will proceed with the physical handshake according to the selected V-series modulation mode.



## 3.7 Support Modules

### 3.7.1 UART module

The UART is the traditional physical interface to the Terminal. UART, or Universal Asynchronous Receiver/Transmitter, is an integrated circuit that converts parallel input into serial output.

The MCF5407 contains two independent UARTs:

- UARTO
- UART1 can operate as a standard UART (identical to UART0) or in modem mode (as a USART). In the LDR Soft Modem, it works in modem mode (16-bit CODEC).

In the LDR Soft Modem UART0 is responsible for the DTE interface, and UART1 is responsible for the interface to the Si3044 DAA.

The UART module routines can be found in the "mcf5407\_uart.c" file.

The UART parameters structure uart params t is defined in the "mcf5407\_uart.h"

#### **Support Modules**

```
enum parity_t
                   {parity none,
                   parity even,
                   parity odd,
                   parity mark,
                   parity space};
                   {stop bits 1,
enum stop_bits_t
                   stop_bits_1_5,
                   stop bits 2};
enum channel_mode_t {channel_mode_normal,
                   channel mode automatic echo,
                   channel mode local loopback,
                   channel mode remote loopback};
//Only for UART1
enum uart1 mode t
                   {uart mode,
                   codec 8,
                   codec 16};
enum shift direction t {msb first,
                      lsb first};
/*****************************
* UART Parameters Structure
*************************
struct uart params t
  //Only for UARTO and UART1 in UART mode
  uint8 uart number;
  uint32 bits rate;
  enum data bits t data bits;
  enum parity_t parity;
  enum stop_bits_t stop_bits;
  enum channel_mode_t channel_mode;
  bool flow_control_transmit_on_off;
  //Only for UART1
  uint8 rx fifo threshold;
  uint8 tx_fifo_threshold;
  enum uart1_mode_t mode;
  //Only for UART1 in modem mode
  enum shift direction t shift direction;
};
```

The uart params t structure parameter descriptions:

- **uart\_number** UART number. The value must be equal to 0 (for UART0) or 1 (for UART1). Initialized in *mcf5407\_uart\_init()*.
- **bits\_rate** Bit rate of the UART. This is not used in modem mode. Initialized in *mcf5407\_uart\_init()*.
- **data\_bits** Bits per character. This value does not include start, parity, or stop bits. It can be equal to:

```
data_bits_5 = 5 bitsdata_bits_6 = 6 bits
```

#### **Support Modules**

- data\_bits\_7 = 7 bitsdata\_bits\_8 = 8 bits
- This is not used in modem mode. Initialized in *mcf5407\_uart\_init()*.
  - parity Parity type. It can be equal to:
    - parity\_none = No parity
    - parity\_even = Even parity
    - parity\_odd = Odd parity
    - parity\_mark = High parity
    - parity\_space = Low parity

This is not used in modem mode. Initialized in *mcf5407\_uart\_init()*.

**stop\_bits** - Stop-bit length. Selects the length of the stop bit appended to the transmitted character. It can be equal to:

- -- stop\_bits\_1 = 1 bit
- stop\_bits\_1\_5 = 1.5 bit
- -- stop\_bits\_2 = 2 bit

This is not used in modem mode. Initialized in *mcf5407\_uart\_init()*.

- **channel\_mode** Channel mode. It can be equal to:
  - channel\_mode\_normal = Normal
  - channel\_mode\_automatic\_echo = Automatic echo
  - channel\_mode\_local\_loopback = Local loop-back
  - channel\_mode\_remote\_loopback = Remote loop-back

This is used in both UART and modem modes. Initialized in *mcf5407\_uart\_init()*.

- **flow\_control\_transmit\_on\_off** The on/off flag for transmission. This is used to control flow between the DTE and DCE in the *flow\_control()* function. It can be equal to:
  - True = Data transmission from DTE to DCE is enabled
  - False = Data transmission from DTE to DCE is disabled

This is not used in modem mode. Initialized in *mcf5407\_uart\_init()*.

• **rx\_fifo\_threshold** - The Rx FIFO threshold. The threshold is one less than the value at which the Rx FIFO is considered to be full for the purpose of alerting the CPU that the Rx FIFO needs to be read. The value should be no more than 0x1F. It supports UART1 only and is used in both UART and modem modes. Initialized in *mcf5407\_uart\_init()*.

#### **Support Modules**

- **tx\_fifo\_threshold** Tx FIFO threshold. This threshold is the value at which the Tx FIFO is considered to be empty for the purpose of alerting the CPU that the Tx FIFO requires more data/samples. The value should be no more than 0x1F. It supports UART1 only and is used in both UART and modem modes. Initialized in *mcf5407\_uart\_init()*.
- **mode** UART1 mode. It can be equal to:
  - uart\_mode = UART mode
  - codec\_8 = 8-bit CODEC interface mode
  - codec\_16 = 16-bit CODEC interface mode

This supports UART1 only. Initialized in *mcf5407\_uart\_init()*.

- **shift\_direction** Shift direction. It can be equal to:
  - msb\_first = Samples/time slots are transferred msb first
  - lsb\_first = Samples/time slots are transferred lsb first

This supports UART1 only. Initialized in *mcf5407\_uart\_init()*.

## mcf5407\_uart\_init

#### Call(s):

#### **Arguments:**

#### Table 3-70. mcf5407\_uart\_init arguments

channel	in	Pointer to the Channel control data structure
uart_control	in	Array of the UART parameter structure. This array should only contain two elements (The MCF5407 only contains two UARTs).

#### **Description:**

This function initialises the *uart\_control[0]* and the *uart\_control[1]* structures to the default parameters, it then sets the UART module parameters by using the *mcf5407\_uart\_parameters\_set()* function.

Default parameters for UART0:

- data\_bits=data\_bits\_8;
- parity=parity\_none;
- stop\_bits=stop\_bits\_1;
- channel\_mode=channel\_mode\_normal;
- flow\_control\_transmit\_on\_off=FALSE;

#### Default parameters for UART1:

- channel\_mode=channel\_mode\_normal;
- rx\_fifo\_threshold=16;
- tx\_fifo\_threshold=16;
- mode=codec\_16;
- shift\_direction=msb\_first;

mcf5407\_uart\_init() enables interrupt detection for UART0 on RxRDY (receiver ready), and for UART1 on RxRDY and TxRDY (transmitter ready) by using mcf5407\_uart\_interrupt\_mask\_set().

It must be called before calling *daa\_init()*.

This function is called in *main()*.

#### **Returns:** None.

```
...
struct channel_t channel;
struct uart_params_t uart_control[2];
...
mcf5407_uart_init (&channel,uart_control);
...
```

# mcf5407\_uart\_parameters\_set

#### Call(s):

void mcf5407\_uart\_parameters\_set(struct uart\_params\_t \* uart\_params);

#### **Arguments:**

#### Table 3-71. mcf5407\_uart\_parameters\_set arguments

	uart_params	in	Pointer to the UART parameters structure.	
<b>Description:</b>		This function sets the parameters of the UART module according to the values contained in the structure pointed to by <i>uart_params</i> .		
		This	function is called in <i>mcf5407_uart_init()</i> and <i>state_machine()</i> .	
Return	s:	None		

```
...
struct uart_params_t uart_control;
...
mcf5407_uart_parameters_set(&uart_control);
...
```

# mcf5407\_uart\_interrupt\_mask\_set

Call(s):

#### **Arguments:**

#### Table 3-72. mcf5407\_uart\_interrupt\_mask\_set arguments

uart_number	in	UART number. This value must be equal to 0 (for UART0) or 1 (for UART1).
cos	in	Interrupt on change-of-state. Not used by UART1 in modem mode: FALSE = disabled TRUE = enabled
db	in	Interrupt on detection of a break. Not used by UART1 in modem mode: FALSE = disabled TRUE = enabled
rx_rdy	in	Interrupt on receiver ready:  FALSE = disabled  TRUE = enabled
tx_rdy	in	Interrupt on transmitter ready: FALSE = disabled TRUE = enabled

**Description:** This function sets the fields of the UART Interrupt Mask Registers

(UIMR) according to the input parameters.

This function is called in *mcf5407\_uart\_init(*).

**Returns:** None.

```
...
mcf5407_uart_interrupt_mask_set(1, FALSE, FALSE, TRUE, TRUE);
...
```

#### **Support Modules**

# Rx\_uart

#### Call(s):

```
void Rx uart (struct channel t * channel);
```

#### **Arguments:**

#### Table 3-73. Rx\_uart arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This function writes a character received from the UART module to

the Tx\_uart\_data[] circular buffer (using *in\_char\_uart()*).

**Returns:** None.

```
...
struct channel_t channel;
...
Rx_uart(&channel);
...
```

# Tx\_uart

### Call(s):

```
void Tx uart (struct channel t * channel);
```

### **Arguments:**

#### Table 3-74. Tx\_uart arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This function reads a character from the Rx\_uart\_data[] and sends it

to the UART module (using out\_char\_uart ()).

**Returns:** None.

```
struct channel_t channel;
...
Tx_uart(&channel);
...
```

## Tx\_uart\_all

## Call(s):

```
void Tx uart all (struct channel t * channel);
```

#### **Arguments:**

### Table 3-75. Tx\_uart\_all arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This function reads all available characters from Rx\_uart\_data[] and

sends them to the UART module (using out\_char\_uart ()).

**Returns:** None.

```
...
struct channel_t channel;
...
Tx_uart_all(&channel);
...
```

# in\_char\_uart

Call(s):

uint8 in char uart (void);

**Arguments:** None.

**Description:** This function gets a character from UART0.

**Returns:** The character obtained from UART0.

```
...
uint8 character;
...
character=in_char_uart();
...
```

# out\_char\_uart

## Call(s):

```
void out char uart (uint8 ch);
```

#### **Arguments:**

### Table 3-76. out\_char\_uart arguments

ch	in	Character for sending to UART0
----	----	--------------------------------

**Description:** This function sends the character (ch) to the UARTO.

**Returns:** None.

```
uint8 character='a';
...
out_char_uart(character);
...
```

# char\_present\_uart

Call(s):

bool char present uart(void);

**Arguments:** None.

**Description:** This function checks the availability of a character in UARTO.

**Returns:** TRUE = A character is available in UART0

FALSE = A character is not available in UARTO

```
...
bool character_is_available=FALSE;
...
character_is_available=char_present_uart();
...
```

# in\_sample\_codec

Call(s):

int16 in sample codec (void);

**Arguments:** None.

**Description:** This function gets a sample from UART1.

**Returns:** The sample obtained from UART1.

```
int16 sample;
...
sample=in_sample_codec();
...
```

# out\_sample\_codec

### Call(s):

void out sample codec (int16 sample);

#### **Arguments:**

#### Table 3-77. out\_sample\_codec arguments

sample	in	Sample for sending to UART1
--------	----	-----------------------------

**Description:** This function sends the *sample* to UART1.

**Returns:** None.

```
...
1nt16 sample=0;
...
out_sample_codec(sample);
...
```

# sample\_present\_codec

Call(s):

bool sample\_present\_codec (void);

**Arguments:** None.

**Description:** This function checks the availability of a sample in UART1.

**Returns:** TRUE = A sample is available in UART1

FALSE = A sample is not available in UART1

```
...
bool sample_is_available=FALSE;
...
sample_is_available = sample_present_codec();
...
```

## flow\_control

#### Call(s):

void flow\_control (struct channel\_t \* channel);

#### **Arguments:**

#### Table 3-78. flow control arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

#### **Description:**

This function performs Data Flow Control according to the S[18] register settings.

The Soft Modem implements flow control by detecting that <code>Tx\_uart\_data[]</code> is full enough (defined by the <code>UART\_DATA\_FLOW\_STOP</code> value, in "modem.h") and then negates the Request to Send (RTS) signal in the case of Hardware flow control or sends a special character XOFF into the data stream in the case of Software flow control, to stop the flow of data.

When the buffer  $T_{x\_uart\_data[]}$  is empty enough again (defined by the UART\_DATA\_FLOW\_START value, in "modem.h"), it asserts the RTS signal in the case of Hardware flow control or sends the special character XON into the data stream in the case of Software flow control to restart the flow of data.

The problem with software flow control is that the characters used to stop and start the flow of data can occur naturally in the data flow, enabling software flow control instructions. Using software flow control may prove satisfactory if you are transferring text files only.

Hardware flow control is much faster and much more reliable than software flow control, so it is highly recommended to use this.

This function is called in *state machine()*.

**Returns:** None.

#### Ketui 115.

```
...
struct channel_t channel;
...
flow_control(&channel);
```

### 3.7.2 DAA Interface

The DAA interface is a set of functions that implement control and interact with the modem daughter card based on the integrated direct access arrangement (DAA) Si3044. The data communication, between the M5407C3 and the modem daughter card, is performed via the UART1 module in modem mode (USART).

The DAA's behavior is modified by programming new coefficients into the memory register of the data codec and the line chip, thus it can adhere to country-specific global telephone line standards without hardware changes.

For detailed information about the Si3044 refer to the Si3044 Data Sheet, R2.01 (Si3044-DS201).

The DAA module routines can be found in the "si3044\_daa.c" file.

The DAA control data structure *daa\_control\_t* is defined in "si3044\_daa.h".

```
* States of DAA Rx&Tx Interface
*************************
enum tx_daa_state_t {TX_DAA_NORMAL,
                 TX DAA PRE READ REG,
                 TX_DAA_PRE_WRITE_REG,
                 TX DAA READ REG,
                 TX DAA WRITE REG };
enum rx_{daa}state_t \{RX_{DAA}NORMAL,
                 RX DAA READ REG,
                 RX DAA WRITE REG};
/****************************
* DAA Interface control Structure
*********************
struct daa_control_t
  volatile enum tx daa state t tx state;
  volatile enum rx daa state t rx state;
  uint8 reg number;
  volatile uint8 value;
  volatile bool value is true;
  uint8 delay;
  volatile uint32 sample counter;
};
```

The *daa\_control\_t* structure parameter descriptions:

- tx\_state Inner State of the Tx DAA. It can be equal to:
  - TX\_DAA\_NORMAL = Normal mode
  - TX\_DAA\_PRE\_READ\_REG = The mode before a Read register cycle
  - TX\_DAA\_PRE\_WRITE\_REG = The mode before a Write register cycle
  - TX\_DAA\_READ\_REG = Read register cycle

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— TX\_DAA\_WRITE\_REG = Write register cycle

It is initialized in *daa init()* to TX DAA NORMAL.

- **rx\_state** Inner State of the Rx DAA. It can be equal to:
  - RX DAA NORMAL = Normal mode
  - RX\_DAA\_READ\_REG = Read register cycle
  - RX\_DAA\_WRITE\_REG = Write register cycle

It is initialized in *daa\_init()* to *RX\_DAA\_NORMAL*.

- **reg\_number** Register number. This can be a value from 1 to 19 (the Si3044 has 19 control registers). It is initialized in *daa\_init()* to 0.
- **value** Register value. It is initialized in *daa\_init()* to 0.
- **value\_is\_true** Flag that the *value* is TRUE. It is initialized in *daa\_init()* to FALSE.
- **delay** Delay between sending a command and receiving a response. It is initialized in *daa\_init()* to 0.
- sample\_counter Counter of received samples. It is initialized in daa\_init() to 0.

## daa init

### Call(s):

#### **Arguments:**

### Table 3-79. daa\_init arguments

channel	In	Pointer to the Channel control data structure
daa_control	In	Pointer to the DAA Interface control structure

#### **Description:**

This function initialises the structure pointed to by <code>daa\_control</code>. It programs the clock generator setting registers 7, 8 and 9 to the appropriate divider ratios to obtain the desired sample rate. Then it writes a 0x80 into Register 6. This enables the charge pump and powers up the line-side chip (Si3015). The function sets the desired line interface parameters (calling <code>daa\_country\_set()</code>), sets Analog Receive&Transmit and AOUT (Speaker) Attenuation Levels (calling <code>daa\_rx\_level\_set()</code>, <code>daa\_tx\_level\_set()</code>, <code>daa\_aout\_level\_set()</code>).

This function is called once in *main()*.

**Returns:** 

None.

```
...
struct channel_t channel;
struct daa_control_t daa_control;
...
daa_init (&channel,&daa_control);
...
```

# daa\_country\_set

## Call(s):

void daa\_country\_set (struct daa\_control\_t \* daa\_control);

#### **Arguments:**

### Table 3-80. daa\_country\_set arguments

	daa_control	in	Pointer to the DAA Interface control structure
Descrip	otion:	17 an The f	function sets the desired line interface parameters (registers 16, d 18) according to the current chosen country (H[0] register). ull list of supported countries by the LDR Soft Modem is placed a SUPPORT_COUNTRIES[] array (in the modem.c file)
			function is called in <i>daa_init()</i> and in the Command Handler by GCI= <country according="" code="" t.35="" the="" to=""> command.</country>

**Returns:** None.

```
...
struct daa_control_t daa_control;
...
daa_country_set(&daa_control);
...
```

## daa\_rx\_level\_set

### Call(s):

```
void daa_rx_level_set (struct daa_control_t * daa_control);
```

#### **Arguments:**

### Table 3-81. daa\_rx\_level\_set arguments

daa control	in	Pointer to the DAA Interface control structure
aaa_00111101		Towner to the Drive mended definer directors

**Description:** This function sets the DAA Analog Receive Gain level (Register 15,

ARX[2:0]) according to the current setting in the S[16] register.

It can be set to 0dB, 3dB, 6dB, 9dB or12dB gain.

**Returns:** None.

```
...
struct daa_control_t daa_control;
...
daa_rx_level_set (&daa_control);
...
```

# daa\_tx\_level\_set

### Call(s):

```
void daa_tx_level_set (struct daa_control_t * daa_control);
```

#### **Arguments:**

#### Table 3-82. daa\_tx\_level\_set arguments

	daa_control	in	Pointer to the DAA Interface control structure
--	-------------	----	--

**Description:** 

This function sets the DAA Analog Transmit Attenuation level (Register 15, ATX[2:0]) according to the current setting in the S[17]

register.

It can be set to 0dB, 3dB, 6dB, 9dB or12dB attenuation.

**Returns:** None.

```
struct daa_control_t daa_control;
...
daa_tx_level_set (&daa_control);
...
```

## daa\_aout\_level\_set

## Call(s):

```
void daa aout level set (struct daa control t * daa control);
```

#### **Arguments:**

#### Table 3-83. daa\_aout\_level\_set arguments

daa_control	in	Pointer to the DAA Interface control structure

**Description:** 

This function sets the DAA Transmit&Receive Attenuation level for the call progress AOUT pin (Register 6 ATM[1:0], ARM[1:0]) according to the current setting in the H[1] register.

**Returns:** None.

```
...
struct daa_control_t daa_control;
...
daa_aout_level_set (&daa_control);
...
```

## daa\_aout\_mute

## Call(s):

```
void daa aout mute (struct daa control t * daa control);
```

#### **Arguments:**

#### Table 3-84. daa\_aout\_mute arguments

daa_control	in	Pointer to the DAA Interface control structure
-------------	----	--

**Description:** This function mutes the receive and transmit path for the call

progress AOUT pin (Register 6 ATM[1:0], ARM[1:0]).

**Returns:** None.

```
...
struct daa_control_t daa_control;
...
daa_aout_mute (&daa_control);
...
```

## daa\_ring\_detect

### Call(s):

bool daa ring detect (struct daa control t \* daa control);

#### **Arguments:**

### Table 3-85. daa\_ring\_detect arguments

			<u> </u>
daa_cc	ontrol	in	Pointer to the DAA Interface control structure

**Description:** This function reads and returns a value of the Ring detect field

(RDT) from Register 5.

This function is called in *state\_machine()*.

**Returns:** The value of field RDT of Register 5:

'TRUE = Indicates a ring is occurring.

FALSE = Reset either 4.5-9 seconds after last positive ring is

detected or when the system executes an off-hook.

```
...
struct daa_control_t daa_control;
bool ring=FALSE;
...
ring= daa_ring_detect (&daa_control);
...
```

## daa\_go\_off\_hook

## Call(s):

```
void daa_go_off_hook (struct daa_control_t * daa_control);
```

#### **Arguments:**

#### Table 3-86. daa\_go\_off\_hook arguments

daa_control	in	Pointer to the DAA Interface control structure
-------------	----	--

**Description:** This function causes the line-side chip to go off-hook. It generates an

off-hook command by applying a logic 0 to the  $\overline{OFHK}$  pin (if DAA\_HW\_HANG\_UP is defined) or by setting the OH bit in

Register 5 (if DAA\_HW\_HANG\_UP is not defined).

This function is called in *state\_machine()*, *Tx\_pulse()* and

*AT\_handler\_routine().* 

**Returns:** None.

```
...
struct daa_control_t daa_control;
...
daa_go_off_hook (&daa_control);
...
```

# daa\_go\_on\_hook

### Call(s):

void daa go on hook (struct daa control t \* daa control);

#### **Arguments:**

### Table 3-87. daa\_go\_on\_hook arguments

daa_control	in	Pointer to the DAA Interface control structure
-------------	----	--

#### **Description:**

This function causes the line-side chip to go on-hook It generates an on-hook command by applying a logic 1 to the OFHK pin (if DAA\_HW\_HANG\_UP is defined) or by resetting the OH bit in Register 5 (if DAA\_HW\_HANG\_UP is not defined).

This function is called in *state\_machine()*, *Tx\_pulse()* and

*AT\_handler\_routine()*.

#### **Returns:**

None.

```
struct daa_control_t daa_control;
...
daa_go_on_hook (&daa_control);
...
```

## daa\_read\_reg

### Call(s):

```
uint8 daa_read_reg
(struct daa_control_t * daa_control, uint8 reg_number);
```

#### **Arguments:**

#### Table 3-88. daa\_read\_reg arguments

daa_control	in	Pointer to the DAA Interface control structure
reg_number	in	Register number

#### **Description:**

This function reads the value of the DAA control register by number *reg\_number*.

It sets the register number

(daa\_control->reg\_number=reg\_number), changes the state of the

Tx DAA to TX\_DAA\_PRE\_READ\_REG

(daa\_control->tx\_state=TX\_DAA\_PRE\_READ\_REG), and waits till daa\_control->value\_is\_true becomes TRUE. It then returns the

value of the register contained in *daa\_control->value*.

#### **Returns:**

Value of the DAA control register by number reg\_number.

```
struct daa_control_t daa_control;
uint8 reg_value;
...
reg_value=daa_read_reg (&daa_control, 0x6);
...
```

## daa\_write\_reg

### Call(s):

```
void daa_write_reg
(struct daa control t * daa control, uint8 reg number, uint8 value);
```

#### **Arguments:**

#### Table 3-89. daa\_write\_reg arguments

daa_control	in	Pointer to the DAA Interface control structure	
reg_number	in	Register number	
value	in	Register Value	

### **Description:**

This function writes the *value* to the DAA control register by *reg\_number*.

It sets the register number

(daa\_control->reg\_number=reg\_number), sets the new register value (daa\_control->value=value), changes the state of the Tx DAA

to TX\_DAA\_PRE\_WRITE\_REG (daa\_control->tx\_state=

TX\_DAA\_PRE\_WRITE\_REG), and waits till daa\_control->value\_is\_true becomes TRUE.

**Returns:** None.

```
struct daa_control_t daa_control;
uint8 reg_value=0;
...
daa_write_reg (&daa_control, 0x6, reg_value);
...
```

## Tx\_daa

#### Call(s):

void Tx\_daa (struct channel\_t \* channel);

#### **Arguments:**

#### Table 3-90. Tx\_daa arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

#### **Description:**

This function is responsible for transmitting data to the DAA.

States of operation:

- TX\_DAA\_NORMAL = If the Tx\_sample[] circular buffer is not empty, the function reads one sample from it and sets the least significant bit to 0 (no secondary frame), otherwise it sets the resultant sample equal to 0 (silence). Finally it sends the resultant sample to the DAA by using out\_sample\_codec ().
- TX\_DAA\_PRE\_READ\_REG = If the Tx\_sample[] circular buffer is not empty, the function reads one sample from it and sets the least significant bit to 1 (calls secondary frame), otherwise it sets the resultant sample equal to 1 (generates silence and calls secondary frame). Changes the Tx DAA state to TX\_DAA\_READ\_REG. Finally it sends the resultant sample to the DAA by using out\_sample\_codec ().
- TX\_DAA\_PRE\_WRITE\_REG = If the Tx\_sample[] circular buffer is not empty, the function reads one sample from it and sets the least significant bit to 1 (calls secondary frame), otherwise it sets the resultant sample equal to 1 (generates silence and calls secondary frame). Changes the Tx DAA state to TX\_DAA\_WRITE\_REG. Finally it sends the resultant sample to the DAA by using out\_sample\_codec ().
- TX\_DAA\_READ\_REG = Prepares a special frame for the Read cycle. The format of the Read frame is
   (0x2000/((0x1F&(channel->daa\_control->reg\_number))<<8)).
   Changes the Tx DAA state to TX\_DAA\_NORMAL. Changes the Rx DAA state to RX\_DAA\_READ\_REG. Calculates the delay between sending the frame and receiving a response. Finally it sends the result frame to the DAA by using out\_sample\_codec ().</li>
- TX\_DAA\_WRITE\_REG = Prepares a special frame for the Write cycle. The format of the Write frame is (((0x1F&(channel->daa\_control->reg\_number))<<8)|(0xFF&(</p>

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channel->daa\_control->value))). Changes the Tx DAA state to TX\_DAA\_NORMAL. Changes the Rx DAA state to RX\_DAA\_WRITE\_REG. Calculates the delay between sending the frame and receiving a response. Finally it sends the resultant frame to the DAA by using out\_sample\_codec().

This function is called by the Interrupt Service Routine of UART1, on a Transmitter ready event.

**Returns:** None.

```
...
struct channel_t channel;
...
Tx_daa(&channel);
...
```

## Rx daa

#### Call(s):

```
void Rx daa (struct channel t * channel);
```

#### **Arguments:**

#### Table 3-91. Rx\_daa arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

#### **Description:**

This function is responsible for receiving data from the DAA. States of operation:

- RX\_DAA\_NORMAL = It reads the received sample from the DAA by using in\_sample\_codec () and writes it into the Rx\_sample[] circular buffer.
- RX\_DAA\_READ\_REG = If daa\_control->delay is equal to 0, it reads the received sample from the DAA by using in\_sample\_codec() and writes it into daa\_control->value, sets daa\_control->value\_is\_true=TRUE and changes the Rx DAA state to RX\_DAA\_NORMAL, otherwise, it decrements daa\_control->delay and reads the received sample from the DAA by using in\_sample\_codec() and writes it into the Rx\_sample[] circular buffer.
- RX\_DAA\_WRITE\_REG = If daa\_control->delay is equal to 0, it reads the received sample from the DAA by using in\_sample\_codec(), sets daa\_control->value\_is\_true=TRUE and changes the Rx DAA state to RX\_DAA\_NORMAL, otherwise, it decrements daa\_control->delay and reads the received sample from the DAA by using in\_sample\_codec() and writes it into the Rx\_sample[] circular buffer.

This function is called by the Interrupt Service Routine of UART1, on a Receiver ready event.

#### **Returns:**

None.

```
...
struct channel_t channel;
...
Rx_daa(&channel);
```

#### 3.7.3 Tone Generator and Detector

The Tone Generator & Detector module provides generation and detection of a variety of signals associated with call progress.

#### 3.7.3.1 Tone Generator

The signals that are produced by the Tone Generator are represented in Table 3-92.

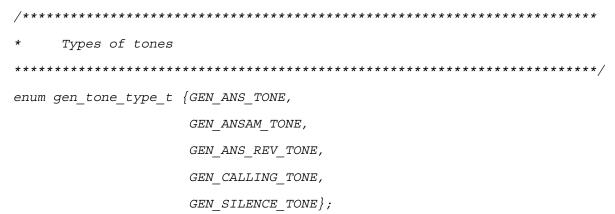


Table 3-92. Tones that are supported by the Tone Generator

Correspondence with gen_tone_type_t type	Signal	Description
GEN_ANS_TONE	Answering tone (ANS)	The tone transmitted from the answering end. The answering tone is a continuous 2100±15 Hz tone with a duration of 3.3±0.7 s.
GEN_ANS_REV_TONE	Answer tone with phase reversals (ANS)	The tone transmitted from the answering end. It consists of a sine wave signal at 2100±15 Hz with phase reversals at an interval of 450±25ms.
GEN_ANSAM_TONE	Modified answer tone (ANSam)	The tone transmitted from the answering end. It consists of a sine wave signal at 2100±15 Hz with phase reversals at an interval of 450±25ms, amplitude-modulated by a sine wave at 15±0.1 Hz. The modulated envelope ranges in amplitude between (0.8±0.01) and (1.2±0.01) times its average amplitude.
GEN_CALLING_TONE	Calling tone	The tone transmitted from the calling end. The calling tone consists of a series of interrupted 1300Hz±15 Hz signals. ON for a duration of not less than 0.7s and OFF for a duration of not less than 1.5s not more than 2.0s
GEN_SILENCE_TONE	Silence	The signal with a zero frequency.

The Tone Generator routines can be found in the "tone\_gendet.c" file.

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The Tone Generator control structure *Tx\_control\_tone\_t* is defined in the "tone\_gendet.h" file:

```
/*****************************
     Tone generator Strucrure
*************************
struct Tx control tone t
   uint32 tone_length_ms;
   uint32 tone pause ms;
   uint32 tone_rev_ms;
   int16 amplitude;
   uint32 tone frequency;
   uint32 omega;
   uint32 phase;
   uint32 tone_samples left;
   uint32 samples left;
   uint32 tone samples rev left
   uint32 amplitude frequency;
   int16 amplitude_ripple;
   uint32 amplitude omega;
   uint32 amplitude phase;
} ;
```

The *Tx\_control\_tone\_t* structure parameter descriptions:

- tone length ms Tone duration (ON) in ms. It is initialized in Tx tone init().
- **tone\_pause\_ms** The duration of silence or a pause (OFF) after tone generation, in ms. It is initialized in *Tx\_tone\_init()*
- **tone\_rev\_ms** Interval of phase reversals, in ms. It is initialized in *Tx\_tone\_init()*
- **amplitude** Amplitude of a generated tone, in Q14 format. It is initialized in *Tx tone init()*.
- **tone\_frequency** Frequency representing the current tone, in Hz. It is initialized in *Tx\_tone\_init()*.
- **omega** Phase shift of the cosine wave, during transfer from one generated sample to next, in Q15. It is calculated from *tone\_frequency* and Sample Rate. It is initialized in *Tx\_tone\_init()*.
- **phase** Current phase, in Q15. It is initialized in *Tx\_tone\_init()* to 0.
- **tone\_samples\_left** Number of samples left to generate current tone only (ON). It is initialized in *Tx\_tone\_init()* to 0.
- **samples\_left** Number of samples left for generating current tone and pause (ON+OFF). It is initialized in *Tx\_tone\_init()* to 0.
- **tone\_samples\_rev\_left** Number of samples left to generate the current phase reversed interval. It is initialized in *Tx\_tone\_init()* to 0.

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- **amplitude\_frequency** Sine wave Frequency for the amplitude-modulated tone, in Hz. It is initialized in *Tx\_tone\_init()*.
- **amplitude\_ripple** Sine wave Ripple Amplitude for the amplitude-modulated tone, in Q14 format. It is initialized in *Tx\_tone\_init()*.
- **amplitude\_omega** Sine wave Phase Shift, during transferring from one generated sample to next, for the amplitude-modulated tone, in Q15 format.
- **amplitude\_phase** Sine wave Current Phase for amplitude-modulated tone, in Q15 format. It is initialized in *Tx\_tone\_init()* to 0.

## Tx\_tone\_init

### Call(s):

```
void Tx_tone_init(struct channel_t * channel,
struct Tx_control_tone_t *Tx_control_tone,
enum gen tone type t tone type );
```

#### **Arguments:**

### Table 3-93. Tx\_tone\_init arguments

channel	in	Pointer to the Channel control data structure
Tx_control_tone	in	Pointer to the Tone Generator control data structure
tone_type	in	Type of tone for generation (see table 3.6.3.1.)

### **Description:**

This function initialises the Tone Generator control data structure according to the chosen type of the tone (*tone\_type*).

It also initializes the fields of the Channel control data structure, pointed to by *channel*, that are responsible for the Tx Data Pump (*data\_pump\_ptr, baud\_rate, number\_samples*,

data\_pump\_call\_func, state, process\_count).

After calling this function, the Tx Data Pump is initialized to work as a Tone generator.

This function is called in *state\_machine()*.

**Returns:** None.

#### ixctui iis.

```
...
struct channel_t channel;
struct Tx_control_tone_t Tx_control_tone;
...
Tx_tone_init(&channel,&Tx_control_tone, GEN_ANSAM_TONE);
...
```

## Tx\_tone

### Call(s):

```
void Tx tone(struct channel_t * channel);
```

#### **Arguments:**

#### Table 3-94. Tx\_tone arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

## **Description:**

This function generates the signal and places the produced samples into the Tx\_sample[] buffer. The number of generated samples, per call, is defined by Tx\_control->number\_samples.

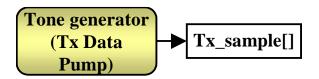


Figure 3-44. Tone generator

The parameters of the signal are defined by the Tx\_tone\_init() function.

This function is called by the  $Tx\_data\_pump()$  function via  $channel->Tx\_control\_ptr-> data\_pump\_call\_func()$ .

**Returns:** None.

```
struct channel_t channel;
...
Tx_tone(&channel);
```

#### 3.7.3.2 Tone Detector

The Tone Detector provides the possibility of parallel detection of three tones. It can detect continuous and cadence tones. It also distinguishes tones, consisting of one and two frequencies.

The signals that are supported by the Tone Detector are represented in Table 3-95.

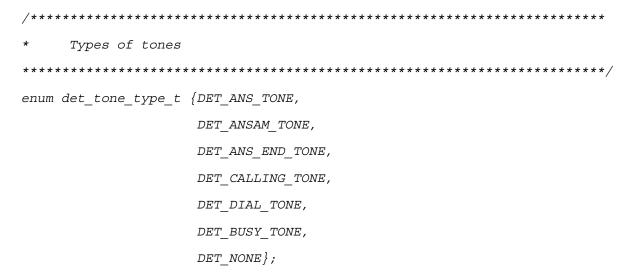


Table 3-95. Tones that are supported by Tone Detector

Correspondence with gen_tone_type_t type	Signal	Description
DET_ANS_TONE	Answering tone without (ANS) or with phase reversals (ANS)	The tone transmitted from the answering end. The detected tone should be a continuous 2100±15 Hz tone.
DET_ANS_END_TONE	Answering tone without (ANS) or with phase reversals (ANS) and pause after it.	The tone transmitted from the answering end. The detected tone should be a continuous 2100±15 Hz tone and with a pause after it.
DET_ANSAM_TONE	Modified answer tone (ANSam)	The tone transmitted from the answering end. The detected tone should be a continuous 2100±15 Hz tone and is amplitude-modulated by a sine wave at 15±0.1 Hz.
DET_CALLING_TONE	Calling tone	The tone transmitted from the calling end. The detected tone should consist of a series of interrupted 1300Hz±15 Hz signals. ON for a duration of not less than 0.7s and OFF for a duration of not less than 1.5s not more than 2.0s

DET_DIAL_TONE	Dial tone	The tone is used in the PSTN to indicate that the telephone network switching equipment has recognized that a telephone has gone off-hook, and is prepared to receive a call. The Tone Detector supports several types of Dial tone (determined by S[24]):  • ex-USSR countries, Brazil: Continuous 425 Hz tone.  • North American countries: Continuous 350Hz + 440Hz tone  • United Kingdom: Continuous 350Hz tone
DET_BUSY_TONE	Busy tone	The tone is used in the PSTN to indicate that the called party is already taking another call. The Tone Detector supports several types of Busy tone (determined by S[24]):  • ex-USSR countries, Brazil: Cadence 425 Hz tone, 0.25 sec. ON, 0.25 sec. OFF.  • North American countries: Cadence 480Hz + 620Hz tone, 0.5 sec. ON, 0.5 sec. OFF.  • United Kingdom: Cadence 400Hz tone, 0.4 sec. ON, 0.4 sec. OFF.
DET_NONE	None	It means the tone detector is disabled

The Tone Detector routines can be found in the "tone\_gendet.c" file.

The scope of the Tone Detector control structure is defined in the "tone\_gendet.h" file:

The biquad filter is represented as:

```
Filter_output = a[0]*input_sample + a[1]*x[0] + a[2]*x[1] - a[3]*y[0] - a[4]*y[1];
```

The *biquad\_param\_t* structure parameter descriptions:

- **a** Array of biquad filter coefficients. It is initialized in Rx\_tone\_init () to values from the filter\_coef\_group[][] depending on the type of tone.
- **x** Array of biquad filter inputs. It is initialized in Rx\_tone\_init () to 0's.

• y - Array of biquad filter outputs. It is initialized in Rx\_tone\_init () to 0's.

```
/***********************
* Double Biquad Filter parameters Structure

**************************

struct double_biquad_param_t
{
   bool enabled;
   struct biquad_param_t filter_1_param;
   struct biquad_param_t filter_2_param;
};
```

The *double\_biquad\_param\_t* structure parameter descriptions:

- **enabled** It is equal to TRUE if the frequency detector is enabled, and to FALSE otherwise. It is initialized in Rx\_tone\_init () depending on the type of tone.
- **filter\_1\_param** Parameters of the first Biquad Filter.
- **filter\_2\_param** Parameters of the second Biquad Filter.

```
States of Tone Detector
*****************************
enum detector_state_t {RX_TONE_ON_DETECTION, RX_TONE_OFF_DETECTION};
/****************************
     Tone detector parameters Structure
#define DET TONE FREQUENCY NUMBER (2)
struct Rx tone t
   struct double biquad param t frequency param[DET TONE FREQUENCY NUMBER];
   enum detector state t detector state;
   int16 y out;
   int16 x out;
   int16 x out max;
   uint32 tone on ms;
   uint32 max_tone_on_ms;
   uint32 tone off ms;
   uint32 detection time;
   uint32 tone on samples left;
   uint32 max_tone_on_samples_left;
   uint32 tone off samples left;
   uint32 detection_saples_left;
   bool tone on detected;
   bool tone all detected;
   uint32 process count;
   bool enabled;
} ;
```

The *Rx\_tone\_t* structure parameter descriptions:

- **frequency\_param** Array of parameters of Double Biquad Filters
- **detector\_state** Determine the state of the Tone detector It can be equal to:
  - RX\_TONE\_ON\_DETECTION = state of Tone ON detection
  - RX\_TONE\_OFF\_DETECTION = state of Tone OFF detection
- y\_out Output of the exponential filter based on the filtered signal. It is initialized in  $Rx\_tone\_init()$  to 0.
- x out Output of the exponential filter based on the global signal. It is initialized in  $Rx\_tone\_init()$  to 0.
- **x\_out\_max** Maximum value of *x\_out*, determined during an ON detection phase. It is initialized in Rx tone init() to 0.
- tone on ms Minimum Tone ON duration in ms (for true detection). It is initialized in Rx tone init().
- max\_tone\_on\_ms Maximum Tone ON duration in ms (for true detection). It is initialized in Rx tone init().
- tone off ms Tone OFF duration in ms (for true detection). It is initialized in  $Rx\_tone\_init()$ .
- **detection time** Maximum time of detection process in ms. It is initialized in Rx tone init().
- tone\_on\_samples\_left Number of samples left for detecting tone ON. It is initialized in Rx tone init() to 0.
- max\_tone\_on\_samples\_left Maximum number of samples for detecting tone ON. It is initialized in Rx tone init() to 0.
- tone off samples left Number of samples left for detecting tone OFF. It is initialized in *Rx\_tone\_init()* to 0.
- **detection\_saples\_left** Number of samples left for the detection process. It is initialized in *Rx\_tone\_init()* to 0.
- tone\_on\_detected It is equal to TRUE if tone ON is detected, and to FALSE otherwise. It is initialized in *Rx\_tone\_init()* to 0.
- tone all detected It is equal to TRUE if a tone is detected (ON and OFF phases), and to FALSE otherwise. It is initialized in Rx tone init() to TRUE.
- **process\_count** Maximum number of phases (ON+OFF) left for the whole tone detection. It is initialized in *Rx\_tone\_init()*.
- enabled It is equal to TRUE if the Tone detector is enabled, and to FALSE otherwise. It is initialized in Rx\_tone\_init ().

Tone detector Control Structure

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#### **Support Modules**

```
*********************************
#define DET_TONE_NUMBER (3)

struct Rx_control_tone_t
{
    struct Rx_tone_t tone[DET_TONE_NUMBER];
};

Rx_control_tone_t is a root structure of the Tone Detector.

Rx_control_tone_t structure parameter descriptions:
```

• **tone** - Array of Tone Detector parameter structure.

## Rx\_tone\_init

#### Call(s):

#### **Arguments:**

#### Table 3-96. Rx\_tone\_init arguments

channel	in	Pointer to the Channel control data structure
Rx_control_tone	in	Pointer to the Tone Detector control data structure
tone	in	Array of tone types (see table 3.6.3.4.)

#### **Description:**

This function initialises the Tone Detector control data structure according to the chosen types of the tones (*tone[]*).

It also initializes the fields of the Channel control data structure, pointed to by *channel*, that are responsible for the Rx Data Pump (data\_pump\_ptr, baud\_rate, number\_samples, data\_pump\_call\_func, state, process\_count).

After calling this function, the Rx Data Pump is initialized to operate according to V.23.

This function is called in *state\_machine()*.

**Returns:** None.

#### ----

```
struct channel_t channel;
struct Rx_control_tone_t Rx_control_tone;
enum det_tone_type_t tone [3];
...
tone[0]=DET_ANS_END_TONE;
tone[1]=DET_BUSY_TONE;
tone[2]=DET_ANSAM_TONE;
...

Rx_tone_init(&channel,&Rx_control_tone, tone);
...
```

# Rx tone

#### Call(s):

void Rx tone(struct channel t \* channel);

### **Arguments:**

Table 3-97. Rx\_tone arguments

C	channel	in	Pointer to the Channel control data structure
---	---------	----	---

# **Description:**

This function detects tones according to the parameters that were initialized in Rx\_tone\_init().

It can perform detection of three tones simultaneously. The tone can consist of one or two frequencies and it can be continuous or cadence.

This function reads samples from Rx\_sample[]. The number of samples is defined by the Rx\_control->number\_samples.

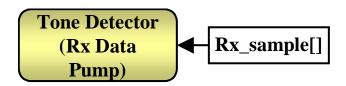


Figure 3-45. Tone detector

The tone detection procedure is divided into different stages as shown in Fig 3.6.3.6.2.

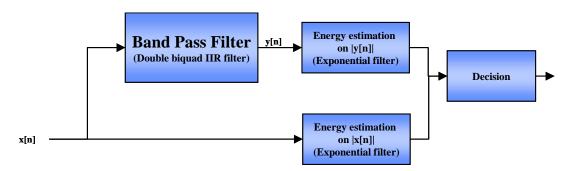


Figure 3-46. Stages of the Tone Detection

The first stage is bandpass filtering (based on double biquad filter) of the input signal. This is followed by an energy estimation by means of exponential filters based on the filtered signal and the global signal. The exponential filters are given by:

 $y_{out}[n] = DET_ALPHA*|y[n]|+(1-DET_ALPHA)*x_{out}[n-1];$  $x_{out}[n] = DET_ALPHA*|y[n]|+(1-DET_ALPHA)*y_{out}[n-1];$ 

The last stage consists of the decision whether a tone has been detected or not.

If one of the tones is detected, or is not detected during the maximum time of the detection process, the *channel->Rx\_control\_ptr->process\_count* becomes equal to 0.

If the actual tone is detected or not is detected, is defined by  $tone_all_detected$  in the structure of  $Rx\_tone\_t$  type.

This function is called by the  $Rx\_data\_pump()$  function via  $channel->Rx\_control\_ptr-> data\_pump\_call\_func()$ .

**Returns:** None.

### **Code example:**

```
struct channel_t channel;
...
Rx_tone(&channel);
```

# 3.7.4 Ring Detector

The Ring detector module uses the serial communication interface of the DAA to receive ring data. When a ring is detected, the samples in the Rx\_sample[] buffer become equal to +32767 while the ring signal is positive, then go back to -32768 when the ring is near zero and negative. Thus a near square wave is presented in Rx\_sample[] that swings from -32768 to 32767 in cadence with the ring signal.

The Ring detector module routines can be found in the "ring\_det.c" file.

The Ring detector control data structure Rx\_ring\_det\_t is defined in "ring\_det.h"

The *Rx\_ring\_det\_t* structure parameter descriptions:

- **number\_on** Current number of positive ringing signals. It is initialized in  $Rx\_ring\_det\_init()$  to 0.
- **number\_off** Current number of negative ringing signals. It is initialized in  $Rx\_ring\_det\_init()$  to 0.

# Rx\_ring\_det\_init

## Call(s):

### **Arguments:**

# Table 3-98. Rx\_ring\_det\_init arguments

channel	in	Pointer to the Channel control data structure
Rx_ring_det	in	Pointer to the Ring detector control data structure

#### **Description:**

This function initialises the Ring detector control data structure. It also initializes the fields of the Channel control data structure, pointed to by *channel*, that are responsible for the Rx Data Pump (data pump ptr, baud rate, number samples,

data\_pump\_call\_func, state, process\_count).

After calling this function, the Rx Data Pump is initialized to work

according to V.23.

This function is called in *state\_machine()*.

**Returns:** None.

```
struct channel_t channel;
struct Rx_ring_det_t Rx_ring_det;
...
Rx_ring_det_init(&channel,&Rx_ring_det);
...
```

# Rx\_ring\_detect

### Call(s):

void Rx ring detect(struct channel t \* channel);

#### **Arguments:**

### Table 3-99. Rx\_ring\_detect arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

#### **Description:**

This function detects Ring signals. It uses the features of the DAA. When a ring is detected, the samples in the Rx\_sample[] buffer become equal to +32767 while the ring signal is positive, then go back to -32768 while the ring is near zero and negative. Thus a near square wave is presented in Rx\_sample[] that swings from -32768 to 32767 in cadence with the ring signal.

This function handles the received sample from the Rx\_sample[] buffer in order to detect a true Ring signal.

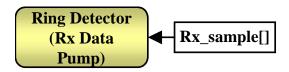


Figure 3-47. Ring Detector

This function is called by the  $Rx\_data\_pump()$  function via  $channel->Rx\_control\_ptr-> data\_pump\_call\_func()$ .

**Returns:** None.

# **Code example:**

```
...
struct channel_t channel;
...
Rx_ring_detect (&channel);
...
```

# 3.7.5 DTMF Dialer

The DTMF dialer implements the DTMF generator.

#### **Support Modules**

Dual Tone Multi-Frequency or DTMF is a method for instructing a telephone switching system of the telephone number to be dialed.

The DTMF system uses eight different frequency signals transmitted in pairs to represent sixteen different numbers, symbols and letters. Table 3-100 shows how the frequencies are organized.

	•		•	`	-
		Co	lumn Freq	uency Grou	lb dr
		1209Hz	1336H z	1477Hz	1633H z
	697Hz	1	2	3	Α
Row Frequency Group					
	770Hz	4	5	6	В
	852Hz	7	8	9	С
	941Hz	*	0	#	D

Table 3-100. Frequencies used by the DTMF generator

The DTMF Dialer module routines can be found in the "dtmf.c" file.

The DTMF control data structure *Tx\_DTMF\_control\_t* is defined in "dtmf.h"

```
/****************************
* DTMF generator control data structure
************************

typedef struct Tx_DTMF_control_t
{
    uint32 digit_length_ms;
    uint32 digit_pause_ms;

    uint16 omega_row;
    uint16 omega_column;
    uint16 phase_row;
    uint16 phase_column;
    uint16 amplitude;
};
```

The *Tx\_DTMF\_control\_t* structure parameter descriptions:

- **digit\_length\_ms** Duration of digit, in ms. It is initialized in *Tx\_DTMF\_init()*.
- **digit\_pause\_ms** Duration of the pause between digits, in ms. It is initialized in *Tx\_DTMF\_init()*.

- **omega\_row** Row frequency Phase Shift per sample, in Q15. It is initialized in  $Tx\_DTMF\_init()$  to 0.
- **omega\_column** Column frequency Phase Shift per sample, in Q15. It is initialized in  $Tx\_DTMF\_init()$  to 0.
- **phase\_row** Row frequency Phase, in Q15. It is initialized in  $Tx_DTMF_init()$  to 0.
- **phase\_column** Column frequency Phase, in Q15. It is initialized in  $Tx\_DTMF\_init()$  to 0.
- **samples\_left** Number of samples left to generate the current digit or a pause. It is initialized in *Tx\_DTMF\_init()* to 0.
- **amplitude** Amplitude of the generated signal, in Q14. It is initialized in *Tx DTMF init()*.

# Tx\_DTMF\_init

## Call(s):

#### **Arguments:**

# Table 3-101. Tx\_DTMF\_init arguments

channel	in	Pointer to the Channel control data structure
Tx_DTMF_control	in	Pointer to the DTMF generator control data structure

# **Description:**

This function initialises the DTMF generator control data structure. It also initializes the fields of the Channel control data structure, pointed to by *channel*, that are responsible for the Tx Data Pump (data\_pump\_ptr, baud\_rate, number\_samples,

data\_pump\_call\_func, state).

After calling this function, the Tx Data Pump is initialized to work as a DTMF generator.

This function is called in *state\_machine()*.

**Returns:** 

None.

```
...
struct channel_t channel;
struct Tx_DTMF_control_t Tx_DTMF_control;
...
Tx_DTMF_init (&channel, &Tx_DTMF_control);
...
```

# Tx DTMF

#### Call(s):

```
void Tx DTMF(struct channel t * channel);
```

### **Arguments:**

#### Table 3-102. Tx\_DTMF arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

#### **Description:**

This function reads the digits (symbols) for DTMF dialing from the Tx\_data[] buffer, until the number of digits equals *channel->Tx\_control\_ptr->process\_count*, and finds the corresponding digit (or symbol) to the row and column frequency according to table 3.6.5. The generator of a DTMF signal simultaneously sends one frequency from the high-group and one frequency from the low group. For example, sending 1209Hz and 770Hz indicates that the '4' digit is being sent.

If the symbol in Tx\_data[] is the wrong digit or is equal to the ',' symbol, it generates a pause.

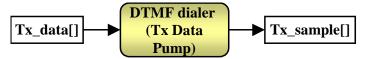


Figure 3-48. DTMF dialer

The pause between digits is defined by the S[11] register, and the pause on comma ',' symbol is defined the by S[8] register.

This function generates the signal and places the produced samples into the Tx\_sample[] buffer. The number of generated samples, per call, is defined by Tx\_control->number\_samples.

This function is called by the  $Tx\_data\_pump()$  function via  $channel->Tx\_control\_ptr->data\_pump\_call\_func()$ .

#### **Returns:**

None.

```
struct channel_t channel;
...
Tx_DTMF (&channel);
...
```

# 3.7.6 Pulse Dialer

Pulse dialing is one of the methods for instructing a telephone switching system of the telephone number to be dialed. Old-style rotary dial phones use this method.

Pulse dialing is accomplished by going off- and on-hook to generate make and break pulses. The nominal rate is 10 pulses per second. It can dial numbers from '0' to '9' only.

The Pulse Dialer module routines can be found in the "pulse.c" file.

The Pulse Dialer control data structure *Tx\_pulse\_control\_t* is defined in "pulse.h"

The *Tx\_pulse\_control\_t* structure parameter descriptions:

- make\_ms Duration of Make (off-hook), in ms. It is initialized in *Tx\_pulse\_init()*.
- **break\_ms** Duration of Break (on-hook), in ms. It is initialized in *Tx\_pulse\_init()*.
- **interval\_ms** Time interval between digits, in ms. It is initialized in *Tx\_pulse\_init()*.
- **samples\_left** Number of samples left to generate the pause between digits or the pause between Make and Break. It is initialized in *Tx\_pulse\_init()* to 0.
- **pulse\_left** Number of pulses left to generate the current digit. It is initialized in *Tx pulse init()* to 0.
- **next\_break** It is equal to TRUE if the next action is Break and to FALSE otherwise. It is initialized in *Tx\_pulse\_init()* to TRUE.

# Tx\_pulse\_init

## Call(s):

# **Arguments:**

#### Table 3-103. Tx\_pulse\_init arguments

channel	in	Pointer to the Channel control data structure
Tx_pulse_control	in	Pointer to the Pulse dialer control data structure

#### **Description:**

This function initialises the *Tx\_pulse\_control* Pulse dialer control data structure. It also initializes the fields of the Channel control data structure, pointed to by *channel*, that are responsible for the Tx Data Pump (*data\_pump\_ptr*, *baud\_rate*, *number\_samples*,

data pump call func, state).

After calling this function, the Tx Data Pump is initialized to work

as a Pulse dialer.

This function is called in *state\_machine()*.

**Returns:** None.

```
...
struct channel_t channel;
struct Tx_pulse_control_t Tx_pulse_control;
...
Tx_pulse_init (&channel, &Tx_pulse_control);
...
```

# Tx\_pulse

#### Call(s):

Tx\_pulse(struct channel\_t \* channel);

#### **Arguments:**

Table 3-104. Tx pulse arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

### **Description:**

This function reads the digits for Pulse dialing from the Tx\_data[] buffer, until the number of digits equals to the *channel->Tx\_control\_ptr->process\_count*. It performs the off- and on-hook signals to generate make and break pulses. The number of make/break pulses directly depends on the digit, see Table 3-105.

Table 3-105. Correspondence of dial digits to the number of Make/Break pulses

Digit	Number of Make/Break pulses
0	10
1	1
2	2
3	3
4	4
5	5
6	6
7	7
8	8
9	9

If the symbol in Tx\_data[] is the wrong digit or is equal to the ',' symbol, it generates a pause.

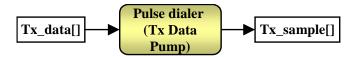


Figure 3-49. Pulse dialer

The pause between digits is defined by the S[15] register, and the pause on a comma ',' symbol is defined by the S[8] register.

This function generates the signal and places the produced samples into the Tx\_sample[] buffer. The number of generated samples, per call, is defined by Tx\_control->number\_samples.

**Support Modules** 

This function is called by the  $Tx\_data\_pump()$  function via  $channel->Tx\_control\_ptr-> data\_pump\_call\_func()$ .

**Returns:** None.

## **Code example:**

```
...
struct channel_t channel;
...
Tx_pulse (&channel);
```

# 3.7.7 Text Response

The text response module is a set of functions that are responsible for informative messaging to the DTE.

# print\_result\_code

## Call(s):

void print result code (struct channel t \* channel, uint8 code);

#### **Arguments:**

#### Table 3-106. print\_result\_code arguments

channel	in	Pointer to the Channel control data structure
code	in	Numeric code of the Result code

#### **Description:**

This function writes the Result code defined by *code*, according to V.25ter, into the Rx\_uart\_data[] buffer.

Result codes consist of three parts: a header, the result text, and a trailer. The characters transmitted for the header and trailer are determined by a user setting. The text may be transmitted as a number or as a string, depending on a user-selectable setting (see Table 3-107).

Table 3-107 shows the effect of the setting of the AT Vn command on the format of the result code. All references to **<cr>>** mean "the character with the ordinal value specified in parameter **S3**"; all references to **<lf>>** likewise mean "the character with the ordinal value specified in parameter **S4**".

Table 3-107. Effect of the AT V command on the Result code format

	AT V0	AT V1
Result codes	<numeric code=""><cr></cr></numeric>	<cr><lf><verbose code=""><cr><lf></lf></cr></verbose></lf></cr>

All numeric and verbose codes are defined in "text\_response.h".

**Returns:** None.

```
...
#include "text_response.h"
struct channel_t channel;
...
print_result_code (&channel, RESPONSE_CODE_CONNECT);
...
```

# print\_text\_response

## Call(s):

```
void print text response (struct channel t * channel, char * text);
```

### **Arguments:**

#### Table 3-108. print\_text\_response arguments

channel	in	Pointer to the Channel control data structure	
text	in	Text of the text response	

#### **Description:**

This function writes the Text response defined by *text*, according to V.25ter, into the Rx\_uart\_data[] buffer.

Information text responses consist of three parts: a header, text, and a trailer. The characters transmitted for the header, text and trailer are determined by a user setting. Table 3-109 shows the effect of setting the AT Vn command to the Information text format. All references to <**cr**> mean "the character with the ordinal value specified in parameter **S3**"; all references to <**lf**> likewise mean "the character with the ordinal value specified in parameter **S4**".

Table 3-109. Effect of AT Vn command on Text response format

	AT V0	AT V1
Information responses	<text><cr><lf></lf></cr></text>	<cr><lf><text><cr><lf></lf></cr></text></lf></cr>

Information text returned in response to manufacturer-specific commands may contain multiple lines, and the text may therefore include CR, LF, and other formatting characters to improve readability.

This function is used by the AT Command handler.

**Returns:** None.

```
...
struct channel_t channel;
...
print_text_response (&channel, "LDR SoftModem");
...
```

# print\_connect

# Call(s):

```
void print_connect (struct channel_t * channel);
```

#### **Arguments:**

### Table 3-110. print\_connect arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** 

This function writes the CONNECT result code, according to

V.25ter, to the Rx\_uart\_data[] buffer.

Table 3-111 shows the effect of setting the AT Xn command to the format of the CONNECT result code. For printing the CONNECT result code the *print\_result\_code()* function is used.

Table 3-111. Effect of the AT Xn command on the CONNECT result code

	AT X0	AT X1; X2; X3; X4
Result code	CONNECT	CONNECT <text></text>

If the result code is equal to CONNECT <text>, the function prints the speed of the Tx and Rx Data Pumps, and the parameters of the correction and compression protocols.

The Global State Machine calls this function upon entering the online data state.

**Returns:** None.

```
...
struct channel_t channel;
...
print_connect (&channel);
...
```

# text\_to\_dte

# Call(s):

```
void text to dte (struct channel t * channel, char *str);
```

# **Arguments:**

### Table 3-112. text\_to\_dte arguments

channel	in	Pointer to the Channel control data structure
str	in	Null-terminated string

**Description:** This function writes the null-terminated string (*str*) into the Rx\_uart\_data[] buffer.

Returns: None.

```
...
struct channel_t channel;
...
text_to_dte (&channel, "Text message");
...
```

# text\_to\_dte\_fixed

# Call(s):

### **Arguments:**

#### Table 3-113. text\_to\_dte\_fixed arguments

channel	in	Pointer to the Channel control data structure
str	in	String
length	in	The length of the str string

**Description:** 

This function writes the string (str) of the fixed length (length) into

the Rx\_uart\_data[] buffer.

**Returns:** None.

```
...
struct channel_t channel;
char tmp_str [2];
tmp_str[0]='O';
tmp_str[1]='K';
...
text_to_dte_fixed (&channel, tmp_str, 2);
...
```

# char\_to\_dte

# Call(s):

void char to dte (struct channel t \* channel, char ch);

#### **Arguments:**

Table 3-114. char\_to\_dte arguments

channel	in	Pointer to the Channel control data structure
ch	in	Character

**Description:** This function writes one character (*ch*) into the Rx\_uart\_data[]

buffer.

**Returns:** None.

```
...
struct channel_t channel;
...
char_to_dte (&channel,'a');
...
```

# uint8\_to\_str\_decimal

# Call(s):

uint8 uint8 to str decimal (char \*num str, uint8 num);

#### **Arguments:**

### Table 3-115. uint8\_to\_str\_decimal arguments

num_str	in/out	Pointer to an output null-terminated string
num	in	Unsigned number

#### **Description:**

This function converts the unsigned number (*num*) into a null-terminated string in the decimal form and places it into the string pointed to by *num\_str*. For example, if *num=128*, then *num\_str="128"*.

#### **Returns:**

The number of characters in the (*num\_str*) string for representing the input number in the decimal form. For the example above it returns 3.

```
...
char tmp_str [4];
uint8 digit_num;
...
digit_num=uint8_to_str_decimal (tmp_str, 100);
...
```

# uint8\_to\_str\_hexadecimal

# Call(s):

```
uint8 uint8 to str hexadecimal (char *num str, uint8 num);
```

#### **Arguments:**

#### Table 3-116. uint8\_to\_str\_hexadecimal arguments

num_str	in/out	Pointer to an output null-terminated string
num	in	Unsigned number

**Description:** This function converts the unsigned number (*num*) into a

null-terminated string in the hexadecimal form and places it into the

string pointed to by num str. For example, if num=43, then

*num\_str="2B"*.

**Returns:** The number of characters in the (num\_str) string for representing the

input number in the hexadecimal form. For the example above it

returns 2.

#### **Code example:**

```
...
char tmp_str [3];
uint8 digit_num;
...
digit_num= uint8_to_str_hexadecimal (tmp_str, 100);
...
```

# 3.8 Miscellaneous Functions

The miscellaneous routines can be found in the "misc.c" file. These functions are widely used in the LDR Soft Modem.

# dsp\_cos

# Call(s):

int16 dsp cos(uint16 phase);

#### **Arguments:**

### Table 3-117. dsp\_cos arguments

phase	in	Phase in Q15 format (0 corresponds to 0 degrees; 32768 (1.0 in Q15) corresponds to 360 degrees)
-------	----	---

**Description:** 

This function calculates the cosine (2\*pi\*phase). The phase is fractional in Q15 format.

**Returns:** Result of the cos(2\*pi\*phase) calculation.

```
...
int 16 phase=(1<<15);
...
dsp_cos (phase);
...</pre>
```

# dsp\_convolution\_frac

Call(s):

int16 dsp convolution frac (const int16 \*tab1, const int16 \*tab2, uint32 n);

### **Arguments:**

Table 3-118. dsp\_convolution\_frac arguments

tab1	in	Pointer to the first vector of the fractional data elements
tab1	in	Pointer to the second vector of the fractional data elements
n	in	Number of elements in one vector

**Description:** 

This function computes a convolution for a vector of fractional data values:

$$result = \sum_{k=0}^{\infty} tab1[k] * tab2[k]$$

The function is used for realization of the Finite Impulse Response (FIR) filters.

**Returns:** 

Result of the  $\sum_{tab1[k]*tab2[k]}$  calculation.

```
...
#define TAB_SIZE (16)
int16 tab1[TAB_SIZE];
int16 tab2[TAB_SIZE];
int16 result_convolution;
...
result_convolution = dsp_convolution_frac(tab1, tab2, TAB_SIZE);
...
```

# memcpy\_mcf

#### Call(s):

```
void memcpy mcf (void *dest, void *src, uint32 n);
```

#### **Arguments:**

#### Table 3-119. memcpy\_mcf arguments

dest	in	Pointer to the destination
src	in	Pointer to the source
n	in	Number of bytes

**Description:** 

This function copies n bytes pointed to by (src) to the destination to pointed by (dest)

**Returns:** None.

### **Code example:**

```
#define TAB_SIZE (16)
int16  dest[TAB_SIZE];
int16  src[TAB_SIZE];
...
memcpy_mcf(dest, src, TAB_SIZE* sizeof(int16));
...
```

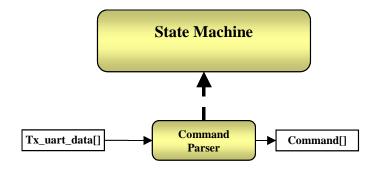
# 3.8.1 AT command set support

The LDR Soft Modem implements the AT command set. The AT command set is the user interface, and in an embedded application, it is the basis of the API. The AT commands allow the user to alter the operation of the Soft Modem in real-time. These commands are sent to the system via the UART from an RS-232 connected terminal device.

Please refer to the "MCF5407 Low Data Rate Soft Modem. AT Command reference. User Manual" for a detailed description of the AT commands and S registers that are supported by the LDR Soft Modem.

In the LDR Soft Modem, an interpretation of the AT commands is performed by two software blocks, the Command Parser and the Command Handler.

# 3.8.2 Command Parser



The at\_parser\_control\_t structure, is defined in "at\_parser.h".

```
*Definition of AT parser structure
struct at parser_control_t
  uint32 at_command_buf[BUF_LENGTH];
                                          //AT_command local buffer
  int8 at buf ptr;
                                                //pointer to at command buf[]
  uint8 at_command_string[BUF_LENGTH];
                                               //AT command enter string //buffer
                                                       //Dial string buffer
  uint8 dial digits[DIAL LENGTH];
  int8 digits number;
  int8 dial;
                               //Used if sign ";" finishes the dial
                                   // string
  int16 backspace counter;
  uint32 sample counter;
  uint8 plus counter;
  uint8 * uart_data_ptr;
  int16 string counter;
  uint32 command code;
  enum parser state t parser state;
  enum sm parser state t sm parser state;
};
```

The at\_parser\_control\_t structure parameter descriptions:

- at\_command\_buf[] Local circular buffer that the AT Parser writes all AT Commands into after processing the received AT command string. After processing the received AT commands, the AT Parser writes competent commands from the at\_command\_buf[] buffer into the Tx\_control->command[] buffer for the AT Handler to process.
- at\_buf\_ptr Pointer to the at\_command\_buf[] buffer.
- **at\_command\_string[]** Array, into which the AT Parser stores all characters received following "AT" characters.

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- **dial\_digits** Array, into which the AT Parser stores dial number characters.
- **digits\_number** Number of characters in the dial\_digits array.
- **dial** This variable contains "1", if the ";" sign has occurred in the command string. It is required for executing of the commands following the ";" sign in the next call of the Command Parser only.
- **backspace\_counter** Counter to the at\_commands\_string[] array.
- **sample\_counter** Sample counter. This is used for calculating the time in two cases: receiving the second character of the "Enter" symbol ("Enter" contains 2 characters CR and LF (see at\_parser.h defined values)); or receiving the escape sequence (see LDR Soft Modem User Manual for escape sequence description).
- **plus\_counter** –Escape sequence symbol counter. The default value of the escape sequence symbol is "+".
- **uart\_data\_ptr** Special pointer to the Tx\_uart\_data[] buffer for monitoring the escape sequence.
- **string\_counter** Pointer to the at\_command\_string[] array, this shows which element is required to begin the processing of the at\_command\_string[] array.
- **command\_code** This variable contains the AT command and the parameters of this command to be written into the at\_command\_buf[].
- **parser\_state** State identifier. It contains the current substate of the AT Parser. It is initialized in AT\_parser\_init() to ATP\_WAIT\_A.
- **sm\_parser\_state** State identifier. It contains the current state of the AT Parser. It is initialized in AT\_parser\_init() to ATP\_COMMAND\_MODE.

# AT\_parser\_init

### Call(s):

void AT\_parser\_init (struct channel\_t \* channel, struct at\_parser\_control\_t
\*at parser control);

#### **Arguments:**

# Table 3-120. AT\_parser\_init arguments

channel	in	Pointer to the Channel control data structure
at_parser_control	in	Pointer to the AT Parser data structure

**Description:** Initialisation of the AT Command Parser. The function initialises the

channel->Tx\_control\_ptr->command\_handler\_call \_func to AT\_parser\_routine. It must be called before the

AT parser routine() function.

This function is called by the Global State Machine.

**Returns:** None.

```
...
struct at_parser_control_t *at_parser_control;
...
struct channel_t channel;
...
AT_parser_init (channel, &at_parser_control);
...
```

# AT\_parser\_routine

# Call(s):

void AT parser routine (struct channel t \* channel);

#### **Arguments:**

#### Table 3-121. AT\_parser\_init arguments

channel in	Pointer to the Channel control data structure
------------	---

**Description:** This function implements the AT Command Parser.

It gets characters from the tx\_uart\_data[] buffer, interprets the AT Commands and writes the interpreted AT Command into the command[] buffer.

This function is called by the command handler () routine.

**Returns:** None.

# **Code example:**

```
struct channel_t channel;

...

AT_parser_init (channel, &at_parser_control);

...

AT_parser_routine(&channel);
```

### **Code explanation:**

The state identifier sm\_parser\_state can be in one of three modes: ATP\_COMMAND\_MODE, ATP\_ONLINE COMMAND\_MODE, ATP\_ONLINE\_DATA\_MODE.

In the case of ATP\_COMMAND\_MODE and ATP\_ONLINE\_COMMAND\_MODE the AT Parser can process the AT Commands received from the tx\_uart\_data[] buffer, and store them at first in the local at\_command\_buf[] buffer. After processing all the AT Commands string they are then stored in the command[] buffer.

In these cases the AT Parser operation depends on the parser\_state state identifier:

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case ATP\_WAIT\_A:

Monitors the tx\_uart\_data[] buffer to determine an "a" or "A" character. If this character has occurred, the AT Parser goes to the ATP\_WAIT\_T state. If another character was occurred, the AT Parser will ignore it.

case ATP\_WAIT\_T:

Monitors the tx\_uart\_data[] buffer to determine a "t", "T" or "\"character. If a "t" or "T" character has occurred, the AT Parser goes to the ATP\_WRITE\_BUFFER state. If "\" character has occurred, the AT Parser goes to the ATP\_BUFFER\_HANDLING state to repeat the execution of the previous AT Command string. If another character has occurred, the AT Parser goes back to the ATP\_WAIT\_A state.

case ATP\_WRITE\_BUFFER:

In this case the AT Parser stores all the characters received from the tx\_uart\_data[] buffer after receiving the "AT" characters until receiving the "Enter" symbol in the <code>at\_command\_string</code>. If the "Enter" character has occurred, the AT Parser goes to the <code>ATP CR WAIT</code> state.

The received AT command string can't be more than 40 character (END\_STRING=40). The characters, following the 40<sup>th</sup> character in the string will be ignored and AT Parser will go to the ATP\_BUFFER\_HANDLING state.

case ATP\_CR\_WAIT:

This case is used for processing the second "Enter" symbol ("Enter" symbol can contain two characters – CR and LF (their values are specified in the modem.h file)). If the "Enter" symbol contains two characters, the second character will be read from tx\_uart data. The AT Parser then goes to the ATP\_BUFFER\_HANDLING state.

case ATP\_BUFER\_HANDLING:

In this case the AT Parser at first determines the number of characters in the at\_command\_string and ignores all space characters.

Then the AT Parser interprets commands and parameters in the at\_command\_string according to the V.25ter ITU-T specification, stores the interpreted commands into the local at\_command\_buf[] buffer, and if no error has occurred, places the contents of the at\_command\_buf[] buffer into the command[] buffer for these commands to be handled by the AT Command Handler.

In the case of ATP\_ONLINE\_DATA\_MODE the AT Parser only monitors the tx\_uart\_data[] buffer to determine the escape sequence.

# command\_parameter\_identification

## Call(s):

int16 command\_parameter\_identification(struct channel\_t \* channel, int16
current k, int16 min, int16 max, uint32 command code);

### **Arguments:**

### Table 3-122. AT\_parser\_init arguments

channel	in	Pointer to the Channel control data structure
current_k	in	Pointer to the element in the at_command_string that must to be processed.
min	in	Minimal value of parameter
max	in	Maximal value of parameter
command_code	in	The variable contains the number and parameters of the current AT command.

**Description:** This function determines whether the parameter following the

command in at\_command\_string doesn't exceed the bounds of minimal and maximal parameter values. If the parameter doesn't

exceed these bounds the function stores it in the variable command\_code together with the stored command, if not the function writes in the variable command\_code an "ERROR".

**Returns:** Pointer to the element in the at\_command\_string that must be

processed.

```
...
struct channel_t channel;
...
int16 k=0;
...
#define E_RANGE_MIN 0
#define E_RANGE_MAX 1
...
k=command_parameter_identification(channel, k, E_RANGE_MIN, E_RANGE_MAX, at_parser_control->command_code);
...
```

# command\_buf\_write

## Call(s):

uint32 command buf write (struct channel\_t \* channel, uint32 command\_code);

#### **Arguments:**

## Table 3-123. AT\_parser\_init arguments

channel	in	Pointer to the Channel control data structure
command_code	in	The variable contains number and parameters of the current AT command.

# **Description:**

This function stores the interpreted AT command and its parameters in the local at\_command\_buf[] buffer. After processing all of the AT Command string and if no error has occurred the contents of at\_command\_buf[] buffer will be placed into the command[] buffer.

**Returns:** Zero.

```
...
struct channel_t channel;
...
uint32 command_code;
...
command_buf_write(channel, at_parser_control->command_code);
...
```

# 3.8.3 Command Handler

The Command Handler is a top-level entity that is invoked by the scheduler to interpret AT commands and execute them. It reads formatted commands from the command[] circular buffer, that are written into it by the Command Parser.

The Command Handler implements the AT commands and controls the state of the Soft Modem. It also passes the command responses and status messages to the DTE. See section 3.7.2.

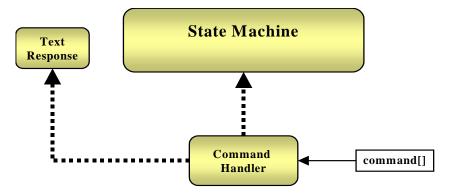


Figure 3-50. Command Handler

The Command Handler routines and definitions can be found in the "at\_handler.h" and "at\_handler.c" files.

# AT\_handler\_init

## Call(s):

```
void AT handler init (struct channel t * channel);
```

### **Arguments:**

#### Table 3-124. AT\_handler\_init arguments

channel	in	Pointer to the Channel control data structure
---------	----	---

**Description:** This function initialises the

channel->Tx\_control\_ptr->command\_handler\_call \_func to the AT\_handler\_routine. It must be called before the command handler() function.

This function is called in *state\_machine()*.

**Returns:** None.

```
...
struct channel_t channel;
...
AT_handler_init (&channel);
...
```

# AT\_handler\_routine

Call(s):

void AT handler routine(struct channel t \* channel);

#### **Arguments:**

### Table 3-125. AT\_handler\_routine arguments

channel in	Pointer to the Channel control data structure
------------	---

#### **Description:**

This function implements the AT Command Handler.

It reads commands, interpreted by the AT Command Parser, from the command[] buffer. The format of the command[] buffer element is:

- 4-th byte is the command number (Command numbers are specified in the "at\_parser.h" file.)
- 3-rd byte is the 1st parameter (optional)
- 2-nd byte is the 2nd parameter (optional)
- 1-st byte is the 3rd parameter (optional)

Using parameters is optional and depends on a concrete command. Depending on the AT commands, the Command Handler can:

- Modify S registers
- Control minimum and maximum values of the S registers
- Change the Global State Machine's current state
- Change the H registers
- Send informative messages to the DTE
- Send command response to the DTE

If the command number or any of the parameters is wrong the Command Handler breaks commands handling, clears the command[] buffer and sends an "ERROR" response to the DTE.

This function is called by the *command\_handler()* function via *channel->Tx\_control\_ptr-> command\_handler\_call\_func()*.

For more information about AT Command Handler actions on AT commands, please refer to the "MCF5407 Low Data Rate Soft Modem. AT Command reference. User Manual".

**Returns:** None.

#### **Miscellaneous Functions**

```
...
struct channel_t channel;
...
AT_handler_init (&channel);
...
AT_handler_routine(&channel);
```

# choose\_data\_pump\_protocol

Call(s):

void choose data pump protocol(void);

**Arguments:** None.

**Description:** This function controls the value of the H[3] register (Current Data

Pump Protocol). If the H[3] register is equal to H3\_DP\_NONE (Data Pump protocol is not chosen), the function makes a decision about the Data Pump protocol on the basis of the S[19] register (Data Pump protocol defined by user), after that it changes the H[3] register to

correspond to that Data Pump protocol.

This function is called in *state\_machine()*.

**Returns:** None.

**Code example:** 

. .

choose\_data\_pump\_protocol();

. . .

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